# Linear Dynamical Systems with Sparsity Constraints: Theory and Algorithms

A thesis submitted in partial fulfilment of the requirements for the degree of **Doctor of Philosophy** 

in the Faculty of Engineering

by Geethu Joseph



Electrical Communication Engineering Indian Institute of Science, Bangalore Bangalore – 560 012 (INDIA)

2019

To Daniel and Hannah

## Acknowledgments

My Ph.D journey has been an extremely inspiring and enjoyable experience, a time of fantastic intellectual and personal growth. I feel very fortunate to receive adequate support and encouragement from my advisor, my family, and friends. They helped me to achieve a healthy work-life balance.

Firstly, I would like to express my sincere gratitude to my advisor *Prof. Chandra R. Murthy* for his wholehearted technical and moral support, patience, and valuable guidance. He initiated me into the world of research, and gave me enormous freedom to work on my own research problems. He taught me how to pursue research and present results elegantly. Our weekly progress meetings helped me to keep motivated when I stumped about the direction of my research. I also appreciate his assistance in writing scholarship application for Intel fellowship and this thesis.

I thank my collaborators *Ranjitha Prasad* who is currently at Tata Consultancy Services Innovation Labs, Delhi, and *Prof. Basker D. Rao* who is at the University of California, San Diego. I am indebted to the faculty members of ECE, EE, CSA and Math departments, where I have attended many interesting courses. I thank the members of ECE department, especially, *Ms. Suma* and *Mr. Srinivasa Murthy* for all the administrative help. I acknowledge with gratitude Intel for awarding me a fellowship which supported my Ph.D. and conference travels.

I have been blessed with a very loving and supportive family. My heart-felt gratitude goes to my wonderful parents *Sherly* and *Joseph* who have been a source of constant and unconditional love and inspiration. They have always supported me in my decisions and encouraged me to pursue my dream. I thank with love my husband *Jomon* with whom I shared the most part of this amazing journey. I appreciate his patient listening, timely advice, and sacrifices to make my career a priority in our lives. I am grateful to my little bundles of joy and laughter: *Daniel* and *Hannah*. Their bright smiles put it all into perspective for me. A special thanks go to my kind and helpful brother *Jitthu* who had happily ran a hundred errands in IISc for me. I owe a debt of gratitude to my enthusiastic grandmother, *Mary*. Despite her old age, she stayed with me and gave invaluable support during my pregnancy and early days of parenthood. I also extend my deepest gratitude to my mother-in-law *Gracykutty* mainly for taking care of my babies very well.

Finally, I would like to convey sincere gratitude to all my IISc friends. I am privileged to have shared the lab space with Saurabh, Partha, Mohit, Shilpa, Jobin, Thirupathaiah, Suma, Praveen and Ribhu. The stimulating discussions, lunch and ice cream treats, and all the fun we have had will be cherished all my life. I also thank Saurabh, Mohit, Ranjitha, Abhay, Venu and other members of SPC lab for the invaluable technical exchanges. I offer gratitude to all my hostel mates, particularly, Arathi, Indu, Jeenu, Arathi, Nazreen, Vandhana, Anu, Priti, Chenju and Seena. Our long chats in the C-mess hall, birthday celebrations, and photo sessions made my hostel life colorful. Also, my friend Rahul deserves a special mention for all the interesting times together.

Thanks for all for your encouragement!

## Abstract

This thesis develops new mathematical theory and presents novel recovery algorithms for discrete linear dynamical systems (LDS) with sparsity constraints on either control inputs or initial state. The recovery problems in this framework, manifest as the problem of reconstructing one or more sparse signals from a set of noisy underdetermined linear measurements. The goal of our work is to design algorithms which can exploit the underlying structure in the measurement matrix and the unknown sparse vectors, and to reveal the impact of these structures on the efficacy of the recovery.

We answer three fundamental and interconnected questions on sparse signal recovery problems that arise in the context of LDS. First, what are the necessary and sufficient conditions for the existence of a sparse solution? Second, given that a sparse solution exists, what are good low-complexity algorithms that exploit the underlying signal structure? Third, when are these algorithms guaranteed to succeed? These questions are considered in the context of three different sparsity models as described below.

Within the LDS framework, we first consider the simplest sparsity model of a single unknown sparse initial state vector with no additional structure. This problem is known as the observability problem in the control theory literature, and the initial state can be recovered using standard compressed sensing (CS) algorithms. However, the recovery guarantees for this case are different from the classical sparse recovery guarantees because the measurement matrix that arises in LDS is fundamentally different from the matrices that are typically considered in the CS literature. Hence, we seek the conditions for observability of LDS when the initial state is sparse and the observation matrix is random. Taking advantage of randomness in the measurements, we use concentration inequalities to derive upper bound on the minimum number of measurements that can ensure the faithful recovery of the sparse initial state.

#### Abstract

Next, we move to a more complicated sparsity model, which is concerned with the recovery of a set of sparse control input vectors. In this setting, we first derive necessary and sufficient conditions for the existence of a sparse solution for any given pair of initial and final states in the LDS. These conditions enable us to develop a simple procedure to test the controllability of LDS using sparse inputs, which is non-combinatorial in nature, unlike the existing sparse-controllability tests.

Following the existence test, we address the second question, namely that of devising lowcomplexity recovery algorithms. We develop online non-iterative algorithms for the same sparsity model. Motivated by the wideband wireless channel estimation problem, we assume that the control inputs are jointly sparse, and the system transfer matrix is diagonal. We devise two online algorithms based on using the sparse Bayesian learning framework. The algorithms are implemented using the sequential expectation-maximization procedure, combined with Kalman smoothing. Consequently, they require minimal computational and memory resources, and have bounded delays. Further, we rigorously examine the properties of the algorithm to answer the third question on recovery guarantees. The analysis involves elegant use of tools from the stochastic approximation theory.

Finally, we present the most sophisticated sparsity model considered in the thesis, where both the control inputs and observation matrix are assumed to be unknown. This problem is referred to as the dictionary learning problem in the CS literature. Here, we focus on algorithm development and establishing its guarantees. We adopt a Bayesian approach for recovering the observation matrix and the control inputs, and solve the resulting optimization problem using two schemes: alternating minimization procedure and Armijo line search procedure. We then give insights to recovery guarantees by characterizing the properties of the algorithm using Kurdyka-Łojasiewicz-based analysis. We also show that the algorithm is likely to converge to a sparse representation.

Apart from the above set of algorithms and theoretical results, we also apply the sparse signal recovery framework to anomaly imaging for structural health monitoring. The goal here is to recover the anomaly map of a structure using multi-sensor measurements. We develop an algorithm that exploit the inherit clustered sparsity in the map, and its performance is benchmarked against two state-of-the-art algorithms using real-world data damage measurements.

Overall, the thesis presents rigorous theoretical analysis and accurate yet low complexity

algorithms for sparse recovery problems in the context of LDS.

# Glossary

AM: Alternating Minimizationa.s.: almost surely	
ALS : Armijo Line Search	
AR : Auto-Regressive	
CoSaMP : Compressive Sampling Matching Pursuit	
CS : Compressive Sensing	
EM : Expectation Maximization	
i.i.d. : Independent and Identically Distributed	
ITH : Iterative Hard Thresholding	
KL : Kullback-Leibler	
KM-SBL : Kalman Multiple Sparse Bayesian Learning	
LASSO : Least Angle Absolute Shrinkage and Selection Opera	ntor
LHS : Left Hand Side	
LDS : Linear Dynamical Systems	
MAP : Maximum a Posteriori	
ML : Maximum Likelihood	
MSE : Mean Squared Error	
MMV : Multiple Measurement Vector	
M-SBL : Multiple Sparse Bayesian Learning	
PBH : Popov-Belevitch-Hautus	
OFDM : Orthogonal Frequency Division Multiplexing	
OMP : Orthogonal Matching Pursuit	
RHS : Right Hand Side	
RMSE : Relative Mean Square Error	
RIC : Restricted Isometry Constant	
RIP : Restricted Isometry Property	
SBL : Sparse Bayesian Learning	
SNR : Signal-to-Noise Ratio	
SRR : Support Recovery Rate	

# Notation

Vectors and matrices are denoted by boldface small letters and boldface capital letters, respectively. Sets are denoted by calligraphy letters. Rest of the notation is listed below.

## Vector

$oldsymbol{a}_i$	:	$i^{th}$ element of vector $\boldsymbol{a}$
$\ \cdot\ $	:	Euclidean norm of a vector
$\left\ \cdot\right\ _{0}$	:	Number of nonzero entries of a vector
$\left\ \cdot\right\ _{1}$	:	$\ell_1$ -norm of a vector
$\ \cdot\ _{\infty}$	:	Infinity norm of a vector
$Diag\left\{\cdot\right\}$	:	Diagonal matrix of with entries of a vector on the diagonal
$Supp\left\{\cdot\right\}$	:	Support set of a vector
$ \begin{array}{c} \left\ \cdot\right\ _{1} \\ \left\ \cdot\right\ _{\infty} \\ Diag\left\{\cdot\right\} \end{array} $	:	Infinity norm of a vector Diagonal matrix of with entries of a vector on the diagonal

## Matrix

1120001111		
$oldsymbol{A}_{ij}$	:	$(i,j)^{th}$ entry of matrix $\boldsymbol{A}$
		$i^{th}$ column of matrix $\boldsymbol{A}$
$\left( oldsymbol{A}^{T}  ight)_{i}$	:	$i^{th}$ row of matrix $oldsymbol{A}$
		Set of columns of matrix $\boldsymbol{A}$ indexed by the set $\boldsymbol{\mathcal{S}}$
$(\cdot)^{T}$	:	Transpose of a matrix
$ \cdot $	:	Determinant of a matrix
$(\cdot)^{\dagger}$	:	Pseudo-inverse of a matrix
$Tr\left\{ \cdot  ight\}$	:	Trace of a matrix
$Rank\left\{\cdot\right\}$	:	Rank of a matrix
$\left\ \cdot\right\ _{F}$	:	Frobenius norm of a matrix
$\left\ \cdot\right\ _{2}$	:	Spectral norm of a matrix
$\mathcal{D}\left\{ \cdot ight\}$	:	Diagonal matrix with same diagonal entries as the argument matrix
$vec\left\{\cdot\right\}$	:	Vectorized version of a matrix
$\mathcal{CS}\left\{ \cdot  ight\}$	:	Column space of a matrix
$\odot$	:	Khatri-Rao product of matrices

 $\mathbb{1}_{\{\cdot\}}$ 

: Indicator function

# Contents

A	cknowledgments				
A	bstra	.ct		iii	
$\mathbf{G}$	Glossary				
N	otati	on		vii	
1	Intr	oducti	on	1	
	1.1	Review	v of Compressed Sensing	3	
		1.1.1	Sparse Recovery Algorithms	3	
		1.1.2	Recovery Guarantees	4	
		1.1.3	Extensions	6	
	1.2	Sparsi	ty Models in LDS	7	
		1.2.1	SM1: Known inputs and observation matrix	7	
		1.2.2	SM2: Unknown inputs and known observation matrix $\ldots$ $\ldots$ $\ldots$	8	
		1.2.3	SM3: Unknown inputs and observation matrix	8	
	1.3	Scope	of the Thesis	9	
	1.4	Thesis	Outline and Contributions	12	
		1.4.1	Measure bounds for observability	12	
		1.4.2	Sparse-controllability	12	
		1.4.3	Online Bayesian algorithms for the recovery of state vectors	13	
		1.4.4	Bayesian algorithms for learning state and system observation matrix	14	
		1.4.5	Anomaly imaging for structural health monitoring	15	
	1.5	List of	Publications from this Thesis	17	

## CONTENTS

<b>2</b>	Obs	ervabi	lity of Sparse Initial State	19			
	2.1	Backg	round	19			
		2.1.1	Related Work	21			
	2.2	System	n Model	23			
	2.3	Prelim	linaries	26			
	2.4	RIP F	or Independent Observation Matrices	27			
		2.4.1	Special Cases	30			
		2.4.2	Number of Measurements	31			
		2.4.3	RIP of the Product of Matrices	32			
		2.4.4	Extension to Robust Recovery	33			
		2.4.5	Comparison With Prior Work	36			
		2.4.6	Extension to Identical Observation Matrices Case $\ . \ . \ . \ . \ .$	37			
	2.5	RIP F	or Identical Observation Matrices Case	38			
		2.5.1	Special Cases	39			
		2.5.2	Number of Measurements	39			
	2.6	Joint 1	Recovery of Sparse Initial State and Sparse Inputs	43			
	2.7	Summ	ary	47			
3	Spa	Sparse-Controllability 48					
	3.1	Backg	round	49			
		3.1.1	Related Literature	49			
	3.2	System	n Model	53			
	3.3	Necess	sary and Sufficient Conditions for Sparse-Controllability	54			
		3.3.1	Verification Procedure	57			
		3.3.2	Comparison with the Kalman-type Rank Test	59			
		3.3.3	Inputs with Common Support	60			
		3.3.4	Illustrative Examples	61			
	3.4	Minim	num Number of Control Input Vectors	63			
	3.5	Decomposing Sparse-controllable States					
	3.6	Outpu	t Sparse-Controllability	71			
		3.6.1	Necessary and Sufficient Conditions for Output Sparse-Controllability	72			
		3.6.2	Minimum Number of Input Vectors for Output Controllability	76			

X

## CONTENTS

		3.6.3 Change of Basis Property	77
	3.7	Summary	78
4	Bay	vesian Recovery Algorithms for Jointly Sparse Control Inputs	79
	4.1	Background	79
	4.2	Problem Formulation	82
		4.2.1 Estimation Objectives	83
		4.2.2 Offline KM-SBL Algorithm	85
	4.3	Iterative Online Algorithm Development	87
		4.3.1 Fixed Lag KSBL	87
		4.3.2 Sawtooth Lag KSBL	89
		4.3.3 Complexity Analysis	91
	4.4	Non-iterative Online Algorithm Development	92
		4.4.1 Implementation of the Algorithm	93
		4.4.2 Discussion	98
		4.4.3 Refinements	99
		4.4.4 Complexity Analysis	100
	4.5	Numerical Results: Non-iterative Algorithm	101
		4.5.1 Comparison with Existing Algorithms	108
		4.5.2 Sparse OFDM Channel Estimation	110
	4.6	Summary	113
<b>5</b>	Cor	nvergence Analysis of Online M-SBL Algorithm	114
	5.1	Uncorrelated Case	115
	5.2	Perfectly Correlated Case	121
	5.3	Simulation Results	123
		5.3.1 Convergence	124
	5.4	Summary	126
6	Bay	vesian Learning Algorithm for Sparse Control Inputs and Observation	
	Mat	trix	127
	6.1	Background	128
	6.2	SBL Based Dictionary Learning	131

		6.2.1 Alternating Minimization (AM)	3
		6.2.2 Armijo Line Search (ALS) $\ldots \ldots 135$	5
		6.2.3 Comparison of the two optimization procedures	3
		6.2.4 Comparison with other Bayesian techniques	)
	6.3	Convergence Analysis	1
		6.3.1 AM Procedure	2
		6.3.2 ALS Procedure	2
	6.4	DL-SBL Cost: Analysis of Minima	3
	6.5	Simulation Results	3
		6.5.1 Convergence	)
		6.5.2 Performance of the Algorithms	1
	6.6	Summary	3
7	Anc	maly Imaging for Structural Health Monitoring 157	7
	7.1	Background	3
	7.2	System Model	)
	7.3	Map Recovery Algorithm	1
	7.4	Experimental Results	3
	7.5	Summary	5
8	Con	clusions 168	3
	8.1	Summary of Contributions	3
		8.1.1 SM1: Known inputs and observation matrix	)
		8.1.2 SM2: Unknown inputs and Known observation matrix	)
		8.1.3 SM3: Unknown inputs and observation matrix	2
		8.1.4 Anomaly Imaging Exploiting Clustered Sparsity	3
	8.2	Future Work	1
$\mathbf{A}$	App	bendix to Chapter 2 176	3
	A.1	Proof of Proposition 2.1	3
	A.2	Proof of Theorem 2.1	7
	A.3	Proof of Theorem 2.3	3
		A.3.1 Toolbox	3

		A.3.2 Proof	.81
	A.4	Proof of Lemma A.3	.85
	A.5	Proof of Lemma A.4	87
	A.6	Proof of Lemma A.5	.88
	A.7	Proof of Lemma A.6	.89
	A.8	Proof of Proposition 2.2	91
	A.9	Proof of Proposition 2.3	.92
в	App	pendix to Chapter 3 19	94
	B.1	Proof of Theorem 3.1	.94
	B.2	Proof of Corollary 3.2	.96
	B.3	Proof of Theorem 3.3	.97
		B.3.1 Characterizing $\mathcal{H}^*_{(K)}$	.98
		B.3.2 Characterizing $R^*_{(K)}$	.99
		B.3.3 First part of the upper bound	202
		B.3.4 Upper bounding $K^*$	203
		B.3.5 Lower bounding $K^*$	205
$\mathbf{C}$	App		205 <b>06</b>
С	<b>Арр</b> С.1		06
С		pendix to Chapter 5 20	<b>06</b> 206
С	C.1	pendix to Chapter 5       20         Proof of Proposition 5.1       2	<b>06</b> 206 210
С	C.1 C.2 C.3	pendix to Chapter 5       20         Proof of Proposition 5.1       2         Proof of Theorem 5.1       2         Proof of Proposition 5.2       2	<b>06</b> 206 210 215
C	C.1 C.2 C.3 C.4	Dendix to Chapter 5       20         Proof of Proposition 5.1       2         Proof of Theorem 5.1       2         Proof of Proposition 5.2       2         Proof of Theorem 5.3       2	<b>06</b> 206 210 215
	C.1 C.2 C.3 C.4 App	Dendix to Chapter 520Proof of Proposition 5.12Proof of Theorem 5.12Proof of Proposition 5.22Proof of Theorem 5.32Dendix to Chapter 62	<b>06</b> 206 210 215 216
	C.1 C.2 C.3 C.4 App	Dendix to Chapter 520Proof of Proposition 5.12Proof of Theorem 5.12Proof of Proposition 5.22Proof of Theorem 5.32Proof of Theorem 5.32Dendix to Chapter 62Proof of Proposition 6.12	<ul> <li>06</li> <li>206</li> <li>210</li> <li>215</li> <li>216</li> <li>19</li> </ul>
	C.1 C.2 C.3 C.4 App D.1 D.2	Dendix to Chapter 520Proof of Proposition 5.12Proof of Theorem 5.12Proof of Proposition 5.22Proof of Theorem 5.32Proof of Theorem 5.32Proof of Proposition 6.12Proof of Theorem 6.12	<ul> <li>06</li> <li>206</li> <li>210</li> <li>215</li> <li>216</li> <li>19</li> <li>219</li> </ul>
	C.1 C.2 C.3 C.4 App D.1 D.2	Dendix to Chapter 524Proof of Proposition 5.12Proof of Theorem 5.12Proof of Proposition 5.22Proof of Theorem 5.32Proof of Theorem 5.32Proof of Proposition 6.12Proof of Theorem 6.12Proof of Theorem 6.22	<ul> <li>06</li> <li>206</li> <li>210</li> <li>215</li> <li>216</li> <li>19</li> <li>220</li> </ul>
	C.1 C.2 C.3 C.4 App D.1 D.2	Dendix to Chapter 5       20         Proof of Proposition 5.1       2         Proof of Theorem 5.1       2         Proof of Proposition 5.2       2         Proof of Theorem 5.3       2         Proof of Proposition 6.1       2         Proof of Theorem 6.1       2         Proof of Theorem 6.2       2         D.3.1       Proof of Theorem 6.2       2	<ul> <li>06</li> <li>206</li> <li>210</li> <li>215</li> <li>216</li> <li>19</li> <li>220</li> <li>222</li> </ul>
	C.1 C.2 C.3 C.4 <b>App</b> D.1 D.2 D.3	Dendix to Chapter 520Proof of Proposition 5.12Proof of Theorem 5.12Proof of Proposition 5.22Proof of Theorem 5.32Proof of Theorem 5.32Proof of Proposition 6.12Proof of Theorem 6.12Proof of Theorem 6.22Proof of Theorem 6.22Proof of Proposition 6.22	<ul> <li>06</li> <li>206</li> <li>210</li> <li>215</li> <li>216</li> <li>19</li> <li>220</li> <li>222</li> <li>225</li> </ul>
	C.1 C.2 C.3 C.4 <b>App</b> D.1 D.2 D.3 D.4	pendix to Chapter 5       24         Proof of Proposition 5.1       2         Proof of Theorem 5.1       2         Proof of Proposition 5.2       2         Proof of Theorem 5.3       2         Proof of Proposition 6.1       2         Proof of Theorem 6.1       2         Proof of Theorem 6.2       2         Proof of Proposition 6.2       2         Proof of Theorem 6.3       2	<ul> <li>06</li> <li>206</li> <li>210</li> <li>215</li> <li>216</li> <li>19</li> <li>220</li> <li>222</li> <li>225</li> <li>228</li> </ul>

D.8 Proof of Theorem 6.5	 233
Bibliography	233

# List of Figures

2.1	Variation of $K/L^{2}(\mathbf{D}, K)$ with K when $\mathbf{D}$ generated using different distri-	
	butions. We see that $K/L^{2}(\boldsymbol{D},K)$ is a (linearly) non-decreasing function	
	of <i>K</i>	40
2.2	Variation of $K/L^{2}(\mathbf{D}, K)$ with K for Fourier, Hadamard and identity con-	
	structions of <b>D</b> . We see that $K/L^{2}(\mathbf{D}, K)$ is not an increasing function of	
	<i>K</i>	41
4.1	The sawtooth lag processing scheme	95
4.2	Performance of our algorithms relative to the offline algorithm for $D$ =	
	${\bf 0}$ (uncorrelated case, where we use the M-SBL based algorithm). Other	
	paramters are $\Delta = 5$ and SNR = 20 dB	102
4.3	Performance of our algorithms relative to the offline algorithm for $oldsymbol{D}$ =	
	${\bf 0}$ (uncorrelated case, where we use the M-SBL based algorithm). Other	
	paramters are $\Delta = 5$ and $K = 120$	103
4.4	Performance of our algorithms relative to the offline algorithm for $D$ =	
	${\bf 0}$ (uncorrelated case, where we use the M-SBL based algorithm). Other	
	paramters are $K = 150$ , $\Delta = 5$ and SNR = 20 dB	104
4.5	Performance of our algorithms relative to the offline algorithm for $\boldsymbol{D}=0.9\boldsymbol{I}$	
	(correlated case, where we use the KM-SBL algorithm). Other paramters	
	are $\Delta = 5$ and SNR = 20 dB.	105
4.6	Performance of our algorithms relative to the offline algorithm for $\boldsymbol{D}=\rho \boldsymbol{I}$	
	(correlated case, where we use the KM-SBL algorithm). Other paramters	
	are $K = 150$ , $\Delta = 5$ and SNR = 20 dB	106

## LIST OF FIGURES

4.7	Comparison of RMSE, SRR and run time of our algorithm with the existing online schemes.	109
4.8	Comparison of the BER, RMSE and run time of our algorithm with existing	105
	schemes	112
5.1	Convergence of the hyperparameters to the true value	123
6.1	Convergence of ALS procedure ((a), (b)) and comparison with AM (c), with $K = 1000, m = 20, N = 60, s = 6$ , and SNR = 20 dB, for the first iteration of EM algorithm.	149
6.2	Comparison of DL-SBL with KSVD, MOD and Gaussian hierarchical model based SBL algorithms when the number of input vectors is varied. The performance of DL SPL is superior to the other three algorithms	159
6.3	performance of DL-SBL is superior to the other three algorithms An example of image denoising using different dictionary learning algorithms with $K = 1000$ , $m = 64$ , and $N = 256$ . We see that all algorithms	102
	improve the visual quality of the noisy image	153
7.1	The figure shows $i^{\text{th}}$ sensor-actuator pair and the direct path between them. The pixels in blue correspond to the nonzero entries of $j^{\text{th}}$ row of $L$ , and the non-zero value equals the length of the path overlapping the pixel	160
7.2	Comparison of the damage outlines estimated by three different algorithms along with corresponding Sørenson-Dice similary index. The method of this	
	chapter provides the best results out of the three methods	166

# List of Tables

4.1	Comparison of Online schemes against Offline scheme when $K$ observations	
	are available	92
4.2	Comparison of the online schemes with the offline scheme when $K$ observa-	
	tions are available.	100
4.3	Simulation parameters for OFDM channel estimation	111
5.1	Value of error-fit power function parameter $p$ when $D = 0$	124
6.1	Comparison of ALS convergence behavior with varying step size parameters	
	$\beta$ and $\alpha$	150
6.2	Comparison of ALS and AM convergence behavior	150

# Chapter 1

## Introduction

#### Connecting sparsity and state space model

Linear dynamical systems (LDS) are well studied and widely accepted mathematical models for describing and analyzing a variety of physical systems that evolve in time. The model serves as core engine in diverse fields such as automatic control systems [1], signal processing [2], communications [3], economics [4], mechanical and civil engineering [5,6], health care [7,8], etc. The canonical representation of the linear dynamics is the discrete state space model as given below:

$$\boldsymbol{x}_k = \boldsymbol{D}\boldsymbol{x}_{k-1} + \boldsymbol{H}\boldsymbol{h}_k \tag{1.1}$$

$$\boldsymbol{y}_k = \boldsymbol{A}_{(k)} \boldsymbol{x}_k + \boldsymbol{w}_k, \qquad (1.2)$$

for time instants  $k = 0, 1, \ldots$ . Here,  $\boldsymbol{x}_k \in \mathbb{R}^N$  denote the the state vector that characterizes the system. The state is influenced by the input  $\boldsymbol{h}_k \in \mathbb{R}^L$ , and the output  $\boldsymbol{y}_k \in \mathbb{R}^m$ represents the measurable response of the system. The output is corrupted by the noise term denoted by  $\boldsymbol{w}_k \in \mathbb{R}^m$ . Also,  $\boldsymbol{D} \in \mathbb{R}^{N \times N}$  is the system transfer matrix,  $\boldsymbol{H} \in \mathbb{R}^{L \times N}$  is the input matrix, and  $\mathbf{A}_{(k)} \in \mathbb{R}^{m \times N}$  is the observation matrix of the system at time instant k. For example, in an orthogonal frequency division multiplexing (OFDM) wireless system, the state denotes the successive instantiations of a time-varying wireless channel. The temporal correlation of the channel modeled using a first-order auto-regressive process is captured by (1.1). Also, (1.2) denotes the linear relation between the received signal, pilot signal, and the channel instantiation.

An important problem associated with a linear dynamical system is the estimation of the system state vectors  $\boldsymbol{x}_k$ , for  $k = 0, 1, \ldots$  using the corresponding measurements  $\boldsymbol{y}_k$ . This problem is equivalent to solving a system of linear equations. In the context of wireless channel explained above, this problem translates to the classical channel estimation problem. In general, for solving a linear system, the number of measurements should not be less than the number of unknowns. However, if the solution is known to admit a sparse representation in a suitable basis, the number of measurements required can be potentially reduced by exploiting this additional information. The reduction in the number of measurements is advantageous in most of the real world systems. For instance, in wireless channel estimation, the channel vector is known to be sparse in the lag domain, and thus exploiting sparsity help to reduce the pilot overhead. Hence, the whole thesis is devoted to the mathematics underlying the state estimation of LDS when the state vectors admit sparse representation. Our work is motivated by the results from the area of sparse signal recovery or compressive sensing (CS). These results serve as the point of departure for our work, and we discuss them in the next section.

## 1.1 Review of Compressed Sensing

The CS framework studies the theory and algorithmic development for finding sparse solutions to underdetermined linear systems of equations [9–11]. The standard CS problem is to reconstruct a s-sparse vector  $\boldsymbol{x} \in \mathbb{R}^N$  from linear measurements:

$$\boldsymbol{y} = \boldsymbol{A}\boldsymbol{x} \in \mathbb{R}^m, m < N. \tag{1.3}$$

There are two aspects to the CS problem: one, design of linear measurement process, and second, design of recovery algorithm. These two problems are equally important and delicate, and they form the foundation of the thesis. Thus, in the next subsection, we discuss the popular and well-understood algorithms and the associated recovery guarantees available in the literature.

## 1.1.1 Sparse Recovery Algorithms

Some of the existing algorithms for the recovery of a sparse vector are as follows:

• *Basis pursuit*: It is a convex optimization method with polynomial time complexity [9, 12]. It solves the following optimization problem:

$$\arg\min_{\boldsymbol{x}\in\mathbb{R}^N} \|\boldsymbol{x}\|_1$$
 subject to  $\boldsymbol{y} = \boldsymbol{A}\boldsymbol{x}$ . (1.4)

There are other variants of the algorithm like LASSO, Dantzig selector,  $\ell_p$  norm minimization, etc [13, 14].

• *Thresholding algorithms*: These are iterative algorithms based on carefully designed thresholding functions. Basic thresholding, iterative hard thresholding (IHT) and

hard thresholding pursuit (HTP) are some of the algorithms that belong to this class of algorithms [15].

- *Greedy algorithms*: These algorithms are designed based on making a locally optimal choice of the support of the vector at each stage. Orthogonal matching pursuit (OMP) and compressive sampling matching pursuit (CoSaMP) are examples of greedy algorithms [16–21].
- Bayesian methods: These algorithms impose a fictitious Bayesian prior on the unknown vector and solve the sparse recovery problem using probabilistic estimation.
   Sparse Bayesian learning (SBL) and approximate message passing are some wellreceived Bayesian methods in literature [22–26].

Clearly, these algorithms are not fitted for arbitrary measurement matrices. For example, if the number of rows m of the measurement matrix is less than the sparsity s, the recovery is information theoretically impossible. The desired properties of the measurement matrices are discussed next.

## 1.1.2 Recovery Guarantees

One can look for two types of guarantees for a CS problem:

- *Existence and uniqueness:* When is a solution to a compressed sensing problem is guaranteed to exists?
- *Recovery guarantee:* When is the solution recovered by the compressed sensing algorithm guaranteed to be unique?

The first question is relatively easy to answer, as the solution is guaranteed to exist if the column space spanned by the any of the *s* columns of  $\mathbf{A}$  is  $\mathbb{R}^m$ . Further, the solution is unique if the null space of  $\mathbf{A}$  does not contain any 2s- sparse vector. However, the second question is by far not trivial and has received a lot of research attention. The most popular approach to establishing guarantees for the exact recovery is through the so-called restricted isometry property (RIP) [27,28], defined as follows. An measurement matrix  $\mathbf{A}$ is said to satisfy the *s*-RIP with restricted isometry constant (RIC)  $\delta_s$  if  $\delta_s \in (0, 1)$ , where

$$\delta_s \triangleq \inf \left\{ \delta : 1 - \delta \le \|\boldsymbol{A}\boldsymbol{z}\|^2 \le 1 + \delta, \ \forall \|\boldsymbol{z}\| = 1, \text{ and } \|\boldsymbol{z}\|_0 \le s \right\}.$$
(1.5)

Some examples of RIP based guarantees for exact recovery of sparse vectors with  $\ell_0$  norm at most *s* are as follows:

- $\delta_s < 1/3$  and  $\delta_{2s} < \sqrt{1/2}$ , (more generally,  $\delta_{ts} < \sqrt{(t-1)/t}$  for  $t \ge 4/3$  and  $\delta_{ts} < \sqrt{t/(4-t)}$  for 0 < t < 4/3) are sharp for recovery using basis pursuit [29–31]
- $\delta_{3s} < 1/8$  is sufficient for recovery via the iterative hard thresholding (IHT) algorithm [32]
- $\delta_{s+1} < \frac{1}{\sqrt{s+1}}$  is sufficient for recovery via the orthogonal matching pursuit (OMP) algorithm [33]

The RIP also ensures that the recovery process is robust to noise and is stable when the unknown vector is not precisely sparse.

#### 1.1.3 Extensions

The theory and algorithms we discussed so far are appropriate for the recovery of a single sparse vector with no other constraint. However, in practice, there are LDS related problems for which one has to recover more than one sparse vector and the sparse vectors exhibit additional structural properties. For example, in the wireless channel estimation problem, one has to recover successive instantiations of a sparse time-varying wireless channel. These instantiations have the same power delay profile, and the nonzero coefficients of these instantiations are temporally correlated. Thus, a recovery algorithm exploiting the common support and temporal correlation yield better recovery performance. We list the different types of sparsity models studied in the literature:

- *Block sparsity:* Sparsity pattern in which the non-zero entries occur in multiple clusters [34,35].
- *Piecewise sparsity:* Sparsity pattern formed by the concatenation of a set of sparse vectors [36].
- Joint sparsity: Sparsity pattern in which a set of sparse vectors share the same support [37–41].
- *Temporally correlated joint sparsity:* A model in which vectors exhibit joint sparsity along with temporal correlation of the nonzero entries [3, 42, 43].

To summarize this section, for a given sparse recovery problem, the first step is to identify any structure other than sparsity that exists in the signal. Once the structure is revealed, one can explore two facets of the problem: development of efficient recovery algorithms and theoretical guarantees on existence and uniqueness of the solution, and recovery performance. Hence, in this thesis, we build on the theory of sparse signal recovery arise in LDS, which deals with the above aspects.

With the above background, we next introduce the sparsity models associated with LDS that are investigated in this thesis.

## 1.2 Sparsity Models in LDS

We consider three versions (SM1-SM3) of spare recovery problems for the *state estimation* of LDS. The categorization is based on knowledge of recovery algorithm about the inputs and the observation matrix, as we describe next. Here, we highlight a selection of LDS problems that can be reduced to or can be modeled using the CS framework, without detailing the mathematical model.

#### 1.2.1 SM1: Known inputs and observation matrix

In version 1(SM1) of the problem where the inputs and observation matrix are known, the estimation of states implies the estimation of the initial sparse vector. Thus, this version is concerned with the question of how well the initial state of a linear dynamical system can be inferred from its observations and inputs. Here, we assume that the initial state of the system is known to admit a sparse representation in a suitable basis. For example, diffusion processes in complex networks that model phenomena like disease or epidemic spreading in human society [7, 8], air or water pollution [5, 6], virus spreading in computer and mobile phone networks [44, 45], information propagation in online social networks [46], etc., are known to have a sparse initialization. Identifying the initial state of these processes accurately is a critical first step towards their control [47]. Thus, a key problem here is the recoverability of the sparse system state which simplifies to a single measurement model as in (1.3).

#### 1.2.2 SM2: Unknown inputs and known observation matrix

In the next model, version 2 (SM2), the goal is to estimate inputs and the initial state which is equivalent to the estimation of all the states. Thus, this version refers to the estimation of a sequence of vectors, which are assumed to be sparse. In other words, we assume that the initial state and the inputs are sparse. A motivating application for sparse control is a networked control system. The system comprises of controllers, plants and sensors, connected over a network medium. Due to the limited bandwidth of the physical communication network, the communication in the network only support low data rates [48, 49]. Thus, to reduce the size of data exchanged between controllers and plants, one can use sparse signals as control inputs. This is because the sparse signals are known to admit compact representations [9–11]. Another motivating real-world problem that can be modeled using sparse inputs is the wireless channel estimation problem described at the beginning of this chapter. For this problem, the inputs refer to the difference between the consecutive instantiations of the channel, and the goal is to recover the sequence of sparse channel instantiations.

#### **1.2.3** SM3: Unknown inputs and observation matrix

In this level of sparse recovery problem, version 3(SM3), one needs to learn both the matrix that characterizes LDS and the external inputs that influence the state. In the CS setup, this problem is known as *dictionary learning*. Learning the system specific adaptive

measurement matrices are particularly beneficial when the measurement model is not precisely known, as in the case of an image. The use of adaptive dictionaries often leads to more compact representations and better performance in signal denoising, inpainting, and restoration. This method is known to outperform the traditional way of using predefined dictionaries like wavelets or union of orthogonal bases like Fourier and Dirac.

In a nutshell, the above three sparsity patterns arise in LDS are highly motivated by different real-world applications. Therefore, it is incentive to explore the theoretical and algorithmic aspects of the aforementioned sparsity models. In the next section, we sketch the territory of research presented in this thesis.

## 1.3 Scope of the Thesis

The central research questions addressed in this thesis are as follows:

- **Q1.** When is a sparse solution is guaranteed to exist for a given sparsity model?
- **Q2.** If the solution exists, what are some efficient reconstruction algorithms?
- **Q3.** For a given reconstruction algorithm, when is the solution guaranteed to be unique?

These questions are not entirely independent, as the recovery guarantees are algorithm dependent, and the reconstruction algorithm assumes the existence of at least one sparse solution. The thesis addresses these fundamental questions for the three versions of sparsity patterns presented in Section 1.2.

Next, we refine these questions for each case separately. We do not delve into the technical details that would be necessary for total comprehension. Instead, we adopt an informal style and we focus on giving an overview of CS problems that arise in the LDS framework.

As previously revealed, we first mention about the special structures associated with the measurement matrices or the sparse signals. This helps us to connect the LDS problems with the existing CS literature to recognize the literature gaps.

### Known inputs and observation matrix (SM1)

As we mentioned in the previous section, this version of the problem reduces to a standard CS problem discussed in Section 1.1. Also, a special structure arises in the measurement matrix due to the LDS model, which makes the problem different from the canonical CS model. Since we do not assume any special structure on the initial state vector as mentioned in Section 1.1.3, and hence, all the conventional recovery algorithms can be applied to this version of the problem. However, the theoretical guarantees which depend on the properties of the measurement model require a fresh look. Therefore, the key questions to be answered in this context are Q1 and Q3. Although Q1 which deals with the existence of the solution is an intriguing mathematical problem, from the application point of view, it is not of much interest. In contrast, Q3 which refers to how well the state of a LDS can be inferred from its observations and inputs is a well-studied topic for a general nonsparse system. This problem is called *observability problem* according to the control theory terminology. To sum up, under the model SM1, we look at only question Q3 which establishes the recovery guarantee of the measurement matrix that arises in the observability problem of an LDS system.

### Unknown inputs and known observation matrix (SM2)

For this version of the sparsity model, the goal is to estimate a set of sparse vectors, which makes this mathematical model extremely different from SM1. Also, unlike SM1, the existence of a solution is an important problem for this version of the problem. In control theory literature, the existence of a set of control inputs to drive the system to any desired state is referred to as the *controllability* problem. So, revisiting the controllability problem taking the sparsity constraints into account is a vital problem to be solved. Further, designing sparse recovery algorithms tailored for LDS system is also a critical problem. An algorithm with good recovery guarantees and low complexity for the recovery of multiple sparse vectors, exploiting all the structures associated with the LDS is the requirement in this setting. In short, we address questions Q1-Q3 under this version of the problem.

### Unknown inputs and observation matrix (SM3)

In this version, the problem is not equivalent to solving a system of linear equations with some constraints or signal structure. Instead, we need to find a decomposition that can explain our measurements and ensuring the control inputs are sparse. Thus, the theoretical analysis demanded by question Q1 is often hard to carry out for this version of the problem. Hence, we restrict our work to the algorithm design and the related recovery guarantees, i.e., we only seek answers to questions Q2 and Q3, for SM3.

To brief, we develop rigorous recovery results for the described three levels of sparsity models that emerge in the LDS framework, answering some fundamental questions on existence, recoverability, and uniqueness of the solution. We next give a road map of the primary contributions of this thesis.

## **1.4** Thesis Outline and Contributions

The overall organization follows a path from simple to more complicated versions of CS problems. In the following, the major finding in the thesis is listed chapterwise.

### 1.4.1 Measure bounds for observability

Chapter 2 addresses question Q3 for sparsity model SM1: the recovery guarantees in the context of the observability problem of an LDS. We show that, if the initial state vector admits a sparse representation, the number of measurements can be significantly reduced by using random projections for obtaining the measurements. Our analysis gives sufficient conditions for the RIP of the observability matrix to hold, which leads to guarantees for the observability of the system. Our results depend only on the properties of system transfer and observation matrices and are derived using tools from probability theory and compressed sensing. Our results are stronger than the existing results in the regime where they are comparable. These results appear in [50, 51].

## 1.4.2 Sparse-controllability

Chapter 3 answers question Q1 for sparsity model SM2: the fundamental problem of controllability of an LDS with sparse control inputs. We derive necessary and sufficient conditions for ensuring controllability of an LDS with arbitrary transfer matrices. Our characterizations are in terms of algebraic conditions, which require verifying rank conditions on an appropriately defined set of matrices. The number of conditions to be verified grows with the state dimension, but does not depend on the number of input vectors required to steer the system to the desired state. In this way, the results have a similar flavor as the classical results for unconstrained input system, although the proof technique is completely different. Using these conditions, we designed a non-combinatorial test to check the controllability of LDS using sparse inputs. Further, we characterize the minimum number of input vectors required to satisfy the derived conditions for controllability. Finally, we present a generalized Kalman decomposition-like procedure that separates the state-space into subspaces corresponding to sparse- controllable and sparse-uncontrollable parts. Our results form a theoretical foundation for designing networked linear control systems with sparse inputs, by introducing and investigating the notion of controllability under sparsity constraints. These results appear in [52].

## 1.4.3 Online Bayesian algorithms for the recovery of state vectors

In Chapter 4, we address the question Q2 on the recovery of sparse state vectors of LDS for model SM2. For this work, we assume that the system transfer matrix is a diagonal matrix. Hence, the problem is to reconstruct temporally sparse vectors sharing a common support, from noisy underdetermined linear measurements. We device two Bayesian algorithms that sequentially recover the vectors, without waiting for the entire measurements to arrive. The online algorithms are formulated using the sparse Bayesian learning(SBL) framework and are implemented using a sequential expectation-maximization procedure combined with Kalman smoothing. The first set of algorithms is iterative in nature, which is then modified to develop a noniterative algorithm. Due to the online nature of the algorithm, it requires less computational and memory resources compared to offline processing. Thus, our algorithms are of significant value in practical settings. We illustrate the efficacy of the algorithms using sparse orthogonal frequency division multiplexing channel estimation through numerical results. These results are published in [53, 54]

## Convergence analysis of online Kalman multiple sparse Bayesian algorithms

In Chapter 5, we present the solution to question Q3 for model SM2, in the context of algorithms presented in Chapter 4. We analyze the convergence of the algorithms in the special case when the sparse vectors are uncorrelated, using tools from stochastic approximation theory. We show that the sequence of the covariance estimates converges either to the global minimum of the offline equivalent cost function or to all-zero vector, regardless of the sparsity level of the signal. These results appear in [54]

## 1.4.4 Bayesian algorithms for learning state and system observation matrix

Chapter 6 focuses on Q2 and Q3 for model SM3: to learn a dictionary from a finite set of noisy training signals, such that the training data admits a sparse representation over the dictionary. For this work, we assume that the system transfer matrix is a zero matrix. The joint recovery of the sparse representation and dictionary is formulated using the sparse Bayesian learning framework by imposing a fictitious prior on the sparse vectors. The parameters of the prior on the sparse vectors and the dictionary are simultaneously learned using the expectation-maximization algorithm. The dictionary update step turns out to be a nonconvex problem which is solved using either an alternating minimization (AM) procedure or the Armijo line search (ALS). Next, to address Q3, we show that the algorithm converges, and further analyze the stability of the solution by characterizing its limit points. We also analyze the minima of the overall cost function of the presented algorithm and prove that the desired sparse representation is likely to achieve by the algorithm. Through numerical results, we demonstrate the efficacy of the presented algorithm and compare it with existing dictionary learning algorithms for the application of image denoising. These results appear in [55].

## 1.4.5 Anomaly imaging for structural health monitoring

In Chapter 7, we include a different application of structured sparse signal recovery, namely, anomaly imaging for structural health monitoring. Although this chapter does not discuss an LDS-based sparsity model, the ideas of the chapter is aligned with the main theme of sparsity. This chapter presents a new tomography-based anomaly mapping algorithm for composite structures. The system consists of an array of piezoelectric transducers which sequentially excites the structure and collects the resulting waveform at the remaining transducers. Anomaly indices computed from the sensor waveforms are fed as input to the mapping algorithm. The output of the algorithm is a color map indicating the outline of damage on the structure when present. Unlike prior work on this topic, the algorithm of this chapter explicitly accounts for both sparsity and cluster pattern structures that are typical of structural anomalies. Hence, the algorithm of this paper provides excellent reconstruction accuracy by incorporating the available prior information on the anomaly map. Experimental results on a unidirectional composite plate confirm that the algorithm outperforms two competing methods in terms of reconstruction accuracy. These results appear in [56].

We offer some concluding remarks and questions for further study in Chapter 8. The appendices containing supplementary material, Appendix A for Chapter 2, Appendix B for Chapter 3, and Appendix C for Chapter 5 are included at the end of the thesis.

16

On the whole, the thesis presents three different sparsity models which are of practical relevance. We cover solution guarantees, and a variety of sparse recovery algorithms, together with their theoretical analysis.

## **1.5** List of Publications from this Thesis

#### Journal Articles

- J1 G. Joseph, and C. R. Murthy, A noniterative online Bayesian algorithm for the recovery of temporally correlated sparse vectors, *IEEE Transactions on Signal Processing*, vol. 65, no. 20, pp. 5510–5525, Oct. 2017.
- J2 G. Joseph, and C. R. Murthy, On the observability of a linear system with a sparse initial state, *IEEE Signal Processing Letters*, vol. 25, no. 7, pp. 994998, Jul. 2018.
- J3 G. Joseph, and C. R. Murthy, Measurement bounds for observability of linear dynamical systems under sparsity constraints, *IEEE Transactions on Signal Processing*, vol. 67, no. 8, pp. 1992–2006, Apr. 2019.
- J4 G. Joseph, and C. R. Murthy, Sparse-controllability of linear dynamical systems, submitted to IEEE Transactions on Automatic Control.
- J5 G. Joseph, and C. R. Murthy, A Bayesian algorithm for joint dictionary learning and sparse signal recovery, *submitted to IEEE Transactions on Signal Processing*.

### **Conference Proceedings**

- C1 G. Joseph, C. R. Murthy, R. Prasad, and B.D.Rao, Online recovery of temporally correlated sparse signals using multiple measurement vectors, *IEEE Global Communications Conference*, Dec. 2015.
- C2 G. Joseph, A. B. Zoubi, C. R. Murthy, and V. J. Mathews, Anomaly imaging for

structural health monitoring exploiting clustered sparsity, *IEEE International Con*ference on Acoustics, Speech and Signal Processing, May. 2019.

# Chapter 2

# **Observability of Sparse Initial State**

#### Answering problem Q3 for model version SM1

In this chapter, we look at the LDS sparsity model with only initial state unknown, and all the control inputs and the system matrices known (model SM1). The unknown initial state is assumed to sparse in a suitable basis. For this setting, we consider two cases: one, the observation matrices are independent random matrices, and two, they are identical to a single random matrix. We derive an upper bound on the number of measurements required for the recovery a sparse initial state using the classical compressive sensing algorithms. The results are probabilistic in nature and depends on the properties of the system matrices. Also, the bound is more general and stronger than the existing results in the regime where they are comparable.

# 2.1 Background

Observability is an important notion in control theory. It is concerned with the question of how well the state of a linear dynamical system can be inferred from its observations and inputs [57]. The classical observability problem involves solving a linear system of equations:

$$\widetilde{\boldsymbol{y}}_{(K)} = \widetilde{\boldsymbol{A}}_{(K)} \boldsymbol{x}_0, \qquad (2.1)$$

where the measurement vector  $\tilde{y}_{(K)}$  and the observability matrix  $\tilde{A}_{(K)}$  are known, and we need to estimate  $x_0$  exactly.<sup>1</sup> The standard results from linear algebra state that a discrete time system is observable if the rank of the observability matrix  $\widetilde{A}_{(K)}$  equals the system dimension [58]. This result applies to the general formulation of the problem, and hence, a large number of measurements are required to recover the initial state for systems with a high dimensional state [59–61]. However, if the initial state of the system is known to admit a sparse representation in a suitable basis, the number of measurements required can be potentially reduced by exploiting this additional information. As we mentioned in Chapter 1, diffusion processes in complex networks that model phenomena like disease or epidemic spreading in the human society [7,8], air or water pollution [5,6], virus spreading in computer and mobile phone networks [44, 45], information propagation in online social networks [46], etc., are known to have a sparse initialization. Identifying the initial state of these processes accurately is a critical first step towards their control [47]. Thus, a key problem in this context is the recoverability of the sparse system state using as few measurements as possible. Further, in some cases, the measurements are obtained as random linear projections of the system state. For example, in the problem of finding the source of pollution in a water body or in the atmosphere, measurements collected from sensors placed at spatially random locations can be mathematically modeled as random linear projections of the system state [62–65]. Hence, in this chapter, we provide guarantees

<sup>&</sup>lt;sup>1</sup>We discuss the system model in detail in Section 2.2.

on the observability of a system when the observability matrix is random and possibly rank deficient, and the initial state admits a sparse representation. We establish these guarantees by analyzing the RIP of the *structured random observability matrix* arising out of a linear dynamical system.

## 2.1.1 Related Work

Our work focuses on two aspects: first, we explore the connection between compressed sensing and observability of the state of a linear dynamical system, and second, we derive sufficient conditions for state recovery by analyzing the RIP of the observability matrix. In our case, the observability matrix is a random matrix with a special structure. Hence, the existing results from the compressed sensing literature cannot be directly applied to our problem. In this subsection, we provide review the past literature in this direction.

#### Compressed sensing and observability

The connection between the compressed sensing and linear dynamical systems is a nascent topic, and has only recently been studied in the literature. The design of control algorithms based on sparsity in the state using tools from compressed sensing is presented in [66]. However, this paper does not discuss guarantees for recoverability of the system state in the presented framework. On the other hand, [67,68] assert that a linear dynamical system is observable if the observability matrix satisfies the RIP. However, conditions under which the observability matrix satisfies RIP are not discussed.

The results in [62–64] characterize the number of measurements required for the exact recovery of the initial state in a stochastic setting. However, the results are useful only under somewhat overly restrictive conditions such as the system transfer matrix being unitary, the observation matrices being i.i.d. Gaussian, and the initial state being sparse in the canonical basis. Moreover, those results depend on the smallest singular value of the transfer matrix. As a consequence, they are not independent of scaling of the transfer matrix. Therefore, in this chapter, we derive a more general results on the observability of LDS under sparsity constraints, which are independent of the scaling of the matrices.

#### **RIP** of structured random matrices

We list a few types of structured random matrices which have been shown to satisfy the RIP in the literature:

- Subsampled bounded orthonormal systems [69,70]
- Partial random circulant matrices and partial random Toeplitz matrices [71–73]
- Block diagonal measurement matrices where each block on the main diagonal is a subgaussian random matrix [74]
- The columnwise Khatri-Rao product of two matrices [75].

As we will see, the RIP of the structured random observability matrix that arises in our problem has not been studied in the past. Hence, it requires new analysis using tools from non-asymptotic random matrix theory.

In this chapter, we first derive guarantees on the system observability under a stochastic setting when the observation matrices are i.i.d. subgaussian random matrices and the system transfer matrix is nonzero. However, in many applications, due to hardware constraints, the measurement process could involve linear projection using a single, randomly selected matrix, rather than an independent matrix for each measurement instant. Hence, it is more pertinent to derive recovery guarantees for the case when the observation matrix is fixed, but equal to an instantiation of a random matrix. We present a different, new analysis to obtain guarantees for uniform recovery of the state for the identical observation matrices cases. We also study the problem of joint recovery of the initial state and sparse input vectors. The key novelty in the results is the derivation of sufficient conditions on Kand m required for the recovery of sparse initialization and inputs. The results presented here are of independent interest, since they provide insights to the RIP and NSP of the matrices with similar structure. In summary, we show that systems that are unobservable using classical control theory can be observable under the sparsity constraints.

# 2.2 System Model

We consider discrete-time linear system which is modeled as follows:

$$\boldsymbol{x}_k = \boldsymbol{D}\boldsymbol{x}_{k-1} \tag{2.2}$$

$$\boldsymbol{y}_k = \boldsymbol{A}_{(k)} \boldsymbol{x}_k, \qquad (2.3)$$

for discrete time instants k = 0, 1, ..., K - 1. Here,  $\mathbf{D} \in \mathbb{R}^{N \times N}$  is a nonzero system transfer matrix and  $\mathbf{A}_{(k)} \in \mathbb{R}^{m \times N}, m \ll N$  is the observation matrix of the system at time instant k. We are interested in the observability of the system when the initial state is sparse. We make the following points before proceeding further:

- (a) Observability of the initial sparse state  $\boldsymbol{x}_0$  implies the observability of  $\boldsymbol{x}_k$  for all k.
- (b) In (2.2), we do not include an innovation term as we did in (1.1). Since we are considering the problem of system observability, the system input is assumed to be

known. We can therefore simply subtract its effect from the system evolution as well as observation equations, resulting in the system model given by (2.2) and (2.3). We consider the joint recovery of the initial state and sparse innovation terms in Section 2.6.

(c) The system equations do not consider measurement noise or model mismatch. However, in the presence of these impairments, our results can be extended to robust recovery of the initial state; we discuss this in Section 2.4.4.

In view of the above, we formally define the notion of observability as follows:

**Definition 2.1** (Observability). A system is said to be observable if any unknown *s*-sparse initial state  $\boldsymbol{x}_0$  can be determined uniquely from the outputs  $\{\boldsymbol{y}_k\}_{k=0}^{K-1}$ , the transfer matrix  $\boldsymbol{D}$ , and the observation matrices  $\{\boldsymbol{A}_{(k)}\}_{k=0}^{K-1}$ .

To recover the sparse initial vector, we consider the following equivalent linear system at time K:

$$\widetilde{\boldsymbol{y}}_{(K)} = \widetilde{\boldsymbol{A}}_{(K)} \boldsymbol{x}_0, \qquad (2.4)$$

where the measurement vector  $\tilde{\boldsymbol{y}}_{(K)} \in \mathbb{R}^{Km}$  and the observability matrix  $\tilde{\boldsymbol{A}}_{(K)} \in \mathbb{R}^{Km \times N}$ are defined as

$$\widetilde{\boldsymbol{y}}_{(K)} = \begin{bmatrix} \boldsymbol{y}_0 \\ \boldsymbol{y}_1 \\ \vdots \\ \boldsymbol{y}_{K-1} \end{bmatrix}, \quad \widetilde{\boldsymbol{A}}_{(K)} = \begin{bmatrix} \boldsymbol{A}_{(0)} \\ \boldsymbol{A}_{(1)}\boldsymbol{D} \\ \vdots \\ \boldsymbol{A}_{(K-1)}\boldsymbol{D}^{K-1} \end{bmatrix}. \quad (2.5)$$

In order to ensure the recovery of  $\boldsymbol{x}_0$  from (2.4) using sparse signal recovery techniques, we need to analyze the RIP of the observability matrix  $\widetilde{\boldsymbol{A}}_{(K)}$ . This, in turn, yields bounds on the number of measurement vectors required to recover any sparse initial state. Before launching into the RIP analysis, we note that an overall scaling does not affect the RIP of a matrix. Now, let  $\lambda_{\max} \neq 0$  be the largest singular value of **D**. We can rewrite (2.4) as

$$\widetilde{\boldsymbol{L}}(\boldsymbol{\lambda}_{\max})\widetilde{\boldsymbol{y}}_{(K)} = \widetilde{\boldsymbol{L}}(\boldsymbol{\lambda}_{\max})\widetilde{\boldsymbol{A}}_{(K)}\boldsymbol{x}_{0}, \qquad (2.6)$$

where  $\boldsymbol{\lambda}_{\max} = \begin{bmatrix} 1 & \lambda_{\max} & \dots & \lambda_{\max}^{K-1} \end{bmatrix}^{\mathsf{T}} \in \mathbb{R}^{K}$  and the matrix function  $\widetilde{\boldsymbol{L}}(\boldsymbol{\lambda}) : \mathbb{R}^{K} \to \mathbb{R}^{Km \times Km}$  is defined as

$$\widetilde{\boldsymbol{L}}(\boldsymbol{\lambda}) = \frac{1}{\sqrt{Km}} \begin{bmatrix} \boldsymbol{\lambda}_1 \boldsymbol{I} & & \\ & \boldsymbol{\lambda}_2 \boldsymbol{I} & \\ & & \ddots & \\ & & & \boldsymbol{\lambda}_K \boldsymbol{I} \end{bmatrix}^{-1}.$$
(2.7)

Therefore, we get the following relation:

$$\widetilde{\boldsymbol{L}}(\boldsymbol{\lambda}_{\max})\widetilde{\boldsymbol{A}}_{(K)} = \begin{bmatrix} \boldsymbol{A}_{(0)} \\ \boldsymbol{A}_{(1)}\overline{\boldsymbol{D}} \\ & \\ & \\ & \\ \boldsymbol{A}_{(K-1)}\overline{\boldsymbol{D}}^{K-1} \end{bmatrix}$$
(2.8)

where  $\bar{D} = D/\lambda_{\text{max}}$  has the largest and the smallest singular values as 1 and  $\lambda$ , respectively. Here,  $\lambda$  is the ratio of the smallest to the largest singular value of D. Analyzing the recoverability of  $\boldsymbol{x}_0$  from (2.6), which is equivalent to (2.4), requires one to study the RIP of the matrix  $\tilde{\boldsymbol{L}}(\boldsymbol{\lambda}_{\text{max}})\tilde{\boldsymbol{A}}_{(K)}$ . Therefore, in this chapter, we focus on the RIP of such a scaled version of  $\tilde{\boldsymbol{A}}_{(K)}$ .

# 2.3 Preliminaries

In this section, we define a subgaussian random matrix and summarize some of its properties.

**Definition 2.2** (Subgaussian random variable). A random variable A is said to be subgaussian with parameter c if, for any  $\theta \in \mathbb{R}$ ,

$$\mathbb{E}\left\{\exp\left(\theta A\right)\right\} \le \exp\left(c\theta^{2}\right).$$
(2.9)

**Definition 2.3** (Subgaussian random matrix). A random matrix  $\mathbf{A} \in \mathbb{R}^{m \times N}$  is said to be a subgaussian random matrix if its entries are independent zero mean and unit variance subgaussian random variables with common parameter c.

The subgaussian random matrix includes a large class of random matrices including independent and identically distributed (i.i.d.) Gaussian random matrices, and i.i.d. Bernoulli random matrices, etc. Next, we present two results that are necessary for the derivation of the main results in the chapter.

**Lemma 2.1.** If A is a subgaussian random variable with parameter c, then  $A^2 - \mathbb{E} \{A^2\}$ is a subexponential random variable with parameter 16c, i.e., for  $|\theta| \leq \frac{1}{16c}$ , we have

$$\mathbb{E}\left\{\exp\left[\theta\left(A^2 - \mathbb{E}\left\{A^2\right\}\right)\right]\right\} \le \exp\left(128\theta^2 c^2\right).$$
(2.10)

*Proof.* See [76, Lemma 1.12].

**Proposition 2.1** (Bernstein-type inequality). Let  $\{A_l\}_{l=1,2,...,m}$  be independent subexponential random variables such that  $a_{\min} \leq \mathbb{E}\{A_l\} \leq a_{\max}$ . That is, for all  $t \geq 0$ ,

$$\mathbb{P}\left\{|A_l - \mathbb{E}\left\{A_l\right\}| \ge t\right\} \le c_1 \exp\left(-c_2 t\right),\tag{2.11}$$

for  $l = 1, 2, \ldots, m$ , and some constants  $c_1, c_2 > 0$ . Then, for any  $t > m \max\{a_{\max}, -a_{\min}\}$ ,

$$\mathbb{P}\left\{\left|\sum_{l=1}^{m} A_{l}\right| \ge t\right\} \le \exp\left(-\frac{c_{2}^{2}(t - ma_{\max})^{2}/2}{m(2c_{1} + c_{2}a_{\max}) + c_{2}t}\right) + \exp\left(-\frac{c_{2}^{2}(t + ma_{\min})^{2}/2}{m(2c_{1} + c_{2}a_{\min}) + c_{2}t}\right).$$
(2.12)

*Proof.* See Appendix A.1.

## 2.4 **RIP** For Independent Observation Matrices

In this section, we present the first main result of the chapter and its implications.

**Theorem 2.1** (Independent random observation matrices). Suppose measurement matrices  $A_{(k)}, k = 0, 1, ..., K-1$  are independent subgaussian random matrices with parameter c. Then, if

$$Km\left(\delta - 1 + \lambda^{2(K-1)}\right)^2 \ge \tilde{c}\left[9s\ln\left(\frac{eN}{s}\right) + 2\ln\left(\frac{2}{\epsilon}\right)\right],$$
 (2.13)

the RIC  $\delta_s$  of the scaled version  $\widetilde{\boldsymbol{L}}(\boldsymbol{\lambda}_{\max})\widetilde{\boldsymbol{A}}_{(K)}$  of  $\widetilde{\boldsymbol{A}}_{(K)}$  satisfies  $\delta_s < \delta$  for all  $1 - \lambda^{2(K-1)} < \delta < 1$  with probability at least  $1 - \epsilon$ . Here,  $\widetilde{c}$  is a constant dependent only on c, and  $\lambda \leq 1$  is the ratio of the smallest to the largest singular values of  $\boldsymbol{D}$ . When (2.13) holds, the system is observable for sufficiently large  $\lambda$  with high probability.

*Proof.* See Appendix A.2.

Here, we note that the phrase "sufficiently large"  $\lambda$  refers to the fact that it must be

large enough so as to be able to meet the upper bound on the RIC set by the RIP based guarantees of different algorithms, as discussed in Section 1.1. We discuss this point in detail in Section 2.4.2.

Next, using the proof technique of the above theorem, we can show the following interesting corollaries. We omit their proofs as they are straightforward. The first corollary extends Theorem 2.1 to the case when the measurements are not necessarily taken over a contiguous set of time instants.

**Corollary 2.1** (Non-consecutive measurements). Suppose the available measurements are  $\{\boldsymbol{y}_k\}_{k\in\mathcal{K}}$ , where  $\mathcal{K}$  is an index set of cardinality K, and  $\boldsymbol{A}_{(k)}, k = 0, 1, \dots, K-1$  are independent subgaussian random matrices with parameter c. Then, if

$$Km\left(\delta - 1 + \lambda^{2(\max\{\mathcal{K}\}-1)}\right)^2 \ge \tilde{c}\left[9s\ln\left(\frac{eN}{s}\right) + 2\ln\left(\frac{2}{\epsilon}\right)\right],\tag{2.14}$$

then the RIC  $\delta_s$  of the scaled version  $\widetilde{\mathbf{L}}(\boldsymbol{\lambda}_{\max,\mathcal{K}}) \widetilde{\mathbf{A}}_{(K)}$  of  $\widetilde{\mathbf{A}}_{(K)}$  satisfies  $\delta_s < \delta$  for all  $1 - \lambda^{2\max\{\mathcal{K}\}} < \delta < 1$  with probability at least  $1 - \epsilon$ . Here,  $\widetilde{c}$  is a constant dependent only on c, and  $\lambda \leq 1$  is the ratio of the smallest to the largest singular values of  $\mathbf{D}$ . Also,  $\boldsymbol{\lambda}_{\max,\mathcal{K}} \in \mathbb{R}^K$  has  $j^{\text{th}}$  entry as  $\lambda_{\max}^{\widetilde{j}}$ . Hence, when (2.15) holds, the system is observable for sufficiently large  $\lambda$  with high probability.

The next corollary extends Theorem 2.1 to the case when  $\boldsymbol{x}_0$  is sparse under an arbitrary basis  $\Psi \in \mathbb{R}^{N \times N}$  rather than the canonical basis.

**Corollary 2.2** (Sparsifying basis other than the canonical basis). Suppose the observation matrices  $\mathbf{A}_{(k)}, k = 0, 1, ..., K - 1$  are independent subgaussian random matrices with parameter c, and the initial state is sparse under the basis  $\Psi \in \mathbb{R}^{N \times N}$ , which need not be the canonical basis. Then, if

$$Km\left(\delta - 1 + \widetilde{\lambda}^2\right)^2 \ge \widetilde{c}\left[9s\ln\left(\frac{eN}{s}\right) + 2\ln\left(\frac{2}{\epsilon}\right)\right],\tag{2.15}$$

the RIC  $\delta_s$  of the scaled version  $\widetilde{\mathbf{L}}\left(\widetilde{\boldsymbol{\lambda}}_{\max}\right)\widetilde{\mathbf{A}}_{(K)}$  of  $\widetilde{\mathbf{A}}_{(K)}$  satisfies  $\delta_s < \delta$ , for all  $1 - \widetilde{\lambda}^2 < \delta < 1$ , with probability at least  $1 - \epsilon$ . Here,  $\widetilde{c}$  is a constant dependent only on c, and  $\widetilde{\boldsymbol{\lambda}} \leq 1$  is the ratio of the smallest to the largest singular value of  $\mathbf{D}^{(K-1)}\Psi$ . Also,  $\widetilde{\boldsymbol{\lambda}}_{\max} \in \mathbb{R}^K$  has  $j^{\text{th}}$  entry as the largest singular value of  $\mathbf{D}^{j-1}\Psi$ . Hence, when (2.15) holds, the system is observable for sufficiently large  $\widetilde{\boldsymbol{\lambda}}$  with high probability.

It is also interesting to consider guarantees for the case where the matrix D is an RIPcompliant matrix. The following corollary gives guarantees similar to Theorem 2.3 based on the RIC of an RIP-compliant  $N \times N$  matrix D.

**Corollary 2.3** (Relaxation based on the RIP of the transfer matrix). Suppose  $A_{(k)}, k = 0, 1, ..., K-1$  are independent subgaussian random matrices with parameter c. Then, if

$$Km\left(\delta - 1 + \bar{\lambda}^{2(K-2)}(1 - \tilde{\delta}_s)^2\right)^2 \ge \tilde{c}\left[9s\ln\left(\frac{eN}{s}\right) + 2\ln\left(\frac{2}{\epsilon}\right)\right],\tag{2.16}$$

then the RIC  $\delta_s$  of the scaled version  $\widetilde{\boldsymbol{L}}(\boldsymbol{\lambda}_{\max})\widetilde{\boldsymbol{A}}_{(K)}$  of  $\widetilde{\boldsymbol{A}}_{(K)}$  satisfies  $\delta_s < \delta$  for all  $1 - \overline{\lambda}^{2(K-2)}(1-\widetilde{\delta}_s)^2 < \delta < 1$  with probability at least  $1-\epsilon$ . Here,  $\widetilde{c}$  is a constant dependent only on c, and  $\overline{\lambda} < 1$  and  $\widetilde{\delta}_s \leq 1$  are the smallest nonzero singular value and the RIC of  $\boldsymbol{D}$  normalized to unit spectral norm. Hence, when (2.16) holds, the system is observable for sufficiently small  $\widetilde{\delta}_s$  with high probability.

*Proof.* When the matrix D is normalized to unit spectral norm, for any unit norm s-sparse

vector  $\boldsymbol{z} \in \mathbb{R}^N$ , we have

$$\left\|\boldsymbol{D}^{k}\boldsymbol{z}\right\| \geq \bar{\lambda}^{k-1} \left\|\boldsymbol{D}\boldsymbol{z}\right\| \geq \bar{\lambda}^{k-1}(1-\tilde{\delta}_{s}), \qquad (2.17)$$

since the vector Dz belongs to the column space of D. Thus, we can replace  $\lambda^{(K-1)}$  with  $\bar{\lambda}^{(K-2)}(1-\tilde{\delta}_s)$  in Theorem 2.3 to obtain the desired result.

We note that  $\bar{\lambda} \geq \lambda$  and  $1 - \tilde{\delta}_s \geq \lambda$ , and thus the above corollary is a stronger result than Theorem 2.1. However,  $\lambda$  is easier to compute than the RIC constant of D. In the following subsections, we discuss some implications of the above results.

## 2.4.1 Special Cases

- 1. Suppose D is a scaled unitary matrix. Then,  $\lambda = 1$ , and Theorem 2.1 simplifies to the recovery condition for the standard compressed sensing problem with Kmmeasurements. Since the RIP of a matrix is invariant to multiplication by a unitary matrix, each new observation vector adds m new measurements to (2.4) as Kincreases.
- 2. Suppose D is rank-deficient. Then,  $\lambda = 0$ , and (2.13) does not hold for any  $\delta < 1$ , unless the following holds:

$$m \ge \frac{\widetilde{c}}{\delta^2} \left[ 9s \ln\left(\frac{eN}{s}\right) + 2\ln(2\epsilon^{-1}) \right].$$
(2.18)

This is intuitive, because when  $\boldsymbol{x}_0$  lies in the null space of  $\boldsymbol{D}$ ,  $\boldsymbol{y}_k = 0$  for  $k \geq 1$ . Hence, the system is observable if it is observable from  $\boldsymbol{y}_0$ . Thus, the uniform recovery guarantee does not hold for a rank deficient  $\boldsymbol{D}$ .

- 3. Suppose that D is an ill-conditioned matrix, i.e.,  $\lambda$  is close to zero. Then, the upper bound on  $\delta$  required to guarantee observability may not hold [29,77,78]. This is because right multiplication of a matrix by another ill-conditioned matrix may severely degrade its RIP. However, Corollary 2.3 guarantees that if  $\tilde{\delta}_s \neq 0$ , it is possible to recover  $\boldsymbol{x}_0$  even if (2.18) is not satisfied.
- 4. For K = 1, Theorem 2.1 reduces to the recovery condition of the standard compressed sensing problem [32]. Also, if the system is observable with m measurements (for example, when (2.18) is satisfied), the conditions in Theorem 2.1 hold for K = 1, as expected.

## 2.4.2 Number of Measurements

Theorem 2.1 shows that  $Km = \mathcal{O}(s \ln(N/s))$  is sufficient for observability. Note that the number of measurements are independent of the scaling of D. Thus, the number of measurements can be greatly reduced for large dimensional systems. In contrast,  $Km = \mathcal{O}(N)$  measurements are necessary for observability of a general non-sparse initial state vector. We also recall from Section 1.1 that the initial state can be recovered using any of the compressed sensing techniques like basis pursuit, thresholding algorithms, or greedy algorithms.

The RIP based recovery guarantees available in the literature set an upper bound on the RIC. For example, using the necessary and sufficient condition for  $\ell_1$  based recovery:  $\delta_s \leq 1/3$  [29], (2.13) reduces to

$$K\left(\lambda^{2(K-1)} - 2/3\right)^2 \ge \frac{\widetilde{c}}{m} \left[9s\ln\left(\frac{eN}{s}\right) + 2\ln(2\epsilon^{-1})\right],\tag{2.19}$$

for  $\lambda^{2(K-1)} \geq 2/3$ . In other words, if (2.19) is satisfied for some K which is less than  $\lfloor (\ln(2/3))/(2\ln(\lambda)) \rfloor + 1$ , then the system is observable. However, note that, if the system is observable for  $K_1$  measurements, it remains observable for  $K > K_1$ .

We note that  $K(\lambda^{2(K-1)} - 2/3)^2$  is an increasing function of K, which gives a lower bound m from (2.19). Therefore, for  $\lambda < 1$ ,

$$m = \mathcal{O}\left(\frac{\ln(N/s)}{K\left(\lambda^{2(K-1)} - 2/3\right)^2}\right).$$
 (2.20)

We also note that value of m required decreases with  $\lambda$  and K. This is in agreement with the fact that as K increases, we get more measurements and a smaller m suffices for ensuring successful recovery of the initial state. Also, as  $\lambda$  increases, the matrix Dbecomes better conditioned, and, consequently, a smaller value of m is sufficient for exact recovery.

## 2.4.3 **RIP of the Product of Matrices**

We can derive an sufficient condition for the product of a subgaussian matrix and a deterministic matrix to satisfy the RIP property as follows:

**Corollary 2.4.** Suppose  $A \in \mathbb{R}^{m \times N}$  is subgaussian random matrix with parameter c. If

$$m\left(\delta - 1 + \lambda^2\right)^2 \ge \tilde{c}\left[9s\ln\left(\frac{eN}{s}\right) + 2\ln\left(\frac{2}{\epsilon}\right)\right],$$
(2.21)

the RIC  $\delta_s$  of a suitably scaled version of AD satisfies  $\delta_s < \delta$ , for all  $1 - \lambda^2 < \delta < 1$ , with probability at least  $1 - \epsilon$ . Here,  $\tilde{c}$  is a constant dependent only on c, and  $\lambda \leq 1$  is the ratio of the smallest to the largest singular values of D.

Corollary 2.4 is an immediate by-product of the proof of Theorem 2.1, but it is an

interesting and potentially useful result in its own right, as it provides conditions under which right-multiplication of a subgaussian random matrix by a deterministic matrix Dpreserves its RIP.

### 2.4.4 Extension to Robust Recovery

The RIP based analysis allows us to extend Theorem 2.1 to bound the  $\ell_1$  and  $\ell_2$  norm error in recovery of the initial state under bounded noise and model mismatch. These impairments correspond to the cases when the measurements are noisy and the initial state is not exactly sparse, respectively. In this case, the system model modifies as follows:

$$\boldsymbol{x}_k = \boldsymbol{D}^k (\boldsymbol{x}_0 + \check{\boldsymbol{x}}_0) \tag{2.22}$$

$$\boldsymbol{y}_k = \boldsymbol{A}_{(k)} \boldsymbol{x}_k + \boldsymbol{w}_k, \qquad (2.23)$$

for discrete time instants k = 0, 1, ..., K - 1. Here,  $\boldsymbol{w}_k \in \mathbb{R}^m$  denotes the bounded measurement noise:  $\|\boldsymbol{w}_k\| \leq W$ ; while  $\check{\boldsymbol{x}}_0 \in \mathbb{R}^N$  represents the error in approximating the initial state by an *s*-sparse vector. That is,  $\boldsymbol{x}_0 = \underset{\boldsymbol{v} \in \mathbb{R}^N : \|\boldsymbol{v}\|_0 \leq s}{\arg \min} \|\boldsymbol{x}_0 + \check{\boldsymbol{x}}_0 - \boldsymbol{v}\|$ . Therefore, the overall set of equations can be written as

$$\tilde{\boldsymbol{y}}_{(K)} = \tilde{\boldsymbol{A}}_{(K)}(\boldsymbol{x}_0 + \check{\boldsymbol{x}}_0) + \tilde{\boldsymbol{w}}, \qquad (2.24)$$

where the bounded noise  $\tilde{\boldsymbol{w}} \in \mathbb{R}^{Km}$  satisfies  $\|\tilde{\boldsymbol{w}}\| \leq \sqrt{K}W$ .

**Corollary 2.5.** Suppose  $A_{(k)}$ , k = 0, 1, ..., K - 1 are independent subgaussian random matrices with parameter c. Suppose that, for some integer p > 0 and positive number  $c_{th}$ ,

$$Km\left(c_{th}-1+\lambda^{2(K-1)}\right)^{2} \geq \tilde{c}\left[9ps\ln\left(\frac{eN}{ps}\right)+2\ln\left(\frac{2}{\epsilon}\right)\right],\tag{2.25}$$

and  $\lambda^{2(K-1)} > 1 - c_{th}$ . Here,  $\tilde{c}$  is a constant dependent only on c, and  $\lambda \leq 1$  is the ratio of the smallest to the largest singular values of D. Then, with probability at least  $1 - \epsilon$ , the initial vector  $\mathbf{x}_0 + \check{\mathbf{x}}_0$  can be recovered from (2.24) with errors as follows:

$$\|\boldsymbol{x}_{0} + \check{\boldsymbol{x}}_{0} - \hat{\boldsymbol{x}}_{0}\|_{1} \le c_{1} \|\check{\boldsymbol{x}}_{0}\|_{1} + c_{2} \sqrt{\frac{s(1 - \lambda_{\max}^{-2K})}{Km(1 - \lambda_{\max}^{-2})}} W$$
(2.26)

$$\|\boldsymbol{x}_{0} + \check{\boldsymbol{x}}_{0} - \hat{\boldsymbol{x}}_{0}\| \leq \frac{c_{1}}{\sqrt{s}} \|\check{\boldsymbol{x}}_{0}\|_{1} + c_{2} \sqrt{\frac{(1 - \lambda_{\max}^{-2K})}{Km(1 - \lambda_{\max}^{-2})}} W,$$
(2.27)

where  $\hat{\boldsymbol{x}}_0$  is the estimate of the initial vector, and  $c_1, c_2 > 0$  are universal constants. The constants p and  $c_{th}$  depend on the recovery algorithms as follows:

- For BP: p = 2 and  $c_{th} = \frac{4}{\sqrt{41}}$ .
- For IHT: p = 6 and  $c_{th} = \frac{1}{\sqrt{3}}$ .

• For compressive sampling matched pursuit (CoSAMP): p = 8 and  $c_{th} = \frac{\sqrt{\sqrt{11/3}-1}}{2}$ .

*Proof.* Follows from the upper bound on the RIC required by the different algorithms to ensure robust recovery [32, Theorem 6.12, 6.21, 6.28].  $\Box$ 

We note the dependence on  $\lambda_{\text{max}}$  in the above expressions is not unexpected: it arises because of the scaling of the measurement matrix. The scaling operation is reasonable due to the following reasons:

- One can always scale the linear equations with no information loss. The scaling operation neither changes the problem nor affects any intuitive notion of SNR.
- The scaling matrix is diagonal, and therefore does not introduce any correlation between the noise terms which might affect the recovery. Moreover, the recovery

guarantees of the algorithms listed in Corollary 2.5 depend only on the  $\ell_2$  norm of the noise vector, and are independent of the individual variances of the noise terms.

- Note that  $\lambda_{\max}$  determines the effective SNR of the system, and hence it plays an important role in recoverability of the initial state. The effect of  $\lambda_{\max}$  appears as the factor  $\sqrt{s}\sqrt{\frac{(1-\lambda_{\max}^{-2K})}{K(1-\lambda_{\max}^{-2})}}\frac{W}{\sqrt{m}}$  in (2.26). Here,  $\sqrt{s}$  and  $W/\sqrt{m}$  capture the same effect as those of the sparsity s and the average noise power per measurement  $W/\sqrt{m}$ , respectively, in the standard compressed sensing results. Further, we intuitively examine the term  $\sqrt{\frac{(1-\lambda_{\max}^{-2K})}{K(1-\lambda_{\max}^{-2})}}$  via three special cases of  $\lambda_{\max}$  below:
  - (i)  $\lambda_{\max} \gg 1$ : When  $\lambda_{\max}$  is large, this term reduces to  $1/\sqrt{K}$ , which has no dependence on  $\lambda_{\max}$ . This is because the effective SNR is large, and hence the noise term is negligible, for all measurements except for the first measurement vector,  $\boldsymbol{y}_0$ . Thus, we have one noisy and K-1 noiseless measurements, which leads to an error bound that decreases with K.
  - (ii)  $\lambda_{\max} \approx 1$ : When  $\lambda_{\max}$  is close to 1, this term reduces to 1. This is equivalent to having K noisy measurements with equal scaling factor and thus the error bound per measurement is independent of K. In this case, the advantage of having multiple observations comes in terms of the Km dependence of the number of measurements in (2.25).
  - (iii)  $\lambda_{\max} \ll 1$ : When  $\lambda_{\max}$  is small, this term reduces to  $\lambda_{\max}^{-(K-1)}/\sqrt{K}$ , which is a new dependence. In this case, the noise in the later measurements gets amplified by the scaling factor. Hence, the noise term in the last measurement dominates the average noise power. However, in practice, one would consider the smallest

value of K for which (2.25) is satisfied, and substitute that value of K in (2.26)and (2.27) to get the bound on robust recovery of the initial state.

## 2.4.5 Comparison With Prior Work

In [63,64], the authors address the same problem as ours and give a sufficient condition on number of measurements Km for successful recovery. In this subsection, we compare and contrast the two results. We begin with the result from [63,64], stated in our notation.

**Theorem 2.2** (Prior work [63,64]). Suppose that  $\mathbf{D} = a\mathbf{U}$  where  $a \neq 0$  and  $\mathbf{U} \in \mathbb{R}^{N \times N}$ is unitary. Define  $b \triangleq \sum_{k=1}^{K} a^{2(k-1)}$ . Assume  $\mathbf{A}_{(k)}, k = 0, 1, \dots, K-1$  are independent Gaussian random matrices with mean zero and variance 1/m. Then, if

$$Km\delta^2 \ge 512 \left[ s\ln\left(\frac{42}{\delta}\right) + 1 + \ln\left(\frac{N}{s}\right) + \ln\left(\frac{2}{\epsilon}\right) \right] \left[ \frac{\|1 - a^2\|K + \min\{1, a^2\}}{\max\{1, a^2\}} \right]$$
(2.28)

the RIC  $\delta_s$  of  $\frac{1}{\sqrt{b}}\tilde{A}_{(K)}$  satisfies  $\delta_s < \delta < 1$  with probability at least  $1 - \epsilon$ .

We make the following observations:

- Restriction on D: Theorem 2.2 is applicable only when D is a scaled unitary matrix. Reference [64] extends the result to a certain type of positive definite matrices. Our results are more general, and hold true for any arbitrary matrix  $D \neq 0$ .
- Bound for scaled unitary matrices: For the special case of D = aU, (2.13) reduces to the following:

$$Km\delta^2 \ge \tilde{c} \left[9s\ln\left(\frac{eN}{s}\right) + 2\ln\left(\frac{2}{\epsilon}\right)\right],$$
(2.29)

for  $0 < \delta < 1$ . We see that there is an extra term on the right hand side of (2.28) of

Theorem 2.2. We can bound this term as follows:

$$\frac{\|1 - a^2\|K + \min\{1, a^2\}}{\max\{1, a^2\}} \ge \frac{\|1 - a^2\| + \min\{1, a^2\}}{\max\{1, a^2\}} = 1,$$
(2.30)

for all  $a \neq 0$ . Hence, our results are stronger than Theorem 2.2 for the scaled unitary matrix case.

• Dependency on the eigenvalue: The condition (2.28) heavily depends on the eigenvalue a of D. The least number of measurements Km are required for |a| = 1, and as |a| moves away from unity, the lower bound on Km increases. However, our results depend only on the ratio of the smallest to the largest singular value of D, and therefore gives the best bound for all values of a. This is because our results make use of the fact that the recovery properties are independent of scaling due to the equivalence of (2.4) and (2.6). This critical observation allowed us to get stronger results compared to Theorem 2.2.

### 2.4.6 Extension to Identical Observation Matrices Case

Suppose we carry out a similar analysis for the case when all observation matrices are identical  $\mathbf{A}_{(k)} = \mathbf{A}$  for k = 0, 1, ..., K-1, where  $\mathbf{A}$  is a subgaussian random matrix with parameter c. The sufficient condition then obtained shows that the system is recoverable if (2.18) is satisfied. However, this condition ensures that the system is observable with K =1. This is a weak result, because it implies that the availability of additional measurements does not improve the sufficient condition for observability. This is indeed true when  $\mathbf{D} = \alpha \mathbf{I}$ , for some  $\alpha \in \mathbb{R}$ , because we are only adding scaled versions of the rows of  $\mathbf{A}$  to  $\widetilde{\mathbf{A}}_{(K)}$  as K increases. For general  $\mathbf{D}$ , a different proof technique has to be used, which is discussed in the next section.

## 2.5 **RIP For Identical Observation Matrices Case**

In this section, we present a result on the RIP of the observability matrix when the observation matrices are identical random matrices. First, we define the following quantities:

$$\widetilde{\boldsymbol{D}}_{(K,i)} \triangleq \begin{bmatrix} \boldsymbol{I}_i & \boldsymbol{D}_i & \dots & \boldsymbol{D}_i^{K-1} \end{bmatrix},$$
 (2.31)

$$L(\boldsymbol{D},K) \triangleq \max_{i} \left\| \widetilde{\boldsymbol{D}}_{(K,i)} \right\|_{2},$$
 (2.32)

where  $I_i$  is the *i*<sup>th</sup> column of identity matrix of size  $N \times N$  and  $D_i^k$  is the *i*<sup>th</sup> column of matrix  $D^k$ .

**Theorem 2.3** (Identical random observation matrices). Suppose all the observation matrices are identical, i.e.,  $\mathbf{A}_{(k)} = \mathbf{A}$  for k = 0, 1, ..., K - 1, where  $\mathbf{A}$  is a subgaussian random matrix with parameter c. Then, if

$$Km\frac{\left(\delta-1+\lambda^{2(K-1)}\right)^2}{L^2\left(\boldsymbol{D},K\right)} \ge \tilde{c}s\max\left\{\ln^2 s\ln^2 N,\ln(2\epsilon^{-1})\right\},\tag{2.33}$$

then the RIC  $\delta_s$  of the scaled version  $\widetilde{\boldsymbol{L}}(\boldsymbol{\lambda}_{\max})\widetilde{\boldsymbol{A}}_{(K)}$  of  $\widetilde{\boldsymbol{A}}_{(K)}$  satisfies  $\delta_s < \delta$  for all  $1 - \lambda^{2(K-1)} < \delta < 1$  with probability at least  $1 - \epsilon$ . Here,  $\widetilde{c}$  is a constant dependent only on c, and  $\lambda \leq 1$  is the ratio of the smallest to the largest singular values of  $\boldsymbol{D}$ . Hence, when (2.33) holds, the system is observable for sufficiently large  $\lambda$  with high probability.

*Proof.* See Appendix A.3.

In the following subsections, we provide more insights into the above results.

## 2.5.1 Special Cases

- 1. Suppose D is a scaled identity matrix. Then,  $\lambda = 1$ , and  $L^2(D, K) = K$ , and hence from Theorem 2.3, we retrieve the recovery condition for a standard compressed sensing problem with m measurements, and the guarantee does not improve with increasing K. This is intuitive, because we are only adding scaled versions of the rows of A to  $\widetilde{A}_{(K)}$  as K increases.
- 2. Suppose D is rank-deficient. Then,  $\lambda = 0$ , and (2.33) does not hold for any  $\delta < 1$ , unless the following holds:

$$m \ge \widetilde{cs} \max\left\{\ln^2 s \ln^2 N, \ln(2\epsilon^{-1})\right\},\tag{2.34}$$

as expected.

- 3. Suppose that D is ill-conditioned, i.e.,  $\lambda$  is close to zero. Then, the upper bound on  $\delta$  required to guarantee observability may not hold [29,77,78], which is in similar vein as explained in the case of Theorem 2.1.
- 4. For K = 1, Theorem 2.3 reduces to the recovery condition of the standard compressed sensing problem [32]. Also, if the system is observable with m measurements (for example, when (2.34) is satisfied), the conditions in Theorem 2.3 hold for K = 1, as expected.

## 2.5.2 Number of Measurements

Theorem 2.3 shows that  $Km = \mathcal{O}(s \ln^2 s \ln^2 N)$  is sufficient for observability, whereas  $\mathcal{O}(N)$  measurements are necessary for observability of a non-sparse initial state vector.

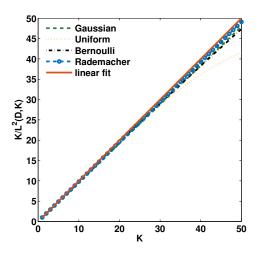


Figure 2.1: Variation of  $K/L^2(\mathbf{D}, K)$  with K when  $\mathbf{D}$  generated using different distributions. We see that  $K/L^2(\mathbf{D}, K)$  is a (linearly) non-decreasing function of K.

Also, as mentioned in Section 1.1, the initial state can be recovered using any of the compressed sensing techniques like basis pursuit, thresholding algorithms, or greedy algorithms. As in the case of Theorem 2.1, the RIP based guarantees fix an upper bound on K, and hence a lower bound on m. However, note that, if the system is observable for  $K_1$  measurements, it remains observable for  $K > K_1$ .

The main difference between the results in Theorem 2.1 and Theorem 2.3 is in the  $L^2(\mathbf{D}, K)$  term. Hence, in order to gain intuition on the number of measurements required in the identical observation matrices case, we study the behavior of the  $L(\mathbf{D}, K)$  term in the following proposition.

**Proposition 2.2.** The term  $K/L^2(\mathbf{D}, K)$  satisfies

$$1 \le K/L^2\left(\boldsymbol{D}, K\right) \le K. \tag{2.35}$$

*Proof.* See Appendix A.8.

40

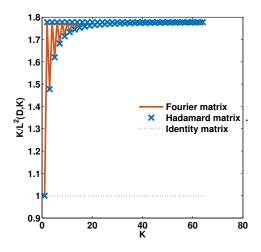


Figure 2.2: Variation of  $K/L^2(\mathbf{D}, K)$  with K for Fourier, Hadamard and identity constructions of  $\mathbf{D}$ . We see that  $K/L^2(\mathbf{D}, K)$  is not an increasing function of K.

We note that the upper and the lower bounds are achieved by D = 0, and D = I, respectively. Further, as discussed in Section 2.5.1, both cases are not favorable from the observability point of view. Although the D = 0 case achieves the upper bound of the term  $K/L^2(D, K)$ , this choice is not desirable since  $\lambda = 0$ .

In Figure 2.1, we empirically illustrate that if D is randomly chosen, the upper bound can be nearly achieved. Also, when D is a random matrix,  $\lambda$  is non-zero with high probability [79], which makes this choice desirable. Random transfer matrices occur in some linear dynamical systems with sparse initial state, which models homogeneous spreading of epidemic or information or computer virus. For example, a dynamical system corresponding to a epidemic spread can be modeled using a Erdos-Renyi model in which case the transfer matrix has independent Bernoulli distributed entries [80–83]. In Figure 2.1, we use N = 50 and vary K from 1 to N [80]. The entries of D are drawn from the following four distributions:

1. Standard Gaussian distribution

- 2. Uniform distribution on [0, 1]
- 3. Bernoulli distribution with parameter 0.5
- 4. Rademacher distribution.

The curve f(K) = K is labeled as *linear fit*. The value of  $K/L^2(\mathbf{D}, K)$  is averaged over 100 trials and plotted along the Y-axis as a function of K. In all the four cases, the behavior of the curves is nearly linear, and  $K/L^2(\mathbf{D}, K) \approx K$ .

Before we discuss the implications of the result, we first give some intuition on reason behind this linear behavior. Since  $L^2(\boldsymbol{D}, K)$  is a complicated function of  $\boldsymbol{D}$ , we focus on the regime where N is large and the distribution of the entries of  $\boldsymbol{D}$  is Gaussian. We note that, from the Marchenko-Pastur theorem [84], the spectral norm of a Gaussian matrix with zero mean and unit variance is close to  $2\sqrt{N}$ . Thus, after normalization, as K increases,  $\boldsymbol{D}^K$  goes to  $\boldsymbol{0}$ , and the extra column that gets added to the matrix  $\tilde{\boldsymbol{D}}_{K,i}$ as K increases is approximately  $\boldsymbol{0}$ . Therefore, we have  $\|\tilde{\boldsymbol{D}}_{(K,i)}\|_2 \approx \|\tilde{\boldsymbol{D}}_{(2,i)}\|_2$ . Further,  $\|\tilde{\boldsymbol{D}}_{(2,i)}\|_2$  is the same as the largest eigenvalue of the matrix:  $\begin{bmatrix} 1 & \boldsymbol{D}_{ii} \\ \boldsymbol{D}_{ii} & \|\boldsymbol{D}_{i}\|^2 \end{bmatrix}$ . Also, for large N,  $\|\boldsymbol{D}_i\|^2 \approx 1$  and  $\boldsymbol{D}_{ii}^2 \approx 0$ , which gives  $\|\tilde{\boldsymbol{D}}_{(K,2)}\|_2 \approx 1$ . Hence, we have the following relation:

$$L(\boldsymbol{D}, K) = \arg\max_{i} \left\| \tilde{\boldsymbol{D}}_{(K,i)} \right\|_{2} \approx \left\| \tilde{\boldsymbol{D}}_{(K,2)} \right\|_{2} \approx 1.$$
(2.36)

Thus, intuitively,  $K/L^2(\mathbf{D}, K) \approx K$  for sufficiently large N. Also, from Figure 2.1, we see that N = 50 is large enough for the argument to hold.

Next, we discuss some implications of Figure 2.1. The RIP of the observability matrix  $\widetilde{A}_{(K)}$  is on par with an unstructured random Gaussian matrix. In turn, this suggests that it is not necessary to use independent observation matrices to ensure observability;

identical observation matrices result in a penalty of only  $O(\ln^2 s \ln N)$  in terms of the number of measurements required, provided D is a random matrix. Hence, we conclude that even though I and 0 are poor choices for D, matrices with good recovery properties are plenty in number.

Another important observation from the plot is that  $K/L^2(\mathbf{D}, K)$  is a non-decreasing function of K. Thus, as K increases, we need a smaller value of m for exact recovery of the sparse initial state. That is, one can trade-off K and m while ensuring observability. The following result shows that the function is increasing for the special case when  $\mathbf{D}$  is a positive semi-definite (psd) matrix.

**Proposition 2.3.** When D is a psd matrix, the function  $K/L^2(D, K)$  is a non-decreasing function of K.

```
Proof. See Appendix A.9.
```

Remark 1: The above result does not, in general, imply that  $K/L^2(\mathbf{D}, K)$  increases with K. For example, from Figure 2.2, we see that  $K/L^2(\mathbf{D}, K)$  for standard dictionaries like Fourier and Hadamard matrices is not strictly increasing with K.

*Remark 2:* Theorem 2.3 also has corollaries similar to Corollary 2.1-Corollary 2.3 and Corollary 2.5. We omit those results to avoid repetition.

# 2.6 Joint Recovery of Sparse Initial State and Sparse Inputs

We now discuss the extension of the results presented thus far to the problem of jointly estimating the initial state as well as the input sequence, under sparsity constraints [68].

The system model in this case is as follows:

$$\boldsymbol{x}_k = \boldsymbol{D}\boldsymbol{x}_{k-1} + \boldsymbol{H}\boldsymbol{h}_k, \qquad (2.37)$$

$$\boldsymbol{y}_k = \boldsymbol{A}_{(k)} \boldsymbol{x}_k, \tag{2.38}$$

where  $\boldsymbol{H} \in \mathbb{R}^{n \times L}$  is the input matrix and  $\boldsymbol{h}_k \in \mathbb{R}^L$  is the input vector such that  $\|\boldsymbol{h}_k\|_0 \leq s_{in}$ . Therefore, the sparse recovery problem is given by the following equation:

$$\widetilde{\boldsymbol{y}}_{(K)} = \widetilde{\boldsymbol{A}}_{(K)} \boldsymbol{x}_0 + \widetilde{\boldsymbol{J}}_{(K)} \widetilde{\boldsymbol{h}}_{(K)}, \qquad (2.39)$$

where the measurement vector  $\tilde{\boldsymbol{y}}_{(K)} \in \mathbb{R}^{Km}$  (as defined in (2.5)), the unknown sparse vector  $\tilde{\boldsymbol{h}}_{(K)} \triangleq \begin{bmatrix} \boldsymbol{h}_1^\mathsf{T} & \dots & \boldsymbol{h}_{K-1}^\mathsf{T} \end{bmatrix}^\mathsf{T} \in \mathbb{R}^{(K-1)L}$  which is at most  $\tilde{s} = s + (K-1)s_{in}$  sparse, and the matrix  $\tilde{\boldsymbol{J}}_{(K)} \in \mathbb{R}^{Km \times (K-1)L}$  is defined follows:

$$\tilde{\boldsymbol{J}}_{(K)} = \begin{bmatrix} \boldsymbol{0} \in \mathbb{R}^{m \times (K-1)L} \\ \boldsymbol{A}_{(1)} \tilde{\boldsymbol{H}}_{(1)} \in \mathbb{R}^{m \times L} & \boldsymbol{0} \in \mathbb{R}^{m \times (K-2)L} \\ \boldsymbol{A}_{(2)} \tilde{\boldsymbol{H}}_{(2)} \in \mathbb{R}^{m \times 2L} & \boldsymbol{0} \in \mathbb{R}^{m \times (K-3)L} \\ \vdots \\ \boldsymbol{A}_{(K-1)} \tilde{\boldsymbol{H}}_{(K-1)} \in \mathbb{R}^{m \times (K-1)L} \end{bmatrix}$$
(2.40)  
$$\tilde{\boldsymbol{H}}_{(k)} = \begin{bmatrix} \boldsymbol{D}^{k-1} \boldsymbol{H} & \boldsymbol{D}^{k-2} \boldsymbol{H} & \dots & \boldsymbol{H} \end{bmatrix} \in \mathbb{R}^{N \times kL}.$$
(2.41)

Comparing (2.39) with (2.4), the effective measurement matrix of the recovery problem takes the form

$$egin{bmatrix} m{A}_{(0)}m{U}_{(0)} \ m{A}_{(1)}m{U}_{(1)} \ dots \ m{A}_{(1)}m{U}_{(1)} \ dots \ m{A}_{(K-1)}m{U}_{(K-1)}, \end{bmatrix},$$

where we define

$$\boldsymbol{U}_{(k)} \triangleq \begin{bmatrix} \boldsymbol{D}^{k} & \tilde{\boldsymbol{H}}_{(k)} & \boldsymbol{0}_{N \times (K-1-k)L} \end{bmatrix} \in \mathbb{R}^{N \times (N+(K-1)L)}.$$
(2.42)

To state results similar to Theorem 2.1 and Theorem 2.3, we define  $\tilde{\delta}_{s,\max}$  as the largest of the RICs among the matrices  $\{U_{(k)}\}_{k=0}^{K-1}$ . The proofs of the two theorems below are similar to that of the earlier results, and hence are omitted.

**Theorem 2.4** (Independent random observation matrices). Suppose the measurement matrices  $A_{(k)}, k = 0, 1, ..., K - 1$  are independent subgaussian random matrices with parameter c. Then, if

$$Km\left(\delta - 1 + (1 - \tilde{\delta}_{s,\max})^2\right)^2 \ge \tilde{c}\left[9s\ln\left(\frac{eN}{s}\right) + 2\ln(2\epsilon^{-1})\right],\tag{2.43}$$

the RIC  $\delta_s$  of a suitably scaled version  $\tilde{\boldsymbol{L}}(\boldsymbol{\delta}) \begin{bmatrix} \tilde{\boldsymbol{A}}_{(K)} & \tilde{\boldsymbol{J}}_{(K)} \end{bmatrix}$  of  $\begin{bmatrix} \tilde{\boldsymbol{A}}_{(K)} & \tilde{\boldsymbol{J}}_{(K)} \end{bmatrix}$  satisfies  $\delta_s < \delta$ for all  $1 - (1 - \tilde{\delta}_{s,\max})^2 < \delta < 1$  with probability at least  $1 - \epsilon$ . Here,  $\tilde{c}$  is a constant dependent only on c, and  $\boldsymbol{\delta}_j = 1 - \tilde{\delta}_{s,j}$  where  $\tilde{\delta}_{s,j}$  is the RIC of matrix  $\boldsymbol{U}_{(j)}$ . Hence, when (2.43) holds for  $s = \tilde{s}$ , the system is observable for sufficiently small  $\tilde{\delta}_{s,\max}$  with high probability.

Next, to state the corresponding result for the identical observation matrices case, we define quantities similar to (2.31) and (2.32) as follows:

$$\widetilde{\boldsymbol{U}}_{(K,i)} \triangleq \begin{bmatrix} \boldsymbol{U}_{(0)i} & \boldsymbol{U}_{(1)i} & \dots & \boldsymbol{U}_{(K-1)i} \end{bmatrix}, \qquad (2.44)$$

$$L_{\boldsymbol{U}}(\boldsymbol{D},\boldsymbol{H},\boldsymbol{K}) \triangleq \max_{i} \left\| \widetilde{\boldsymbol{U}}_{(\boldsymbol{K},i)} \right\|_{2}, \qquad (2.45)$$

where  $\boldsymbol{U}_{(k)i}$  denotes the  $i^{\mathsf{th}}$  column of  $\boldsymbol{U}_{(k)}$ .

**Theorem 2.5** (Identical random observation matrices). Suppose all observation matrices are identical, i.e.,  $\mathbf{A}_{(k)} = \mathbf{A}$  for k = 0, 1, ..., K-1, where  $\mathbf{A}$  is a subgaussian random matrix with parameter c. Then, if

$$Km \frac{\left(\delta - 1 + (1 - \tilde{\delta}_{s,\max})^{2(K-1)}\right)^2}{L_U^2(\boldsymbol{D}, \boldsymbol{H}, K)} \ge \tilde{c}s \max\left\{\ln^2 s \ln^2 N, \ln(2\epsilon^{-1})\right\},$$
(2.46)

then the RIC  $\delta_s$  of a suitably scaled version  $\tilde{\boldsymbol{L}}(\boldsymbol{\delta}) \begin{bmatrix} \tilde{\boldsymbol{A}}_{(K)} & \tilde{\boldsymbol{J}}_{(K)} \end{bmatrix}$  of  $\begin{bmatrix} \tilde{\boldsymbol{A}}_{(K)} & \tilde{\boldsymbol{J}}_{(K)} \end{bmatrix}$  satisfies  $\delta_s < \delta$  for all  $1 - (1 - \tilde{\delta}_{s,\max})^2 < \delta < 1$  with probability at least  $1 - \epsilon$ . Here,  $\tilde{c}$  is a constant dependent only on c, and  $\boldsymbol{\delta}_j = 1 - \tilde{\delta}_{s,j}$  where  $\tilde{\delta}_{s,j}$  is the RIC of matrix  $\boldsymbol{U}_{(j)}$ . Hence, when (2.46) holds for  $s = \tilde{s}$ , the system is observable for sufficiently small  $\tilde{\delta}_{s,\max}$  with high probability.

Remark 1: As before, we can extend the above results to the nonconsecutive measurements, noncanonical basis and robust recovery cases. Also, conditions in Theorem 2.4 and Theorem 2.5 can be made less stringent using the RIC of D. We omit explicitly stating the results to avoid repetition.

Remark 2: The above three theorems show how to extend three main results of the chapter (Theorem 2.1 and Theorem 2.3) to derive a sufficient condition for the structured random matrix in (2.40) to satisfy the RIP. These results could be of independent interest: they provide insight to the RIP of two special types of structured random matrices (resulting from independent and identical  $A_{(k)}$ ).

# 2.7 Summary

In this chapter, we derived the conditions for a linear dynamical system to be observable using the knowledge of its noiseless observations and inputs, when the initial state is sparse. We derived the results in the stochastic setting, both when the observation matrices are independent random matrices and when they are identical to a single random matrix. We characterized the number of measurements that are sufficient to observe the state of the linear dynamical system, using tools from compressed sensing. We also extended the results to the joint sparse input and initial state recovery. Thus, we completed the detailed theoretical analysis for the model with only initial state unknown. In the next chapter, we progress to the next level model in which both control inputs and the initial state is unknown (model SM2).

# Chapter 3

# **Sparse-Controllability**

### Answering problem Q1 for model version SM2

In this chapter, we consider at the LDS sparsity model with both initial state and sparse inputs unknown, and all system matrices known (model SM2). The unknown control inputs are assumed to sparse in a suitable basis. For this setting, we examine the conditions for controllability or the existence of a set of sparse control inputs that can drive the system from any arbitrary state to any other arbitrary final state. We note that unlike the last chapter, we do not assume that the initial state is sparse, and it can take any possible real vector value. To the best of our knowledge, we are the first ones to study the controllability of LDS for sparse inputs. This chapter covers the necessary and sufficient conditions for the controllability, upper and lower bounds on the length of the minimum number inputs that ensure controllability, a state space decomposition to separate sparse-controllable and sparse-uncontrollable spaces, and extensions to output controllability case.

# **3.1** Background

Networked control systems have attracted intense research attention from both academia and industry over the past decades [85–89]. In such a system, the notion of controllability refers to the ability to drive the system from an arbitrary initial state to a desired final state within a finite amount of time. Complete characterization of controllability of linear dynamical systems using unconstrained inputs have pure algebraic forms, and are rather easily verifiable [57, 90]. These conditions involve verification of the rank conditions of suitably defined matrices. However, in applications involving networked control systems, it is often necessary to select a small subset of the available sensors or actuators at each time instant, due to cost or energy constraints. For example, in an energy constrained network, energy-aware scheduling of actuators can help to extend the battery life of the nodes [91]. Similarly, in a system where the controller and plant communicate over a network, sparse control signals are required to meet the bandwidth constraints imposed by the links over which the control signals are exchanged [48, 49]. Now, when the number of actuators or input variables that can be activated is limited, the system may become uncontrollable because all the feasible control signals are restricted to lie in the union of low-dimensional subspaces. Thus, the controllability of linear dynamical systems under sparse input constraints is an important problem, and is the focus of this chapter.

### 3.1.1 Related Literature

We first discuss the relationship between the problem considered in this chapter and the existing literature in control theory and sparse signal porcessing.

#### Time-varying actuator scheduling problem

This problem focuses on the problem of finding a schedule for sparse actuator control, such that the system is sparse-controllable [88,89,91]. These works rely on a well known condition for controllability, namely, an extended version of the Kalman rank test. This test depends on the rank of the so-called Gramian matrix of the sparsity-constrained system.<sup>1</sup> However, finding sequence of control inputs that satisfy the rank condition on the Gramian matrix is a combinatorial problem, and it is known to be NP-hard [92, 93]. Moreover, it has been recently shown that the relatively simpler problem of finding a sparse set of actuators to guarantee reachability of a particular state is hard to approximate, even when a solution is known to exist [94]. Hence, different quantitative measures of controllability based on the Gramian matrix have been considered: smallest eigenvalue, the trace of the inverse, inverse of the trace, the determinant, maximum entry in the diagonal, etc. [91]. Based on these metrics, several algorithms and related guarantees are available in the literature [85,88,89]. However, none of the above mentioned references directly address the fundamental question of whether or not the system can be controlled by sparse inputs. Further, direct extension of the Kalman rank test leads to a combinatorial problem that is computationally infeasible to solve in practice. Thus, the goal of our chapter is to study the controllability of a linear dynamical system under sparsity constraints without directly relying on Gramian matrix. We are not interested in finding the optimal actuator selection; rather we deal with the more basic problem of deriving conditions for the existence of a selection that drives the system from any initial state to any final state.

<sup>&</sup>lt;sup>1</sup>Refer to [91, Section II.B] for details.

#### Minimal input selection problem

The minimal input selection involves selecting a small set of input variables so that the system is controllable using the selected set [92,93,95]. This problem is a special case of our sparse input problem because of the extra constraint that the support of the control input remains unchanged for all time instants. Moreover, the controllability conditions for the minimal input selection problem can be easily be derived from the classical controllability results for the unconstrained system. We discuss and contrast the two cases in detail in Section 3.3.3.

#### Design of sparse control inputs

Some recent works connecting compressive sensing and control theory focus on the design of control inputs [68, 96, 97]. They propose algorithms for the recovery (design) of sparse control inputs based on the observations, and derive conditions under which the input can be uniquely recovered using a limited number of observations [68, 96, 97]. These problems do not deal with controllability related issues, rather assume the existence of sparse control inputs and initial state for reaching a given final state.

### Observability under sparsity constraints

Due to the recent advances in sparse signal processing and compressed sensing, researchers have recently started looking at the observability of linear systems with a sparse initial state [50, 51, 62, 67]. For a system with unconstrained inputs, observability and controllability are dual problems and do not require separate analysis. However, our problem assumes a general initial state and sparse control inputs, whereas [50, 51, 62, 67] consider a sparse initial state and known control inputs. Therefore, the problems have different sparsity pattern models, and consequently require separate analysis.

#### Sparse signal recovery guarantees

The sparse controllability problem studies the conditions that ensure the existence of sparse control inputs to drive a linear system from any given state to any other state. Moreover, it is not required that the solution be unique. In contrast, the focus of traditional sparse signal processing studies is on developing algorithms and guarantees for the cases where the linear system is already known to admit a sparse solution [9–11,32]. Also, the structure of the effective measurement matrix that arises in the context of linear dynamical systems is different from the type of random measurement matrices that are usually considered in the compressed sensing literature.

The problem of controllability using sparse inputs is completely different in flavor compared to the existing work in control theory. Also, the solution to the problem is cannot be obtained using any of the available tools from sparse signal processing.

In the light of the discussion thus far, the primary questions that we address in this chapter are as follows:

- 1. What are necessary and sufficient conditions for ensuring controllability under sparse input constraints? Can we devise a simple method to test for controllability?
- 2. If a system is controllable using sparse inputs, what is the minimum number of control input vectors needed to drive the system from a given initial state to an arbitrary final state?
- 3. If the system is not controllable using sparse inputs, what parts of the state space are reachable using sparse inputs? In other words, how does one decompose the

state space into three subspaces: uncontrollable, uncontrollable using sparse inputs and controllable using sparse inputs?

Answering above questions requires a fresh look at controllability, and we start by deriving a Popov-Belevitch-Hautus (PBH)-like test [90]. Unlike the Gamian matrix based test discussed above, the new approach presented in this chapter allows one to check for sparsecontrollability of a system without solving a combinatorial problem. In a nutshell, this chapter studies theoretical aspects of the one of the most important notion in control theory: controllability under sparsity constraints on the input. We also note that the classical results for the unconstrained system can be recovered as a special case of our results, by relaxing the sparsity constraint.

## 3.2 System Model

We consider the discrete-time linear dynamical system

$$\boldsymbol{x}_k = \boldsymbol{D}\boldsymbol{x}_{k-1} + \boldsymbol{H}\boldsymbol{h}_k \tag{3.1}$$

where the transfer matrix  $\boldsymbol{D} \in \mathbb{R}^{N \times N}$  and input matrix  $\boldsymbol{H} \in \mathbb{R}^{N \times L}$ . Here, the input vectors are assumed to be sparse, i.e.,  $\|\boldsymbol{h}_k\|_0 \leq s$  for all k. We denote the rank of the matrices  $\boldsymbol{D}$  and  $\boldsymbol{H}$  using  $R_{\boldsymbol{D}}$  and  $R_{\boldsymbol{H}}$ , respectively.

We revisit the problem of controllability in the context of sparsity. We formally define the notion of controllability using sparse inputs as follows:

**Definition 3.1** (Sparse-controllability). The system defined by (3.1) is said to be s-sparsecontrollable if for any initial state  $\mathbf{x}_0$  and any final state  $\mathbf{x}_K$ , there exists an input sequence  $h_k, k = 1, 2, \cdots, K$  such that  $\|h_k\|_0 \leq s$ , which steers the system from the state  $x_0$  to  $x_K$  for some finite K.

Next, to characterize the sparse-controllability of the system, we consider the following equivalent system of equations:

$$\boldsymbol{x}_{K} - \boldsymbol{D}^{K} \boldsymbol{x}_{0} = \tilde{\boldsymbol{H}}_{(K)} \boldsymbol{h}_{(K)}, \qquad (3.2)$$

where we define the matrices as follows:

$$\tilde{\boldsymbol{H}}_{(K)} = \begin{bmatrix} \boldsymbol{D}^{K-1}\boldsymbol{H} & \boldsymbol{D}^{K-2}\boldsymbol{H} & \dots \boldsymbol{H} \end{bmatrix} \in \mathbb{R}^{N \times KL}$$
(3.3)

$$\boldsymbol{h}_{(K)} = \begin{bmatrix} \boldsymbol{h}_1^{\mathsf{T}} & \boldsymbol{h}_2^{\mathsf{T}} & \dots & \boldsymbol{h}_K^{\mathsf{T}} \end{bmatrix}^{\mathsf{T}} \in \mathbb{R}^{KL}.$$
(3.4)

Note that  $h_{(K)}$  is a *piecewise sparse vector*, i.e., it is formed by concatenating K sparse vectors, each with sparsity at most s.

# 3.3 Necessary and Sufficient Conditions for Sparse-Controllability

This section addresses question 1 in Section 3.1. It is well-known that the system is sparse-controllable if for some finite K, there exists a submatrix of  $\tilde{H}_{(K)}$  with rank N of the following form:

$$\begin{bmatrix} \boldsymbol{D}^{K-1}\boldsymbol{H}_{\mathcal{S}_1} & \boldsymbol{D}^{K-2}\boldsymbol{H}_{\mathcal{S}_2} & \dots & \boldsymbol{H}_{\mathcal{S}_K} \end{bmatrix} \in \mathbb{R}^{N \times Ks},$$

such that the index set  $S_i \subseteq \{1, 2, ..., L\}$  and  $|S_i| = s$ , for i = 1, 2, ..., K. In the sequel, we refer this condition to as the Kalman-type rank test. Note that the first (K-1)N columns

of  $\hat{H}_{(K)}$  belong to  $\mathcal{CS} \{D\}$ . Hence, to satisfy the Kalman-type rank test,  $\mathcal{S}_K$  should be such that  $\mathcal{CS} \{H_{\mathcal{S}_K}\}$  should contain the left null space of D. Thus, the Kalman-type rank test naturally leads to the necessary condition for sparse-controllability as the existence of an index set  $\mathcal{S}$  with s entries such that rank of the matrix  $\begin{bmatrix} D & H_{\mathcal{S}} \end{bmatrix} \in \mathbb{R}^{N \times (N+s)}$  is N. With this intuition in mind, we next show that the above condition is not only necessary but also sufficient for a controllable system to be s-sparse-controllable.

**Theorem 3.1.** The system given by (3.1) is s-sparse-controllable if and only if the following two conditions hold:

- 1. For all  $\lambda \in \mathbb{C}$ , rank of the matrix  $\begin{bmatrix} \lambda \mathbf{I} \mathbf{D} & \mathbf{H} \end{bmatrix} \in \mathbb{R}^{N \times (N+L)}$  is N.
- 2. There exists an index set  $S \subseteq \{1, 2, ..., L\}$  with s entries such that rank of  $\begin{bmatrix} D & H_S \end{bmatrix} \in \mathbb{R}^{N \times (N+s)}$  is N.

*Proof.* See Appendix B.1.

We make the following remarks from Theorem 3.1:

- From condition 2, if a system is *s*-sparse-controllable, then for all  $s \leq \tilde{s} \leq L$ , it is  $\tilde{s}$ -sparse-controllable. This is intuitive since every *s*-sparse vector is also  $\tilde{s}$ -sparse.
- From condition 2, the system is s-sparse-controllable only if

$$\min\left\{R_{\boldsymbol{H}}, s\right\} \ge N - R_{\boldsymbol{D}}.\tag{3.5}$$

This relation gives a necessary condition on the minimum sparsity s required to ensure the controllability using sparse inputs. Also, we note that for an unconstrained system, min  $\{R_{H}, s\} = R_{H}$ , and thus  $R_{H} + R_{D} \ge N$  is a necessary condition for controllability.

- For s = L, Theorem 3.1 reduces to the PBH test [90] since there is no constraint on the input. Similarly, when L = 1, the notion of controllability and sparsecontrollability are the same, and hence Theorem 3.1 reduces to the PBH test.
- If the system defined by the transfer matrix-input matrix pair (D, H<sub>S</sub>) is controllable for some index set S with s entries, the system is s−sparse-controllable. In particular, a controllable system with R<sub>H</sub> ≤ s is s−sparse-controllable.

Before we present the detailed implications of the theorem, we present some interesting corollaries of Theorem 3.1. The theorem assumes that the input vectors are sparse in the canonical basis. However, the result can be extended to the more general class of inputs that are sparse under a basis  $\Psi \in \mathbb{R}^{L \times L}$  other than the canonical basis to get the following corollary:

**Corollary 3.1.** The system given by (3.1) is controllable using inputs which are s-sparse under a basis  $\Psi \in \mathbb{R}^{L \times L}$  if and only if the following two conditions hold:

- 1. For all  $\lambda \in \mathbb{C}$ , rank of  $\begin{bmatrix} \lambda I D & H \end{bmatrix}$  is N.
- 2. There exists an index set  $S \subseteq \{1, 2, ..., L\}$  with s entries such that the rank of  $\begin{bmatrix} D & H\Psi_S \end{bmatrix}$  is N.

*Proof.* Since the input vector is sparse under the basis  $\Psi$ , the effective input matrix becomes  $H\Psi$ . Thus, replacing H with  $H\Psi$  in Theorem 3.1, we obtain a similar result for a non-canonical basis. Further, To obtain condition 1, we note that

$$\begin{bmatrix} \lambda \boldsymbol{I} - \boldsymbol{D} & \boldsymbol{H} \boldsymbol{\Psi} \end{bmatrix} = \begin{bmatrix} \lambda \boldsymbol{I} - \boldsymbol{D} & \boldsymbol{H} \end{bmatrix} \begin{bmatrix} \boldsymbol{I} & \boldsymbol{0} \\ \boldsymbol{0} & \boldsymbol{\Psi} \end{bmatrix}, \qquad (3.6)$$

and the matrix  $\begin{bmatrix} I & 0 \\ 0 & \Psi \end{bmatrix} \in \mathbb{R}^{N+L \times N+L}$  is invertible as  $\Psi$  is invertible. Therefore, rank of  $\begin{bmatrix} \lambda I - D & H\Psi \end{bmatrix}$  is N for all  $\lambda \in \mathbb{C}$  if and only if rank of  $\begin{bmatrix} \lambda I - D & H \end{bmatrix}$  is N for all  $\lambda \in \mathbb{C}$ , which gives condition 1.

Our next corollary gives a sufficient condition under which controllability without any input constraints is equivalent to sparse-controllability, for the system given by (3.1).

**Corollary 3.2.** A reversible system, i.e., a system with an invertible state transition matrix D, is s-sparse-controllable for any  $0 < s \leq L$  if and only if it is controllable.

*Proof.* See Appendix B.2.

In the following subsection, we bring out the usefulness of Theorem 3.1 by developing a simple procedure to check the controllability of a system under sparsity constraints.

### 3.3.1 Verification Procedure

We present the following procedure to verify the conditions of Theorem 3.1:

- 1. Compute the eigendecomposition of  $D^{\mathsf{T}}$ .
- 2. If  $s < N R_D$ , the system is not sparse-controllable.
- 3. Check condition 1 for  $\lambda$  set to be equal to each of the eigenvalues of  $D^{\mathsf{T}}$ . If the condition fails, the system is not sparse-controllable.
- 4. If none of the eigenvalues of  $D^{\mathsf{T}}$  is zero, the system is sparse-controllable.
- 5. Project the columns of  $\boldsymbol{H}$  onto the null space of  $\boldsymbol{D}^{\mathsf{T}}$  obtained from its eigendecomposition to get  $\boldsymbol{H}^{\perp} \in \mathbb{R}^{N \times L}$ . If rank of the  $\boldsymbol{H}^{\perp}$  is  $N - R_{\boldsymbol{D}}$ , the system is sparse-controllable.

Here, step 2 follows from necessary condition for sparse-controllability given by (3.5). Next, step 3 verifies condition 1 for all values of  $\lambda$ . This is because if there exists  $\boldsymbol{z} \in \mathbb{R}^N, \lambda \in \mathbb{C}$ such that  $\boldsymbol{z}^{\mathsf{T}} \begin{bmatrix} \lambda \boldsymbol{I} - \boldsymbol{D} & \boldsymbol{H} \end{bmatrix} = \boldsymbol{0}$ , then  $\boldsymbol{z}^{\mathsf{T}} \boldsymbol{D} = \lambda \boldsymbol{z}^{\mathsf{T}}$ . Thus, it suffices to verify condition 1 for at most N values of  $\lambda$ . The step 4 follows from Corollary 3.2 if  $\boldsymbol{D}$  has full column rank. Finally, in step 5, since the columns of  $\boldsymbol{H}^{\perp}$  are orthogonal to the columns of  $\boldsymbol{D}$ , we have

$$\mathsf{Rank}\left\{\boldsymbol{H}^{\perp}\right\} \le N - R_{\boldsymbol{D}} \le s. \tag{3.7}$$

Here, the last inequality follows from step 2. Further, we also have

$$\mathsf{Rank}\left\{\begin{bmatrix} \boldsymbol{D} & \boldsymbol{H}_{\mathcal{S}}\end{bmatrix}\right\} = \mathsf{Rank}\left\{\begin{bmatrix} \boldsymbol{D} & \boldsymbol{H}_{\mathcal{S}}^{\perp}\end{bmatrix}\right\}$$
(3.8)

$$= \operatorname{\mathsf{Rank}}\left\{\boldsymbol{D}\right\} + \operatorname{\mathsf{Rank}}\left\{\boldsymbol{H}_{\mathcal{S}}^{\perp}\right\}.$$
(3.9)

Therefore, an index S satisfies condition 2 if and only if  $\operatorname{Rank} \{ H_S^{\perp} \} = N - R_D$ . Moreover, (3.7) ensures that this condition is equivalent to  $\operatorname{Rank} \{ H^{\perp} \} = N - R_D$ , as verified by step 5.

The complexity of the procedure depends on the computations required for the eigendecomposition in step 1, N + 1 rank computations in steps 2 and 5, and the matrix multiplication in step 5 required for the projection operation using the obtained eigendecomposition. It is known that the complexity of finding the eigendecomposition of a matrix is  $\mathcal{O}(N^3)$  [98], and thus, the overall complexity of the procedure is polynomial in N. Therefore, Theorem 3.1 allows us to verify the controllability of any discrete system in polynomial time in N. It is interesting to note that the complexity is independent of the sparsity s. We reiterate that the above procedure ensures that there exists a set of s-sparse control inputs for every pair of initial and final states of the system. However, it does not reveal any insight on the support pattern of the input sequence. The determination of the support sequence is a completely different problem and is known to be NP-hard.

In the following two subsections, we discuss the several implications of Theorem 3.1 by relating it to the Kalman-type rank test and the minimal input selection problem.

### 3.3.2 Comparison with the Kalman-type Rank Test

The two conditions of Theorem 3.1 and the Kalman-type rank test described at the beginning of Section 3.3 are based on the two different characteristics of controllability, and provide insights into two aspects of the problem. The Kalman-type rank test identifies the range spaces of the possible controllability matrices for different sparsity patterns of the input. The union of these range spaces represents the set of all states that can be reached from zero-initial condition. This observation is immediate from (3.2) with  $\mathbf{x}_0 = \mathbf{0}$ . Therefore, the minimum number of input vectors required to satisfy the Kalman-type rank test characterizes the length of the input sequence, K, required to ensure controllability. We exploit this fact to characterize the minimum number of input vectors that ensures controllability in Section 3.4. Moreover, the Kalman-type rank test also identifies the support pattern of the input sequence that can drive the system from any given state to any other final state.

On the other hand, the conditions of Theorem 3.1 are based on recognizing the uncontrollable and sparse-uncontrollable parts of the system. Interestingly, these conditions are independent of the number of input vectors, K. The first condition is the same as the PBH test. Thus, it elegantly separates systems into three categories: *sparse-controllable, controllable but sparse-uncontrollable, and uncontrollable.* The system is *sparse-controllable* if both conditions 1 and 2 of the theorem are satisfied; *controllable but sparse-uncontrollable* if only condition 1 is satisfied; and *uncontrollable* if condition 1 is violated. We exploit this observation to design a recipe to decompose the system into sparse-controllable and sparse-uncontrollable parts in Section 3.5.

Next, we comment on the computational effort required to verify the two tests. To verify the Kalman-type rank test, we need to do at most  $\binom{L}{s}^{N}$  rank computations. However, as outlined in Section 3.3.1, Theorem 3.1 requires one to do at most N + 1 rank computations, one eigendecomposition and a matrix multiplication. Therefore, the computational cost required for Theorem 3.1 is polynomial in N and independent of s. In contrast, the computational complexity of the Kalman-type rank test grows exponentially with N and s. We also note that, since the Kalman-type rank test involves powers of D, numerical stability also needs to be considered. Overall, conditions of Theorem 3.1 are computationally easier to verify compared to the Kalman-type rank test.

### 3.3.3 Inputs with Common Support

We recall the minimal input selection problem discussed in Section 3.1. For such a problem, the system is controlled using sparse inputs with a common support, i.e., when the indices of the nonzero entries of all the inputs coincide. In this case, the effective system has the transfer matrix-input matrix pair as  $(D, H_S)$  for some index set S such that |S| = s. Hence, the controllability conditions are given as follows:

1. For some finite K, there exists a submatrix of  $\tilde{H}_{(K)}$  with rank N of the following form:

$$\begin{bmatrix} \boldsymbol{D}^{K-1}\boldsymbol{H}_{\mathcal{S}} & \boldsymbol{D}^{K-2}\boldsymbol{H}_{\mathcal{S}} & \dots & \boldsymbol{H}_{\mathcal{S}} \end{bmatrix} \in \mathbb{R}^{N \times Ks},$$

such that the index set  $S \subseteq \{1, 2, \dots, L\}$  and |S| = s.

2. For all  $\lambda \in \mathbb{C}$ , rank of the matrix  $\begin{bmatrix} \lambda I - D & H_{\mathcal{S}} \end{bmatrix} \in \mathbb{R}^{N \times (N+s)}$  is N, for some index set  $\mathcal{S} \subseteq \{1, 2, \dots, L\}$  such that  $|\mathcal{S}| = s$ .

We see that, due to the additional constraint of controllability using a common support, the above conditions are more stringent than those in Theorem 3.1. Thus, a system with sparse inputs offers greater flexibility and control, and incurs a similar communication  $\cos^2$  compared to a system restricted to using sparse inputs with common support.

Finally, we provide some illustrative numerical examples in the following subsection.

### 3.3.4 Illustrative Examples

We first give an example to demonstrate that a controllable system which does not satisfy condition 2 of Theorem 3.1 is not sparse-controllable.

**Example 3.1.** Consider a linear system with N = 3, L = 2,

$$\boldsymbol{D} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}, \text{ and } \boldsymbol{H} = \begin{bmatrix} 1 & 1 \\ 1 & 0 \\ 0 & 1 \end{bmatrix}.$$
 (3.10)

Using the PBH test, it is easy to see that the system is controllable. Also, using the procedure given Section 3.3.1, the system does not satisfy the conditions of Theorem 3.1.

We verify that the system is not 1-sparse-controllable using the initial state  $x_0 = 0$  and

<sup>&</sup>lt;sup>2</sup>The communication cost remains of order s, since the support can be conveyed using  $s \log(N)$  bits.

final state  $\boldsymbol{x}_f = \begin{bmatrix} 1 & 1 & 1 \end{bmatrix}^{\mathsf{T}}$ . From (3.2), we have,

$$\begin{bmatrix} 1\\1\\1 \end{bmatrix} = \sum_{k=1}^{K} \boldsymbol{D}^{K-k} \boldsymbol{H} \boldsymbol{h}_{k} = \begin{bmatrix} \sum_{k=1}^{K} \boldsymbol{h}_{k}[1] + \boldsymbol{h}_{k}[2] \\ \boldsymbol{h}_{K}[1] \\ \boldsymbol{h}_{K}[2] \end{bmatrix}.$$
 (3.11)

Since  $h_K$  is 1-sparse, the above system of equations does not have any solution, for any finite value of K. Thus, the system is not 1-sparse-controllable.

Next example illustrates the benefits of using sparse control in a linear system over the sparse control with common support discussed in Section 3.3.3.

**Example 3.2.** Consider a linear system with N = 3, L = 3,

$$\boldsymbol{D} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & -1 \end{bmatrix}, \text{ and } \boldsymbol{H} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 1 & 0 & 0 \end{bmatrix}.$$
 (3.12)

This system satisfies the conditions in Theorem 3.1 for s = 2, and is hence 2-sparsecontrollable. There are three possible unconstrained systems with input matrices of size  $3 \times 2$ :

$$\boldsymbol{H}_{(1)} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \\ 1 & 0 \end{bmatrix} \quad \boldsymbol{H}_{(2)} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{bmatrix} \quad \boldsymbol{H}_{(3)} = \begin{bmatrix} 0 & 0 \\ 0 & 1 \\ 1 & 0 \end{bmatrix}.$$

However, the three subsystems described by the matrix pair  $(\mathbf{D}, \mathbf{H}_{(k)})$  for k = 1, 2, 3 are individually uncontrollable. Hence, sparse control allows the system to be controllable without adding much communication burden.

Finally, we give an example of a system with non-invertible D which is both controllable and sparse-controllable. This example shows that the condition in Corollary 3.2 that D is invertible is not necessary, but sufficient for a controllable system to be sparse-controllable.

**Example 3.3.** Consider a linear system with N = 3, L = 2,

$$\boldsymbol{D} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{bmatrix}, \text{ and } \boldsymbol{H} = \begin{bmatrix} 1 & 1 \\ 1 & 0 \\ 1 & 1 \end{bmatrix}.$$
 (3.13)

We note that D is not an invertible matrix. Further, the system satisfies the conditions in Theorem 3.1 for s = 1, and hence it is 1-sparse-controllable.

In summary, in this section, we derived necessary and sufficient conditions that a system needs to satisfy to ensure sparse-controllability. We also presented a procedure with polynomial complexity to verify the conditions. Next, we address the question 2 in Section 3.1, and derive bounds on the number of sparse input vectors required to drive the system from a given state to any desired final state.

# 3.4 Minimum Number of Control Input Vectors

In this section, we upper and lower bound the minimum number of input vectors that are required to drive the system from any given state to any final state. Before we present the main result of the section, for comparison, we state the corresponding result for the unconstrained system. To state the result, we denote the degree of minimal polynomial of D using q.

**Theorem 3.2.** For a controllable system, the minimum number of input vectors K required to steer the system from any given state to any other state satisfies

$$N/R_{H} \le K \le \min\{q, N - R_{H} + 1\} \le N.$$
(3.14)

*Proof.* See [99, Section 6.2.1].

We note that when we restrict the admissible inputs to sparse vectors, the minimum number of input vectors required can possibly increase. The following theorem gives bounds on the number of sparse control inputs. Thereby, we get insights on the additional input vectors required to control a system when the admissible inputs are restricted to be sparse.

**Theorem 3.3.** For an s-sparse-controllable system, the minimum number of s-sparse input vectors  $K^*$  required to steer the system from any given state to any other state satisfies

$$\frac{N}{\min\left\{R_{\boldsymbol{H}},s\right\}} \leq K^* \leq \min\left\{q\left\lceil\frac{R_{\boldsymbol{H}}}{s}\right\rceil, N - R^*_{\boldsymbol{H},s} + 1\right\} \leq N,\tag{3.15}$$

where 
$$R_{H,s}^* = \max_{\substack{S \subseteq \{1,2,\dots,L\} \\ |S|=s}} \operatorname{Rank} \{H_S\}.$$
  
*Proof.* See Appendix B.3.

We can relax the above upper bound to get a simpler relation without  $R^*_{H,s}$  as follows.

**Corollary 3.3.** For an s-sparse-controllable system, the minimum number of input vectors  $K^*$  required to steer the system from any given state to any other state satisfies

$$\frac{N}{\min\left\{R_{\boldsymbol{H}},s\right\}} \le K^* \le \min\left\{q\left\lceil\frac{R_{\boldsymbol{H}}}{s}\right\rceil, R_{\boldsymbol{D}}+1, N\right\}.$$
(3.16)

*Proof.* The result follows from condition 2 of Theorem 3.1 which gives the following:

$$R_{H,s}^* \ge \max\{N - R_D, 1\}.$$
 (3.17)

We make the following observations from Theorem 3.3.

- *Invariance:* The bound is invariant under right or left multiplication of **H** by a non-singular matrix, and under any similarity transform on **D**.
- *Relation to s:* Both the upper and the lower bounds decrease with *s*. This is intuitive: as *s* increases, the system has more flexibility to choose its inputs, and thus it requires fewer number of input vectors to ensure controllability.
- Equivalence between Theorem 3.2 and Theorem 3.3: We consider three cases for comparison:
  - 1. When s = L, which corresponds to the unconstrained case, Theorem 3.3 reduces to Theorem 3.2, as expected.
  - 2. When  $s \ge R_H$ , Theorem 3.3 reduces to Theorem 3.2, as  $R^*_{H,s} = R_H$ . This follows because when  $s \ge R_H$ ,  $\mathcal{CS}\left\{\tilde{H}_{(K)}\right\}$  is the same as the column space of an  $N \times Ks$  submatrix of  $\tilde{H}_{(K)}$  with maximum rank.
  - 3. When min  $\{q, N \mathbf{R}_{H} + 1\} = N$ , the system requires the same number of control inputs to achieve controllability and *s*-sparse-controllability for any *s*. However, this is possible only if  $\mathbf{R}_{H} = 1$ , and any system with  $s \geq \mathbf{R}_{H}$  is equivalent to an unconstrained system, as discussed above.
- Equality: We note that the upper and lower bounds in Theorem 3.2 meet when  $N/R_{H} = N R_{H} + 1$ , which gives  $R_{H}$  as 1 or N. Similarly, for s = 1, the lower and upper bounds in Theorem 3.3 are equal, and  $K^* = N$ . Further, if  $R_{H} \ge s$ , we get  $R^*_{H,s} = s$ , and thus the bounds are equal when s = N.

# 3.5 Decomposing Sparse-controllable States

In this section, we consider question 3 in Section 3.1, and present a decomposition of the state space into sparse-controllable, sparse-uncontrollable and uncontrollable subspaces. We begin with the observation that s-sparse-controllability inherits the *invariance under a change of basis* property of the conventional controllability as discussed in the proposition below.

**Proposition 3.1** (Invariance under change of basis). The system defined by the matrix pair (D, H) is s-sparse-controllable if and only if the system defined by  $(U^{-1}DU, U^{-1}H)$ is s-sparse-controllable for every nonsingular  $U \in \mathbb{R}^{N \times N}$ .

*Proof.* We note that when D and H are replaced with  $U^{-1}DU$  and  $U^{-1}H$  respectively, in (3.3), we get  $U^{-1}\tilde{H}_{(K)}$  instead of  $\tilde{H}_{(K)}$ . Now, the result follows from the Kalman-type rank test and the fact that the rank of every submatrix of  $\tilde{H}_{(K)}$  and  $U^{-1}\tilde{H}_{(K)}$  are the same.

Inspired by the above proposition and in the same spirit as the Kalman decomposition [100], we transform the original system to an equivalent standard form using a change of basis, such that, the transformed state-space is separated into an s-sparsecontrollable subspace and an orthogonal s-sparse-uncontrollable subspace. As discussed in Section 3.3.2, the key idea of such a decomposition is as follows: if a system is not controllable, then it is not sparse-controllable. Therefore, the first step to decompose the system is to separate the controllable and uncontrollable states using the Kalman decomposition. Next, we identify the sparse-controllable part of the controllable part, for which we use Theorem 3.1. Recall that a controllable system is sparse-controllable if it satisfies condition 2 of Theorem 3.1. Hence, we need to find a basis for the controllable part such that the transformed state-space separates into two subsystems: one which satisfies condition 2 of Theorem 3.1, and the other which does not. The two bases together give the transform that decomposes the system to sparse-controllable and sparse-uncontrollable parts. We next describe the procedure for the decomposition followed by an explanation on why the procedure works.

- 1. Find a basis for  $\mathcal{CS}\left\{\tilde{\boldsymbol{H}}_{(N)}\right\}$  as  $\{\boldsymbol{u}_i\}_{i=1}^{R_1}$ , where  $R_1 \leq N$  is the rank of  $\tilde{\boldsymbol{H}}_{(N)}$ . Extend the basis by adding  $N - R_1$  linearly independent vectors  $\{\boldsymbol{u}_i\}_{i=R_1+1}^N$  to form a basis for  $\mathbb{R}^N$ . Define an invertible matrix  $\boldsymbol{U} \triangleq \begin{bmatrix} \boldsymbol{u}_1 & \boldsymbol{u}_2 \dots \boldsymbol{u}_N \end{bmatrix} \in \mathbb{R}^{N \times N}$ .
- 2. Compute  $\check{\boldsymbol{D}} = \boldsymbol{U}^{-1}\boldsymbol{D}\boldsymbol{U}$  and  $\check{\boldsymbol{H}} = \boldsymbol{U}^{-1}\boldsymbol{H}$  which take the following forms:

$$\check{\boldsymbol{D}} = \begin{bmatrix} \check{\boldsymbol{D}}_{(1)} \in \mathbb{R}^{R_1 \times R_1} & \check{\boldsymbol{D}}_{(2)} \in \mathbb{R}^{R_1 \times N - R_1} \\ \boldsymbol{0} \in \mathbb{R}^{N - R_1 \times R_1} & \check{\boldsymbol{D}}_{(3)} \in \mathbb{R}^{N - R_1 \times N - R_1} \end{bmatrix}$$
(3.18)

$$\check{\boldsymbol{H}} = \begin{bmatrix} \check{\boldsymbol{H}}_{(1)} \in \mathbb{R}^{R_1 \times L} \\ \boldsymbol{0} \in \mathbb{R}^{N-R_1 \times L} \end{bmatrix}.$$
(3.19)

3. Use the QR-decomposition to get  $\check{\boldsymbol{D}}_{(1)} = \check{\boldsymbol{U}}_{(1)} \tilde{\boldsymbol{D}}_{(1)}$ , where  $\check{\boldsymbol{U}}_{(1)} \in \mathbb{R}^{R_1 \times R_1}$  is an orthogonal matrix, and  $\tilde{\boldsymbol{D}}_{(1)} \in \mathbb{R}^{R_1 \times R_1}$  is an upper triangular matrix. The matrix  $\check{\boldsymbol{U}}_{(1)}$  takes the following form:

$$\check{\boldsymbol{U}}_{(1)} = \begin{bmatrix} \check{\boldsymbol{U}}_{(11)} \in \mathbb{R}^{R_1 \times R_2} & \check{\boldsymbol{U}}_{(12)} \in \mathbb{R}^{R_1 \times R_1 - R_2} \end{bmatrix},$$
(3.20)

where  $R_2$  is the rank of  $\boldsymbol{D}_{(11)}$ .

4. Let  $R_3 = \min\left\{s, \mathsf{Rank}\left\{\check{\boldsymbol{U}}_{(12)}^\mathsf{T}\check{\boldsymbol{H}}_{(1)}\right\}\right\}$ . Find a set of  $R_3$  independent rows of  $\check{\boldsymbol{U}}_{(12)}^\mathsf{T}\check{\boldsymbol{H}}_{(1)}$ ,

indexed by  $\mathcal{T}$ . Define  $\bar{U}_{(1)}$  by rearranging the columns of  $\check{U}_{(1)}$  as follows:

$$\bar{\boldsymbol{U}}_{(1)} \triangleq \begin{bmatrix} \check{\boldsymbol{U}}_{(11)} & \check{\boldsymbol{U}}_{(12)\mathcal{T}} & \check{\boldsymbol{U}}_{(12)\mathcal{T}^c} \end{bmatrix} \in \mathbb{R}^{R_1 \times R_1},$$
(3.21)

where the matrices  $\check{\boldsymbol{U}}_{(12)\mathcal{T}} \in \mathbb{R}^{R_3 \times L}$  and  $\check{\boldsymbol{U}}_{(12)\mathcal{T}^c} \in \mathbb{R}^{R_1 - R_2 - R_3 \times L}$  are the submatrices of  $\check{\boldsymbol{U}}_{(12)}$  with columns are indexed by  $\mathcal{T}$  and  $\mathcal{T}^c$ , respectively. Define an invertible matrix  $\bar{\boldsymbol{U}} \in \mathbb{R}^{N \times N}$  using some arbitrary invertible matrix  $\bar{\boldsymbol{U}}_{(2)} \in \mathbb{R}^{N - R_1 \times N - R_1}$  as follows:

$$\bar{\boldsymbol{U}} \triangleq \begin{bmatrix} \bar{\boldsymbol{U}}_{(1)} \in \mathbb{R}^{R_1 \times R_1} & \boldsymbol{0} \in \mathbb{R}^{R_1 \times N - R_1} \\ \boldsymbol{0} \in \mathbb{R}^{N - R_1 \times R_1} & \bar{\boldsymbol{U}}_{(2)} \in \mathbb{R}^{N - R_1 \times N - R_1} \end{bmatrix}.$$
(3.22)

5. Compute  $\bar{\boldsymbol{D}} = \bar{\boldsymbol{U}}^{-1}\check{\boldsymbol{D}}\bar{\boldsymbol{U}}$  and  $\bar{\boldsymbol{H}} = \bar{\boldsymbol{U}}^{-1}\check{\boldsymbol{H}}$ , which take the following forms:

$$\bar{\boldsymbol{D}} = \begin{bmatrix} \bar{\boldsymbol{D}}_{(1)} \in \mathbb{R}^{R_2 + R_3 \times R_2 + R_3} & \bar{\boldsymbol{D}}_{(2)} \\ \boldsymbol{0} \in \mathbb{R}^{N - R_2 - R_3 \times R_2 + R_3} & \bar{\boldsymbol{D}}_{(3)} \end{bmatrix}$$
(3.23)  
$$\bar{\boldsymbol{H}} = \begin{bmatrix} \bar{\boldsymbol{H}}_{(1)} \in \mathbb{R}^{R_2 + R_3 \times L} \\ \bar{\boldsymbol{H}}_{(2)} \in \mathbb{R}^{R_1 - R_2 - R_3 \times L} \\ \boldsymbol{0} \in \mathbb{R}^{N - R_1 \times L} \end{bmatrix}.$$
(3.24)

The  $(R_2+R_3)$ -dimensional part corresponding to  $(\bar{D}_{(1)}, \bar{H}_{(1)})$  is *s*-sparse-controllable, while the remaining part is *s*-sparse-uncontrollable. Also, since  $\bar{D} = (U\bar{U})^{-1} D (U\bar{U})$ and  $\bar{H} = (U\bar{U})^{-1} H$ , the new basis is  $U\bar{U}$ .

Here, in steps 1 and 2 are the same as the Kalman decomposition, and thus the  $R_1$ -dimensional part corresponding to  $(\check{\boldsymbol{D}}_{(1)}, \check{\boldsymbol{H}}_{(1)})$  is controllable, while the part corresponding to  $(\check{\boldsymbol{D}}_{(2)}, \boldsymbol{0})$ is uncontrollable. From the PBH test based conditions, we know that  $(\check{\boldsymbol{D}}_{(1)}, \check{\boldsymbol{H}}_{(1)})$  satisfies condition 1 of Theorem 3.1. Next, in steps 3 and step 4, we find a basis that separates the sparse-controllable part from the controllable part corresponding to  $(\check{\boldsymbol{D}}_{(1)}, \check{\boldsymbol{H}}_{(1)})$ , i.e., the part which satisfies condition 2 of Theorem 3.1. In step 4, since  $R_3 \leq \text{Rank} \left\{ \check{\boldsymbol{U}}_{(12)}^{\mathsf{T}} \check{\boldsymbol{H}}_{(1)} \right\}$ , we can always find  $R_3$  linearly independent rows of  $\check{\boldsymbol{U}}_{(12)}^{\mathsf{T}} \check{\boldsymbol{H}}_{(1)}$ . After step 4, we have

$$\begin{bmatrix} \check{\boldsymbol{U}}_{(1)}^{-1}\check{\boldsymbol{D}}_{(1)}\check{\boldsymbol{U}}_{(1)} & \check{\boldsymbol{U}}_{(1)}^{-1}\check{\boldsymbol{H}}_{(1)} \end{bmatrix} = \begin{bmatrix} \bar{\boldsymbol{D}}_{(11)} & \check{\boldsymbol{U}}_{(11)}^{\mathsf{T}}\check{\boldsymbol{H}}_{(1)} \in \mathbb{R}^{R_2 \times R_1} \\ \mathbf{0} & \check{\boldsymbol{U}}_{(12)\mathcal{T}}^{\mathsf{T}}\check{\boldsymbol{H}}_{(1)} \in \mathbb{R}^{R_3 \times R_1} \\ \mathbf{0} & \check{\boldsymbol{U}}_{(12)\mathcal{T}^c}^{\mathsf{T}}\check{\boldsymbol{H}}_{(1)} \in \mathbb{R}^{(R_1 - R_2 - R_3) \times R_1} \end{bmatrix}$$
(3.25)

since the rank of  $\check{\boldsymbol{D}}_{(1)}$  has rank  $R_2$ . The first  $R_2$  rows of the matrix are linearly independent, as  $\bar{\boldsymbol{D}}_{(11)}$  has full row-rank. Further, we note that  $\mathcal{T}$  is chosen such that it is the largest index set such that  $(\check{\boldsymbol{U}}_{(12)\mathcal{T}})^{\mathsf{T}}\check{\boldsymbol{H}}_{(1)}$  has a submatrix with *s* columns and has rank as  $R_3$ . Thus, we get the following:

$$\max_{\substack{\mathcal{S} \subseteq \{1,2,\dots,L\} \\ \|\mathcal{S}\| = s}} \operatorname{\mathsf{Rank}} \left\{ \begin{bmatrix} \check{\boldsymbol{U}}_{(1)}^{-1} \check{\boldsymbol{D}}_{(1)} \check{\boldsymbol{U}}_{(1)} & \check{\boldsymbol{U}}_{(1)}^{-1} \check{\boldsymbol{H}}_{(1)s} \end{bmatrix} \right\}$$

$$= \operatorname{\mathsf{Rank}} \left\{ \bar{\boldsymbol{D}}_{(11)} \right\} + \max_{\substack{\mathcal{S} \subseteq \{1,2,\dots,L\} \\ \|\mathcal{S}\| = s}} \operatorname{\mathsf{Rank}} \left\{ \check{\boldsymbol{U}}_{(12)}^{\mathsf{T}} \check{\boldsymbol{H}}_{(1)S} \right\}$$
(3.26)

$$= \mathsf{Rank}\left\{\bar{\boldsymbol{D}}_{(11)}\right\} + \min\left\{s, \mathsf{Rank}\left\{\check{\boldsymbol{U}}_{(12)}^{\mathsf{T}}\check{\boldsymbol{H}}_{(1)}\right\}\right\}$$
(3.27)

$$= \operatorname{\mathsf{Rank}}\left\{\bar{\boldsymbol{D}}_{(11)}\right\} + \operatorname{\mathsf{Rank}}\left\{\check{\boldsymbol{U}}_{(12)\mathcal{T}}^{\mathsf{T}}\check{\boldsymbol{H}}_{(1)}\right\}$$
(3.28)

$$= R_2 + R_3. (3.29)$$

Thus, condition 2 of Theorem 3.1 is satisfied by the reduced space of dimension  $R_2 + R_3 \leq R_1$ , and therefore, it is the sparse-controllable part. Also, note that, since  $R_3$  is nondecreasing in s, the dimension of the sparse-controllable part is also nondecreasing in s.

Finally, in step 5, we extend the basis  $\bar{U}_{(1)}$  obtained in step 4, to span  $\mathbb{R}^N$ . Overall,

the basis for the sparse-controllability decomposition is  $U\overline{U}$ , and the dimension of the sparse-controllable part of the system is  $R_2 + R_3 \leq R_1$ .

We illustrate the decomposition procedure with the following example.

**Example 3.4.** Consider a linear system with N = 4, L = 3, s = 1:

$$\boldsymbol{D} = \begin{bmatrix} 5.65 & 0 & -1.25 & -7.95 \\ 3.3 & 0 & -0.9 & -4.7 \\ -0.55 & 0 & 0.35 & 0.85 \\ 3.4 & 0 & -0.8 & -4.8 \end{bmatrix}$$
(3.30)  
$$\boldsymbol{H} = \begin{bmatrix} 0.25 & 1.25 & 1.5 \\ 0.25 & 1.25 & 1.5 \\ -0.5 & -0.75 & -1.25 \\ 0.25 & 1 & 1.25 \end{bmatrix}$$
(3.31)

Following the above procedure, from step 1

$$\boldsymbol{U} = \begin{bmatrix} 1 & 0 & 4 & 1 \\ 2 & -1 & 3 & 0 \\ -2 & 0 & -1 & 1 \\ 1 & 0 & 3 & 0 \end{bmatrix}.$$
 (3.32)

Step 2 gives the following with  $R_1 = 3$ :

$$\check{\boldsymbol{D}}_{(1)} = \begin{bmatrix} 0.2 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}, \\ \check{\boldsymbol{H}}_{(1)} = \begin{bmatrix} 0.25 & 0.25 & 0.5 \\ 0.25 & 0 & 0.25 \\ 0 & 0.25 & 0.25 \end{bmatrix}$$
(3.33)

In step 3, we get  $R_2 = 1$ , and

$$\check{\boldsymbol{U}}_{(11)} = \begin{bmatrix} 1\\0\\0 \end{bmatrix}, \quad \check{\boldsymbol{U}}_{(12)} = \begin{bmatrix} 0 & 0\\1 & 0\\0 & 1 \end{bmatrix}.$$
(3.34)

In step 4, we compute  $\check{\boldsymbol{U}}_{(12)}^{\mathsf{T}}\check{\boldsymbol{H}}_{(1)} = \begin{bmatrix} 0.25 & 0 & 0.25 \\ 0 & 0.25 & 0.25 \end{bmatrix}$ . Thus,  $R_3 = 1$ , and  $\mathcal{T} = \{1\}$  or  $\{2\}$  for s = 1. With  $\mathcal{T} = \{1\}$ , we get

$$\bar{\boldsymbol{U}} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, \quad \boldsymbol{U}\bar{\boldsymbol{U}} = \begin{bmatrix} 1 & 4 & 0 & 1 \\ 2 & 3 & -1 & 0 \\ -2 & -1 & 0 & 1 \\ 1 & 3 & 0 & 0 \end{bmatrix}.$$
 (3.35)

Finally, step 5 gives

$$\bar{\boldsymbol{D}}_{(1)} = \begin{bmatrix} 0.2 & 0\\ 0 & 0 \end{bmatrix}, \bar{\boldsymbol{H}}_{(1)} = \begin{bmatrix} 0.25 & 0.25 & 0.5\\ 0 & 0.25 & 0.25 \end{bmatrix},$$
(3.36)

which corresponds to the 1-sparse-controllable part of the system which has dimension  $R_2+R_3 = 2$ . It can be easily verified that the system defined using  $(\bar{D}_{(1)}, \bar{H}_{(1)})$  is 1-sparsecontrollable using Theorem 3.1.

# 3.6 Output Sparse-Controllability

The notion of controllability we discussed thus far has been concerned with state controllability: we analyzed the conditions for driving the system state from any initial state to any arbitrary final state using a finite number of sparse inputs. In this section, we extend our results to a variant of controllability called *output controllability*, which is related to transferring any initial state to any desired final output. Hence, we consider the following linear dynamical system:

$$\boldsymbol{x}_k = \boldsymbol{D}\boldsymbol{x}_{k-1} + \boldsymbol{H}\boldsymbol{h}_k, \tag{3.37}$$

$$\boldsymbol{y}_k = \boldsymbol{A}\boldsymbol{x}_k, \tag{3.38}$$

where the output matrix  $\mathbf{A} \in \mathbb{R}^{m \times N}$  with m < N, and the state evolution equation (3.37) is same as (3.1). Next, we define the notion of *output sparse-controllability* as follows:

**Definition 3.2** (Output sparse-controllability). The system defined by (3.37) and (3.38) is called output s-sparse-controllable if for any given initial state  $\mathbf{x}_0$  and any desired final output vector  $\mathbf{y}_K$ , there exists an input sequence  $\mathbf{h}_k$ ,  $k = 1, 2, \dots, K$  such that  $\|\mathbf{h}_k\|_0 \leq s$ , which steers the system from initial state  $\mathbf{x}_0$  to a final output  $\mathbf{y}_K$  for some finite K.

Now, to characterize the output sparse-controllability, we consider the following equivalent system of equations:

$$\boldsymbol{y}_{K} - \boldsymbol{A}\boldsymbol{D}^{K}\boldsymbol{x}_{0} = \boldsymbol{A}\tilde{\boldsymbol{H}}_{(K)}\boldsymbol{h}_{(K)}.$$
(3.39)

In the following subsections, we extend the results in the previous sections to the output sparse-controllability case.

# 3.6.1 Necessary and Sufficient Conditions for Output Sparse-Controllability

We begin by stating necessary and sufficient conditions for output controllability of an *unconstrained* system. In [101], a Kalman test for output controllability is derived, which states that an unconstrained system given by (3.37) and (3.38) is output controllable if and only if the matrix  $A\tilde{H}_{(K)}$  has full row rank for some finite K. However, a direct extension

of this result to the case of output sparse-controllability leads to a combinatorial test, which is computationally expensive. Hence, we first derive a PBH-type test for output (unconstrained) controllability, which we present as the following proposition.

**Proposition 3.2.** For an unconstrained system given by (3.37) and (3.38), the system is output controllable only if the rank of the matrix  $A \begin{bmatrix} \lambda I - D & H \end{bmatrix} \in \mathbb{R}^{m \times (N+L)}$  is m for all  $\lambda \in \mathbb{C}$ .

*Proof.* Our proof is by contradiction. Suppose that, for some  $\lambda \in \mathbb{C}$ , the matrix  $A \begin{bmatrix} \lambda I - D & H \end{bmatrix}$ does not have full row rank. Then, there exists a  $\mathbf{0} \neq \mathbf{z} \in \mathbb{C}^m$  such that

$$\boldsymbol{z}^{\mathsf{T}}\boldsymbol{A}\boldsymbol{D} = \lambda \boldsymbol{z}^{\mathsf{T}}\boldsymbol{A} \text{ and } \boldsymbol{z}^{\mathsf{T}}\boldsymbol{A}\boldsymbol{H} = \boldsymbol{0},$$
 (3.40)

which implies  $\boldsymbol{z}^{\mathsf{T}} \boldsymbol{A} \tilde{\boldsymbol{H}}_{(K)} = \boldsymbol{0}$  for all K. Hence, the Kalman test is violated, and thus the system is not output controllable, as required. 

We note that the PBH test for output controllability only gives us a necessary condition for output controllability. We illustrate this using the following example:

**Example 3.5.** Let m = 3, N = 5 and L = 3, and suppose the system given by (3.37) and (3.38) is defined by the following matrices:

$$\boldsymbol{D} = \begin{bmatrix} 1 & 2 & 4 & 5 & 9 \\ 7 & 2 & 3 & 1 & 7 \\ 0 & 0 & 1 & 2 & 5 \\ 0 & 0 & 3 & 4 & 7 \\ 0 & 0 & 1 & 6 & 9 \end{bmatrix}, \quad \boldsymbol{H} = \begin{bmatrix} 1 \\ 2 \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix}, and \quad \boldsymbol{A} = \begin{bmatrix} 0 & 0.019 & -0.14 & 0.02 & 0.99 \\ 0 & -0.08 & 0.24 & 0.97 & 0.018 \\ 1 & 0 & 0 & 0 & 0 \end{bmatrix}.$$

(3.41)

It can be verified that the system fails the Kalman test, as  $\operatorname{Rank}\left\{A\tilde{H}_{(K)}\right\} < m$  for all K. However, for all values of  $\lambda \in \mathbb{C}$ ,  $\operatorname{Rank}\left\{A\begin{bmatrix}\lambda I - D & H\end{bmatrix}\right\} = 3 = m$ . Thus, the condition of the PBH test is necessary but not sufficient for output controllability.

Before we present the main result, we note that the Kalman-type test for output (unconstrained) controllability [101] immediately extends to output sparse-controllability as follows. The system is output controllable if and only if, for some finite K, there exists a submatrix of  $\tilde{AH}_{(K)}$  with rank m of the following form:

$$oldsymbol{A} egin{bmatrix} oldsymbol{D}^{K-1}oldsymbol{H}_{\mathcal{S}_1} & oldsymbol{D}^{K-2}oldsymbol{H}_{\mathcal{S}_2} & \dots & oldsymbol{H}_{\mathcal{S}_K} \end{bmatrix} \in \mathbb{R}^{m imes Ks},$$

such that the index set  $S_i \subseteq \{1, 2, ..., L\}$  and  $|S_i| = s$ , for i = 1, 2, ..., K. Next, we extend Theorem 3.1 to the output sparse-controllability, as follows:

**Theorem 3.4.** The system given by (3.37) and (3.38) is output s-sparse-controllable only if the following conditions are satisfied:

- 1. For all  $\lambda \in \mathbb{C}$ , rank of  $A \begin{bmatrix} \lambda I D & H \end{bmatrix} \in \mathbb{R}^{m \times (N+L)}$  is m.
- 2. There exists an index set  $S \subseteq \{1, 2, ..., L\}$  with s entries such that rank of  $A\begin{bmatrix} D & H_S \end{bmatrix} \in \mathbb{R}^{m \times (N+s)}$  is m.

*Proof.* The result can be proved by using an approach similar to the proof of Theorem 3.1 given in Appendix B.1. We replace z in the third part of the proof with Az to show the necessity of the above conditions.

Theorem 3.4 is the same as Theorem 3.1, except for a pre-multiplication with  $\boldsymbol{A}$ . We make the following observations:

- When A = I, Theorem 3.4 reduces to Theorem 3.1, as expected.
- We know that Rank {AH\*} ≤ Rank {A}, for any matrix H\*. Thus, if Rank {A} < m, the Kalman test for output sparse-controllability fails. Hence, the system is not output sparse-controllable.</li>
- Suppose Rank  $\{A\} = m$  for an *s*-sparse-controllable system. Invoking Sylvester's rank inequality [102], we get

$$m = \operatorname{\mathsf{Rank}} \{\mathbf{A}\} + \operatorname{\mathsf{Rank}} \{\mathbf{H}^*\} - N \le \operatorname{\mathsf{Rank}} \{\mathbf{A}\mathbf{H}^*\} \le \operatorname{\mathsf{Rank}} \{\mathbf{A}\} = m,$$
 (3.42)

where  $\boldsymbol{H}^* \in \mathbb{R}^{N \times Ks}$  is the submatrix of  $\tilde{\boldsymbol{H}}_{(K)}$  that satisfies the Kalman test for state sparse-controllability, for some finite K. Hence, the system is output s-sparsecontrollable. Therefore, the conditions in Theorem 3.4 are less restrictive than those in Theorem 3.1, as the output dimension m < N, provided  $\boldsymbol{A}$  has rank m.

From the last observation, we see that it is possible that the system is output s-sparsecontrollable, even if it is *not* s-sparse-controllable, provided Rank  $\{A\} = m$ . We illustrate this using the following example.

**Example 3.6.** Let m = 2, N = 3 and L = 2, and suppose the system given by (3.37) and (3.38) is defined by the following matrices:

$$\boldsymbol{D} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}, \quad \boldsymbol{H} = \begin{bmatrix} 1 & 1 \\ 1 & 0 \\ 0 & 1 \end{bmatrix} \text{ and } \boldsymbol{A} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix}.$$
(3.43)

It can be verified that the system is not 1-sparse-controllable, but the system is output 1-sparse-controllable.

From Theorem 3.4, we can derive a procedure similar to the one given in Section 3.3.1, to verify output sparse-controllability of a system. We omit it to avoid repetition.

# 3.6.2 Minimum Number of Input Vectors for Output Controllability

A bound on the smallest number of s-sparse input vectors that ensures output controllability is given the following theorem:

**Theorem 3.5.** For an output s-sparse-controllable system, the minimum number of input vectors  $K^*$  required to steer any initial output to any final output satisfies

$$\frac{m}{\min\left\{\mathsf{Rank}\left\{\boldsymbol{AH}\right\},s\right\}} \le K^* \le \min\left\{q\left[\frac{R_{\boldsymbol{H}}}{s}\right], m - R^*_{\boldsymbol{AH},s} + 1\right\} \le m,$$
(3.44)

where  $R_{AH,s}^* = \max_{\substack{\mathcal{S} \subseteq \{1,2,\dots,L\} \\ |\mathcal{S}|=s}} \operatorname{Rank} \{AH_{\mathcal{S}}\}, R_H$  is the rank of H, and q is the degree of the minimal polynomial of D.

*Proof.* The proof is along similar lines as the proof in Appendix B.1. We start by premultiplying the right-hand side of (B.6) by  $\boldsymbol{A}$  to get a modified definition of submatrices of  $\boldsymbol{A}\tilde{\boldsymbol{H}}$ , and then follow the same arguments as in Appendix B.1 to get the desired result.  $\Box$ 

It is interesting to note that the bounds in Theorem 3.5 are smaller than those in Theorem 3.3. This is because the dimension of the output space, m, is smaller than that of the state space, N. Also, the above result depends only on the degree of the minimal polynomial of D, and not on the degree of the minimal polynomial of AD.

A relaxed version of the above result, similar to Corollary 3.3, is as follows:

**Corollary 3.4.** For an s-sparse-controllable system, the minimum number of input vectors  $K^*$  required to steer any initial output to any other final output satisfies

$$\frac{m}{\min\left\{\mathsf{Rank}\left\{\mathbf{AH}\right\},s\right\}} \le K^* \le \min\left\{q\left[\frac{R_H}{s}\right],\mathsf{Rank}\left\{\mathbf{AD}\right\}+1,m\right\}.$$
(3.45)

Theorem 3.5 provides bounds on the minimum number of input vectors to ensure output controllability without any constraints on the input, by substituting s = L. We get the following result:

**Corollary 3.5.** For an output controllable system, the minimum number of input vectors K required to steer any initial output to any final output satisfies

$$\frac{m}{\operatorname{\mathsf{Rank}}\left\{\boldsymbol{AH}\right\}} \le K \le \min\left\{q, m - \operatorname{\mathsf{Rank}}\left\{\boldsymbol{AH}\right\} + 1\right\} \le m.$$
(3.46)

### 3.6.3 Change of Basis Property

Output sparse-controllability possesses invariance under a change of basis, similar to state sparse-controllability.

**Proposition 3.3** (Invariance under change of basis). The linear system defined by the matrix tuple (D, H, A) is output s-sparse-controllable if and only if the system defined by  $(U^{-1}DU, U^{-1}H, AU)$  is output s-sparse-controllable for every nonsingular matrix  $U \in \mathbb{R}^{N \times N}$ .

*Proof.* We note that when D and H are replaced with  $U^{-1}DU$  and  $U^{-1}H$  respectively, in (3.3), we get  $U^{-1}\tilde{H}_{(K)}$  instead of  $\tilde{H}_{(K)}$ . Further, replacing A with AU does not change in rank criterion in the Kalman-type rank test for output sparse controllability.

Unlike state controllability, the change of basis does not change the equivalent linear

system of equations given by (3.39). However, designing a procedure for decomposing the output space into sparse-controllable and sparse-uncontrollable subspaces similar to the one proposed for sparse-controllability is non-trivial, and we defer it to future work.

# 3.7 Summary

This chapter presented the necessary and sufficient conditions for controllability of linear systems subject to sparsity constraints on the input. We derived two easily verifiable conditions equivalent the Kalman-type rank test for the sparse-controllability of the system. We also presented a simple procedure with polynomial complexity to verify the conditions of the theorem. Further, we bounded the minimum number of input vectors that ensures controllability. In addition, the sparse-controllability tests also led to a Kalman decomposition-like procedure for decomposing the system into sparse-controllable, controllable but sparse-uncontrollable and uncontrollable parts. Thus, we have addressed three important aspects of controllability to the output controllability. The chapter dealt with the first question Q1 on the existence of a sparse solution for the second level of sparsity model SM2. Next, we proceed to the next question Q2 in Chapter 1 for the same sparsity model.

# Chapter 4

# Bayesian Recovery Algorithms for Jointly Sparse Control Inputs

Answering problem Q2 for model version SM2

In this chapter, we present recovery algorithms for the recovery of jointly sparse control inputs, i.e., we consider the special case when D is a diagonal matrix. Also, we are interested in designing a sequential (online) algorithm with low complexity. These type of algorithms are particularly useful in case of applications like wireless channel estimation. The measurements are processed one after another in a sequential fashion, without waiting for the complete input to arrive. Such algorithms require significantly lower computational and memory resources compared to their offline counterparts. Also, estimates of the sparse vectors become available after a fixed delay from the time observations arrive.

# 4.1 Background

In many applications, such as wireless channel tracking [3], radar signal processing [103, 104], and biomedical imaging [105–108], the goal is to recover a sequence of sparse vectors

that exhibit additional structure, such as a common support and temporal correlation. For example, successive instantiations of a time-varying wireless channel have the same power delay profile, and the nonzero coefficients of these instantiations are temporally correlated, and can be modeled using a first-order auto-regressive (AR) process. Exploiting this additional structure in the multiple measurement vectors (MMV) can improve the recovery performance, but at the cost of higher latency, memory, and computational complexity. Hence, the goal of this chapter is to develop algorithms that exploit the structure in the signal to reconstruct a sequence of sparse vectors. However, exploiting the additional structure can lead to higher latency, memory, and computational complexity. Therefore, we are particularly interested in developing algorithms with low complexity and bounded latency.

There are many offline recovery algorithm proposed in lietratue, that exploits the temporal correlation across the sparse vectors [3, 42, 43]. The aforementioned algorithms are offline in nature, i.e., they process the entire set of measurement vectors in a single batch. Hence, when the data set is large, these algorithms suffer from poor efficiency and scalability. On the other hand, *online* algorithms process small batches of the measurement vectors at a time and recover the sparse vectors sequentially, resulting in low-complexity implementations. Online algorithms offer the additional benefit of low latency between the measurement and estimation, which may be necessary in certain applications. For example, in a real-time broadband communication system with high data rate and high mobility, offline estimation of the wireless channel is infeasible.

Several sequential algorithms for sparse signal recovery have been presented in the literature [109–114]. An online algorithm for recovery for sparse signal with common support is

81

presented in [109]. However, the algorithm does not account for the temporal correlation in the signal. A non-iterative modified OMP algorithm for sequential recovery of sparse signals is described in [110] for the case when the coefficient in the autoregression is unity. A combination of Kalman filtering and dynamic programming is givem in [111]. This algorithm is slow because it runs  $l_1$  optimization multiple times for every measurement vector. Another iterative sequential algorithm that decouples the support recovery step from the Kalman filtering-based amplitude estimation step is presented in [112]. However, the algorithm requires one to tune a number of parameters beforehand. An alternate iterative online algorithm that jointly estimates the amplitude and support is hierarchical Bayesian Kalman filtering [113]. This algorithm does not require one to tune many parameters, but suffers from high complexity. Another algorithm for the sequential recovery of sparse signals is dynamic sparse coding [114]. The algorithm executes an optimization procedure based on gradient descent, and is also iterative in nature.

The above discussed algorithms do not allow one to improve the current estimate using a small set of future measurements. For scenarios that often arise in communication related applications (e.g., wireless channel estimation), a small delay is allowed if the estimation performance can be improved. Therefore, we need to use a smoothing operation instead of a filtering operation, and then filtering becomes a special case of smoothing when the allowed delay is zero. We present two algorithms in this chapter: *iterative online* algorithm and *non-iterative online* algorithm. The iterative algorithm allows a bounded delay between the measurement and estimation by combining the Kalman smoothing and the SBL framework. The algorithm runs multiple rounds of the expectation-maximization (EM) procedure for every measurement vector. Next, we improve this algorithm to obtain a

non-iterative algorithm which has simpler implementation with minimal resource requirements. The non-iterative online algorithm where as every measurement vector arrives, we do not run an iterative procedure until convergence of some metric. The algorithm does one round of update using the measurement vector, and waits for the next measurement vector. We reiterte that both these algorithms do not require parameter tuning and allows a small delay between the measurement and estimation, for the reconstruction of temporally correlated sparse vectors with common support.

Our online algorithms are based on the SBL framework [22,41]. The SBL approach offers superior performance compared to other algorithms like  $l_1$  minimization and OMP, and does not require one to tune the algorithm parameters. Moreover, it naturally extends to incorporate the temporal correlation structure in the signal model. However, its complexity and memory requirements increase with the number of measurements to be processed, which limits its practical application. Our algorithms overcome this drawback, and is computationally efficient, while retaining the good performance of SBL.

# 4.2 **Problem Formulation**

We consider a special case of LDS presented in Chapter 1, where  $\boldsymbol{x}_0 \triangleq \boldsymbol{0}$  and  $\boldsymbol{D} \in [0, 1)^{N \times N}$ and  $\boldsymbol{H} \in \mathbb{R}^{N \times L}$  are the known diagonal matrices. The system model is given by

$$\boldsymbol{x}_k = \boldsymbol{D}\boldsymbol{x}_{k-1} + \boldsymbol{H}\boldsymbol{h}_k \tag{4.1}$$

$$\boldsymbol{y}_k = \boldsymbol{A}_k \boldsymbol{x}_k + \boldsymbol{w}_k, k = 1, 2, \dots$$
(4.2)

Here,  $\boldsymbol{w}_k$  is a zero mean Gaussian distributed noise with a full rank covariance matrix  $\boldsymbol{R}_k$ . The number of measurements m is assumed to be smaller than the number of unknowns N which makes the system underdetermined. The unknown sequence of vectors  $\{\boldsymbol{h}_k, k = 1, 2, \ldots\}$  are sparse, i.e., the number of nonzero entries, S, is small compared to the size of the vector, N. The  $\boldsymbol{h}_k$  are simultaneously sparse, that is, they share a common support. This implies that the indices of the nonzero entries of all the sparse vectors coincide. Note that, in our model, the sparse vectors are temporally correlated, but because  $\boldsymbol{D}$  and  $\boldsymbol{H}$  are both assumed to be diagonal, there is no intra-vector correlation. Also, the support of  $\boldsymbol{x}_k$  coincides with that of  $\{\boldsymbol{h}_k\}_{k \in \mathbb{N}}$ .

### 4.2.1 Estimation Objectives

The objective of this work is to estimate the sparse vectors on-the-fly, without storing all the measurement data and the corresponding measurement matrices. The maximum delay allowed between the measurement and estimation is  $\Delta < \infty$ , and therefore our goal is to recursively estimate  $\boldsymbol{x}_k$  using the measurements up to time  $k + \Delta$ , denoted by  $\boldsymbol{y}^{k+\Delta}$ . Throughout the chapter, we use subscripts to denote the value of a variable at a particular time instant (e.g.,  $\boldsymbol{y}_k$  denotes the observation at time k), and superscripts to denote the sequence of observations up to a particular time instant (e.g.,  $\boldsymbol{y}^{\ell}$  denotes the sequence of observations { $\boldsymbol{y}_k, k = 1, 2, ..., \ell$ }).

We design an online scheme inspired by the SBL algorithm [22], [41]. The extension of SBL for the recovery of simultaneous sparse vectors imposes a common prior on the unknown vectors, namely,  $\boldsymbol{x}_k \sim \mathcal{N}(\boldsymbol{0}, \boldsymbol{\Gamma})$  [41]. The covariance matrix  $\boldsymbol{\Gamma} \in \mathbb{R}^{N \times N}_+$  is a diagonal matrix with N hyperparameters  $\boldsymbol{\gamma} \in \mathbb{R}^N_+$  along the diagonal. In SBL, we compute the ML estimate  $\boldsymbol{\gamma}_{\text{ML}}$  of  $\boldsymbol{\gamma}$ , which in turn gives the MAP estimate of the sparse vectors. In the following subsections, we contrast the offline and online approaches to estimating the hyperparameters and sparse vectors, which serves to bring out the primary estimation objectives of this work. We start with the online case.

#### Online

Let  $\gamma^{k-1}$  denote the sequence of estimates of the hyperparameters  $\gamma$  till time k - 1. At time k, we want to compute the estimate of the hyperparameter vector  $\gamma_k$ , using  $y^{k+\Delta}$  and  $\gamma^{k-1}$ . Since we do not want to store the complete set of past measurements, we recursively update  $\gamma_k$  using a small set of measurements  $\{y_t, t = k, k+1, \ldots, k+\Delta\}$  and  $\gamma_{k-1}$ . The update rule for  $\gamma_k$  is discussed in Section 4.4.

Using  $\gamma_k$ , the online estimate of  $\boldsymbol{x}_k$  is computed as its conditional mean given  $\boldsymbol{y}^{k+\Delta}$ , with  $\Gamma_t$  as the covariance of  $\boldsymbol{x}_t$  for t = 1, 2, ..., k - 1, and  $\Gamma_k$  as the covariance of  $\boldsymbol{x}_t$  for  $t = k, k + 1, ..., k + \Delta$ . Mathematically,

$$\hat{\boldsymbol{x}}_{k} = \mathbb{E}\left\{\boldsymbol{x}_{k} | \boldsymbol{y}^{k+\Delta}; \boldsymbol{\gamma}^{k-1}, \boldsymbol{\gamma}_{k}\right\}.$$
(4.3)

The estimate  $\hat{x}_k$  is obtained using fixed interval Kalman smoothing on a data block of size  $\Delta + 1$  [2]. That is,  $x_k$  is recursively updated using the set of measurement vectors  $\{y_t, t = k, k+1, \dots, k+\Delta\}$  and  $\gamma_k$ . Note that  $\gamma^{k-1}$  is not used in the estimation of  $x_k$ .

We emphasize that, with the estimate of  $\gamma_k$  in hand, the estimation of  $x_k$  is a straightforward application of the Kalman filtering principle. The key contribution of this chapter is the development of a recursive, online technique for estimating  $\gamma_k$  and its convergence analysis. We next discuss the offline case.

#### Offline

In the offline setting, we find the ML estimate  $\gamma^{\text{OFF}}$  of  $\gamma$  given the entire sequence  $\boldsymbol{y}^{K}$ , where K denotes the total number of measurements [3,41]. The estimation procedure is detailed in Section 4.2.2. The estimate of  $\boldsymbol{x}_{k}$  is computed as its conditional mean given  $\boldsymbol{y}^{K}$ , using Diag { $\gamma^{\text{OFF}}$ } as the signal covariance matrix. Mathematically,

$$\hat{\boldsymbol{x}}_{k}^{\text{OFF}} = \mathbb{E}\left\{\boldsymbol{x}_{k} | \boldsymbol{y}^{K}; \boldsymbol{\gamma}^{\text{OFF}}\right\}, \qquad (4.4)$$

for k = 1, 2, ..., K. These estimates are computed efficiently using fixed interval Kalman smoothing on the data block  $\boldsymbol{y}^{K}$ .

Thus, the primary goal in both the offline and online algorithms is the estimation of  $\gamma$ . In the offline case, a single estimate of  $\gamma$  is computed using the entire set of observations. In the online version, a sequence of estimates are computed using small batches of observations, and in a recursive manner.

In the next subsection, we first describe the offline SBL algorithm for the correlated MMV problem, which we refer to as the *offline Kalman MMV SBL (KM-SBL)* algorithm [3].

### 4.2.2 Offline KM-SBL Algorithm

The offline algorithm uses the expectation-maximization (EM) procedure, which treats the unknowns  $\boldsymbol{x}^{K}$  as the hidden data and the observations  $\boldsymbol{y}^{K}$  as the known data. The EM procedure iterates between two steps: an expectation step (E-step) and a maximization step (M-step). Let  $\boldsymbol{\gamma}^{(r-1)}$  be the estimate of  $\boldsymbol{\gamma}$  at the  $r^{\text{th}}$  iteration.<sup>1</sup> The E-step computes  $Q(\boldsymbol{\gamma}, \boldsymbol{\gamma}^{(r-1)})$ , which is the marginal log-likelihood of the observed data. The M-step

<sup>&</sup>lt;sup>1</sup>For ease of notation, we omit the superscript OFF here.

computes the hyperparameters that maximize  $Q\left(\boldsymbol{\gamma}, \boldsymbol{\gamma}^{(r-1)}\right)$ .

E-step: 
$$Q(\boldsymbol{\gamma}, \boldsymbol{\gamma}^{(r-1)}) = \mathbb{E}_{\boldsymbol{x}^{K}|\boldsymbol{y}^{K};\boldsymbol{\gamma}^{(r-1)}} \{\log p(\boldsymbol{y}^{K}, \boldsymbol{x}^{K}; \boldsymbol{\gamma})\}$$
  
M-step:  $\boldsymbol{\gamma}^{(r)} = \operatorname*{arg\,max}_{\boldsymbol{\gamma} \in \mathbb{R}^{N \times 1}_{+}} Q(\boldsymbol{\gamma}, \boldsymbol{\gamma}^{(r-1)}).$ 
(4.5)

Simplifying  $Q(\boldsymbol{\gamma}, \boldsymbol{\gamma}^{(r-1)})$  we get,

$$Q(\boldsymbol{\gamma}, \boldsymbol{\gamma}^{(r-1)}) = c_{K} - \frac{K}{2} \log |\boldsymbol{\Gamma}| - \frac{1}{2} \operatorname{Tr} \left\{ \boldsymbol{\Gamma}^{-1} \boldsymbol{C}_{1|K, \boldsymbol{\gamma}^{(r-1)}} \right\} - \frac{1}{2} \sum_{t=2}^{K} \operatorname{Tr} \left\{ \boldsymbol{\Gamma}^{-1} \left( \boldsymbol{I} - \boldsymbol{D}^{2} \right)^{-1} \boldsymbol{T}_{t|K, \boldsymbol{\gamma}^{(r-1)}} \right\}.$$
(4.6)

where the constant  $c_K$  is independent of  $\gamma$ , and the  $N \times N$  matrices are defined as follows:

$$\boldsymbol{T}_{t|K,\boldsymbol{\gamma}^{(r-1)}} \triangleq \boldsymbol{C}_{t|K,\boldsymbol{\gamma}^{(r-1)}} + \boldsymbol{D}\boldsymbol{C}_{t-1|K,\boldsymbol{\gamma}^{(r-1)}}\boldsymbol{D} - 2\boldsymbol{D}\boldsymbol{C}_{t,t-1|K,\boldsymbol{\gamma}^{(r-1)}}$$
(4.7)

$$\boldsymbol{C}_{t|K,\boldsymbol{\gamma}^{(r-1)}} \triangleq \boldsymbol{P}_{t|K,\boldsymbol{\gamma}^{(r-1)}} + \widehat{\boldsymbol{x}}_{t|K,\boldsymbol{\gamma}^{(r-1)}} \widehat{\boldsymbol{x}}_{t|K,\boldsymbol{\gamma}^{(r-1)}}^{\mathsf{T}}$$
(4.8)

$$\boldsymbol{C}_{t,t-1|K,\boldsymbol{\gamma}^{(r-1)}} \triangleq \boldsymbol{P}_{t,t-1|K,\boldsymbol{\gamma}^{(r-1)}} + \widehat{\boldsymbol{x}}_{t|K,\boldsymbol{\gamma}^{(r-1)}} \widehat{\boldsymbol{x}}_{t-1|K,\boldsymbol{\gamma}^{(r-1)}}, \qquad (4.9)$$

for  $t \leq K$ . Here, the mean  $\widehat{\boldsymbol{x}}_{t|K,\boldsymbol{\gamma}^{(r-1)}} \triangleq \mathbb{E} \{ \boldsymbol{x}_t | \boldsymbol{y}^K; \boldsymbol{\gamma}^{(r-1)} \}$ ; and the covariance  $\boldsymbol{P}_{t|K,\boldsymbol{\gamma}^{(r-1)}}$ and the cross-covariance  $\boldsymbol{P}_{t,t-1|K,\boldsymbol{\gamma}^{(r-1)}}$  are defined as

$$\boldsymbol{P}_{t|K,\boldsymbol{\gamma}^{(r-1)}} \triangleq \mathbb{E}\left\{ \tilde{\boldsymbol{x}}_{t} \tilde{\boldsymbol{x}}_{t}^{\mathsf{T}} \left| \boldsymbol{y}^{K}; \boldsymbol{\gamma}^{(r-1)} \right. \right\}$$
(4.10)

$$\boldsymbol{P}_{t,t-1|K,\boldsymbol{\gamma}^{(r-1)}} \triangleq \mathbb{E}\left\{\tilde{\boldsymbol{x}}_{t}\tilde{\boldsymbol{x}}_{t-1}^{\mathsf{T}} \left| \boldsymbol{y}^{K}; \boldsymbol{\gamma}^{(r-1)} \right.\right\},\tag{4.11}$$

where  $\tilde{\boldsymbol{x}}_t = \boldsymbol{x}_t - \hat{\boldsymbol{x}}_{t|K,\boldsymbol{\gamma}^{(r-1)}}$ . The calculation of the variables  $\hat{\boldsymbol{x}}_{t|K,\boldsymbol{\gamma}^{(r-1)}}$ ,  $\boldsymbol{P}_{t|K,\boldsymbol{\gamma}^{(r-1)}}$ , and  $\boldsymbol{P}_{t,t-1|K,\boldsymbol{\gamma}^{(r-1)}}$  is implemented using fixed interval Kalman smoothing [2]. Maximizing

 $Q(\boldsymbol{\gamma}, \boldsymbol{\gamma}^{(r-1)})$  with respect to  $\boldsymbol{\gamma}$ , we get the following M-step:

$$\boldsymbol{\gamma}^{(r)} = \frac{1}{K} \operatorname{Diag} \left\{ \left( \boldsymbol{I} - \boldsymbol{D}^2 \right)^{-1} \sum_{t=2}^{K} \boldsymbol{T}_{t|K,\boldsymbol{\gamma}^{(r-1)}} + \boldsymbol{C}_{1|K,\boldsymbol{\gamma}^{(r-1)}} \right\}.$$
(4.12)

We note that the latency in estimating  $\boldsymbol{x}_{K}$  is 0, that of  $\boldsymbol{x}_{K-1}$  is 1, and so on. Hence, the average latency of the offline KM-SBL algorithm is  $\frac{1}{K}\sum_{t=1}^{K}(K-t) = (K-1)/2$ . We now present our online algorithm.

# 4.3 Iterative Online Algorithm Development

### 4.3.1 Fixed Lag KSBL

The natural extension of the offline KSBL to partial data is to run the forward and backward recursions on the available data at each time, as each new measurement vector arrives. However, this approach requires storage of all forward variables, and it is computationally expensive. Hence, we propose to estimate the hyperparameters using data in a small fixed time window of size  $\Delta + 1$ , and produce smoothed estimates of all sparse vectors in that window. This leads us to the fixed lag Kalman smoothing, which estimates the state in a sliding window of time [115]. Note that the conventional fixed lag smoothing algorithms assume complete knowledge of the signal statistics, i.e.,  $\Gamma$ , whereas here we have to adaptively estimate  $\Gamma$  while computing the smoothed estimates. We combine fixed lag Kalman smoothing with the EM algorithm to learn the hyperparameter  $\Gamma$  in the M-step, and perform fixed lag filtering to estimate the sparse vector in the E-step.

When  $\boldsymbol{y}_k$  arrives, the algorithm estimates the hyperparameter  $\boldsymbol{\Gamma}$  using the EM algorithm using data over a time window of length  $\Delta + 1$ . The E-step and the M-step for the fixed

**Algorithm 1** E-Step of Fixed Lag KSBL at time k in the  $r^{\text{th}}$  EM iteration

```
Input: \boldsymbol{y}_k, \boldsymbol{A}_k, \boldsymbol{\Gamma}_k^{(r-1)}
      Initialization: \Gamma = \Gamma_k^{(r-1)}
      Prediction:
      \hat{\boldsymbol{x}}_{k|k-1} = \boldsymbol{D}\hat{\boldsymbol{x}}_{k-1|k-1}
       {m P}_{k|k-1} = {m D} {m P}_{k-1|k-1} {m D} + \Gamma({m I}_N - {m D}^2)
      oldsymbol{J}_k = oldsymbol{A}_k^{\mathsf{T}} \left(oldsymbol{A}_k oldsymbol{P}_{k|k-1} oldsymbol{A}_k^{\mathsf{T}} + oldsymbol{R}
ight)^{-1}
      \boldsymbol{v}_k = \boldsymbol{y}_k - \boldsymbol{A}_k \hat{\boldsymbol{x}}_{k|k-1}
      for t = 0, 1, 2, ..., \Delta do
            Smoothing:
\boldsymbol{G}_{k}^{(t)} = \boldsymbol{P}_{k,k-t|k-1}^{\mathsf{T}} \boldsymbol{J}_{k}
            \hat{oldsymbol{x}}_{k-t|k} = \hat{oldsymbol{x}}_{k-t|k-1} + oldsymbol{G}_k^{(t)} oldsymbol{v}_k
            \boldsymbol{P}_{k-t|k} = \boldsymbol{P}_{k-t|k-1} - \boldsymbol{G}_k^{(t)} \boldsymbol{A}_k \boldsymbol{P}_{k,k-t|k-1}
            if t \neq \Delta then
                   egin{aligned} \dot{m{P}}_{k-t,k-t-1|k} = \left(m{I}_N - m{G}_k^{(t)}m{A}_k
ight)m{P}_{k,k-t-1|k-1} \end{aligned}
            end if
            oldsymbol{P}_{k+1,k-t|k} = oldsymbol{D} \left( oldsymbol{I}_N - oldsymbol{G}_k^{(0)} oldsymbol{A}_k 
ight) oldsymbol{P}_{k,k-t|k-1}
      end for
```

lag smoothing scheme are

$$\begin{split} \textbf{E-step:} \ \mathcal{Q}(\boldsymbol{\Gamma}|\boldsymbol{\Gamma}_{k}^{(r-1)}) &= \mathbb{E} \left\{ \log \left[ p\left(\boldsymbol{Y}_{k-\Delta:k}, \boldsymbol{X}_{k-\Delta:k}\right) \right] \right\} \\ \textbf{M-step:} \ \boldsymbol{\Gamma}_{k}^{(r)} &= \underset{\boldsymbol{\gamma} \in \mathbb{R}_{+}^{N \times 1}; \boldsymbol{\Gamma} = \mathsf{Diag}\{\boldsymbol{\gamma}\}}{\arg \max} \mathcal{Q}(\boldsymbol{\Gamma}|\boldsymbol{\Gamma}^{(r-1)}), \end{split}$$

where the expectation operation is over the distribution of  $\mathbf{X}_{k-\Delta:k}$  conditioned on  $\mathbf{Y}_{k-\Delta:k}$ and parameterized  $\mathbf{\Gamma}_{k}^{(r-1)}$ . Similar to the offline KSBL, the E-step involves computation of the posterior statistics of the sparse vectors over the time window. We employ fixed lag smoothing to estimate these, and the pseudocode is given in Algorithm 1, where  $\mathbf{\Gamma}_{k}^{(0)} = \mathbf{\Gamma}_{k-1}$  and  $\mathbf{\Gamma}_{1}^{(0)} = \mathbf{I}_{N}$ . At the start of algorithm (k = 1), we initialize all vectors and matrices except  $\mathbf{P}_{1|0}$  with all zero entries, and  $\mathbf{P}_{1|0} = \mathbf{I}_{N}$ . For each successive k, the variables are initialized with the estimates obtained in the previous iteration. The M-step in the  $r^{\text{th}}$  iteration can be simplified to a closed form expression for the new estimate of hyperparameters given by

$$\operatorname{Diag}\left\{\boldsymbol{\Gamma}_{k}^{(r)}\right\} = \frac{1}{\Delta+1}\operatorname{Diag}\left\{\left(\boldsymbol{I}_{N}-\boldsymbol{D}^{2}\right)^{-1}\sum_{t=k-\Delta+1}^{k}\boldsymbol{T}_{t|k;\boldsymbol{\Gamma}_{k}^{(r-1)}}+\boldsymbol{C}_{k-\Delta|k;\boldsymbol{\Gamma}_{k}^{(r-1)}}\right\}.$$
 (4.13)

The proof is similar to that for the offline KSBL [3], and hence omitted. The E-step and the M-step are repeated until convergence of  $\Gamma_k$ , and upon convergence the algorithm outputs the estimate of  $\boldsymbol{x}_{k-\Delta}$  as  $\hat{\boldsymbol{x}}_{k-\Delta|k}$ .

### 4.3.2 Sawtooth Lag KSBL

Although our fixed lag KSBL possess low computational complexity and storage, our numerical experiments show that the performance of the fixed lag comes close to that of offline KSBL only for large number of measurements K. The reason for the poor performance of the fixed lag filter is that it has only one step of forward recursion and  $\Delta$  steps of backward recursions. This results in using different values of covariance matrix estimates  $\Gamma$  for updating the estimate of state statistics at each value of t, i.e.,  $\{\hat{x}_t, P_t, P_{t,t-1}\}$ , in different rounds of the EM iterations. In the initial part of the algorithm, the estimates of  $\Gamma$  in consecutive rounds of EM algorithm could possibly have different support. Hence, the support recovery performance of the algorithm is poor when K is small. In this section, we present another online scheme for the same recovery problem using a smoothing scheme known as the sawtooth lag filter, which uses the same value of  $\Gamma$  for computing the state statistics at any given t [116]. The sawtooth lag scheme is a compromise between the fixed interval and the fixed lag smoothing. Here, the fixed interval smoothing is run on overlapping blocks of data. In the E-step, the forward and backward recursions are run on a data block of size  $\Delta + 1$ , as given in Algorithm 2. The M-step is given by (4.13). Upon

Algorithm 2 E-Step of Sawtooth Lag KSBL at time  $k = i\overline{\Delta} + \Delta + 1$  in the  $r^{\text{th}}$  iteration

```
Input: \boldsymbol{Y}_{k-\Delta:k}, \boldsymbol{A}_{k-\Delta:k}, \boldsymbol{\Gamma}_{i}^{(r-1)}
     Initialization: \Gamma = \Gamma_i^{(r-1)}
     for t = k - \Delta, \ldots, k do
          Prediction:
          \hat{x}_{t|t-1} = D\hat{x}_{t-1|t-1}
          P_{t|t-1} = DP_{t-1|t-1}D + \Gamma(I_N - D^2)
          Filtering:
          oldsymbol{G}_t = oldsymbol{P}_{t|t-1}oldsymbol{A}_t^{\mathsf{T}} \left(oldsymbol{R} + oldsymbol{A}_toldsymbol{P}_{t|t-1}oldsymbol{A}_t^{\mathsf{T}}
ight)^{-1}
          \hat{x}_{t|t} = \hat{x}_{t|t-1} + \hat{G}_t(y_t - A_t \hat{x}_{t|t-1})
          \boldsymbol{P}_{t|t} = (\boldsymbol{I}_N - \boldsymbol{G}_t \boldsymbol{A}_t) \boldsymbol{P}_{t|t-1}
     end for
      oldsymbol{P}_{k,k-1|k} = \left(oldsymbol{I}_N - oldsymbol{G}_koldsymbol{A}_k
ight)oldsymbol{D}oldsymbol{P}_{k-1|k-1}
     for t = k, k - 1, ..., k - \Delta + 1 do
          Smoothing:
          J_{t-1} = P_{t-1|t-1} D P_{t|t-1}^{-1}
          \hat{x}_{t-1|K} = \hat{x}_{t-1|t-1} + \hat{J}_{t-1}(\hat{x}_{t|K} - \hat{x}_{t|t-1})
          P_{t-1|K} = P_{t-1|t-1} + J_{t-1}(P_{t|K} - P_{t|t-1})J_{t-1}^{\mathsf{T}}
          if t \neq \Delta then
                \dot{\boldsymbol{P}}_{t,t-1|k} = \boldsymbol{P}_{t|t} \boldsymbol{J}_{t-1}^{\mathsf{T}} + \boldsymbol{J}_t \left( \boldsymbol{P}_{t+1,t|k} - \boldsymbol{D} \boldsymbol{P}_{t|t} 
ight) \boldsymbol{J}_{t-1}^{\mathsf{T}}
          end if
     end for
```

convergence of the EM procedure, the algorithm outputs the estimates of sparse vectors corresponding to the first  $\overline{\Delta} < \Delta$  indices in the block, i.e., at time k, the estimates at times  $t = k - \Delta, \dots, k - \Delta + \overline{\Delta}$  are declared as  $\hat{x}_{t|k}$ . Then, algorithm waits for next set of  $\overline{\Delta}$  measurement vectors before proceeding further, shifting the time window by  $\overline{\Delta}$ .

Unlike the previous algorithm, the sawtooth lag KSBL waits for a block of data of size  $\overline{\Delta}$ , which is typically much smaller than the total number of observations. The EM procedure is run at times  $k = i\overline{\Delta} + \Delta + 1$ , i = 0, 1, ... and when  $i^{\text{th}}$  EM procedure converges estimates of sparse vectors at time  $t = i\overline{\Delta} + 1$ ,  $i\overline{\Delta} + 2, ..., (i+1)\overline{\Delta}$ , are declared. Hence, the latency in estimation is not fixed, but varies between  $\Delta - \overline{\Delta} + 1$  and  $\Delta$ ; the average latency is  $\Delta - (\overline{\Delta} - 1)/2$ . As in the previous case, at the start of algorithm (i = 0), we initialize the algorithm with  $\Gamma_0^{(0)} = I_N$ ,  $\hat{x}_{1|0} = \mathbf{0}_N$ , and  $P_{1|0} = I_N$ . For each successive i, the variables are initialized with the estimates obtained in the previous iteration. Based on our numerical experiments, the computational and storage demand of the sawtooth lag KSBL is comparable to that of the fixed lag KSBL, while its the performance is close to that of the offline KSBL. This happens because the sawtooth lag scheme has both forward and backward recursions, whereas the fixed lag scheme has only backward computations. Thus, it combines the best of both the fixed lag and the fixed interval procedures. *Remark:* Sawtooth lag smoothing reduces to the optimal offline KSBL if when  $\bar{\Delta} = \Delta = K$ . Typically,  $\bar{\Delta}$  is chosen as  $\Delta/2$  or smaller [116].

#### 4.3.3 Complexity Analysis

#### **Computational Cost**

We assume that the multiplication of a  $p \times q$  matrix with a  $q \times r$  matrix is of the order pqrflops, and inversion of a  $p \times p$  matrix is of the order  $p^3$  flops [117]. Also, we assume that  $m \ll N$ , and for simplicity, we neglect lower order terms involved in computational complexity. We also note that the overall computational complexity of the fixed lag smoothing scheme and the sawtooth lag smoothing scheme scale with the number of observation vectors K, but the complexity per EM iteration is independent of K. However, simulation results show that the overall run time of our online algorithms is much smaller then the offline algorithm.

#### Memory Requirement

In the case of the offline algorithm, we need to save all forward variables, which demands memory that scales with K. For the fixed lag and the sawtooth lag smoothing schemes, data is processed over a small time window. Thus, the memory requirements do not scale

Smoothing	Computational cost	Memory	Average latency	
	per EM iteration	Requirement		
Offline scheme	$\mathcal{O}(KN^3)$	$\mathcal{O}\left(KN^2 ight)$	(K-1)/2	
Fixed Lag	$\mathcal{O}\left(\Delta N^2 m\right)$	$\mathcal{O}\left(\Delta N^2 ight)$	Δ	
Sawtooth Lag	$\mathcal{O}(\Delta N^3)$	$\mathcal{O}\left(\Delta N^2 ight)$	$\Delta - (\bar{\Delta} - 1)/2$	

Table 4.1: Comparison of Online schemes against Offline scheme when K observations are available

with K, a primary advantage of our online algorithms. The variables that need to be stored are the statistics of the sparse vectors, which is of the order  $N^2$ .

We compare the computational demands and memory requirements of the three algorithms in Table 4.1. Next, we present an improved version of the above algorithms which demands lesser computational resources.

# 4.4 Non-iterative Online Algorithm Development

In the non-iterative version of KM-SBL, we process the data sequentially, without waiting for the complete input to arrive or storing all the data that has already arrived. Since we do not store data, it is not feasible to compute the mean  $\hat{x}_{t|K}$ ,<sup>2</sup> the covariance  $P_{t|K}$ , and the cross-covariance  $P_{t,t-1|K}$ . Instead, we approximate them with  $\hat{x}_{t|t+\Delta}$ ,  $P_{t|t+\Delta}$ , and  $P_{t,t-1|t+\Delta}$ , respectively. Then,

$$Q_{k}\left(\boldsymbol{\gamma},\boldsymbol{\gamma}^{k-1}\right) \approx a_{k} - \frac{k}{2}\log|\boldsymbol{\Gamma}| - \frac{1}{2}\operatorname{Tr}\left\{\boldsymbol{\Gamma}^{-1}\boldsymbol{C}_{1|\Delta}\right\} - \frac{1}{2}\operatorname{Tr}\left\{\boldsymbol{\Gamma}^{-1}\left(\boldsymbol{I}-\boldsymbol{D}^{2}\right)^{-1}\sum_{t=2}^{k}\boldsymbol{T}_{t|t+\Delta}\right\} \quad (4.14)$$

<sup>&</sup>lt;sup>2</sup>For brevity, we drop  $\gamma$  from the subscript.

where the constant  $a_k$  is independent of  $\gamma$ .

Maximizing  $Q_k(\boldsymbol{\gamma}, \boldsymbol{\gamma}^{k-1})$  with respect to  $\boldsymbol{\gamma}$ , we have the following recursion

$$\boldsymbol{\gamma}_{k} = \frac{1}{k} \mathsf{Diag} \left\{ \left( \boldsymbol{I} - \boldsymbol{D}^{2} \right)^{-1} \sum_{t=2}^{k} \boldsymbol{T}_{t|t+\Delta} + \boldsymbol{C}_{1|\Delta} \right\}$$
(4.15)

$$= \boldsymbol{\gamma}_{k-1} + \frac{1}{k} \operatorname{Diag} \left\{ \left( \boldsymbol{I} - \boldsymbol{D}^2 \right)^{-1} \boldsymbol{T}_{k|k+\Delta} - \boldsymbol{\Gamma}_{k-1} \right\}.$$
(4.16)

Thus,  $\gamma_k$  can be estimated using  $\gamma_{k-1}$  and  $T_{k|k+\Delta}$ . We next present a procedure to recursively estimate  $T_{k|k+\Delta}$ .

#### 4.4.1 Implementation of the Algorithm

In order to compute  $T_{k|k+\Delta}$ , we need to recursively update the mean  $\hat{x}_{k|k+\Delta}$ , the autocovariance  $P_{k|k+\Delta}$ , and the cross-covariance  $P_{k,k-1|k+\Delta}$ . We describe two implementations: a fixed lag scheme and a sawtooth lag scheme.

#### Fixed Lag Scheme

We consider a Kalman filter designed for the following state space model with state variables as  $\boldsymbol{x}_k$  and measurement variables as  $\boldsymbol{\tilde{y}}_k \triangleq \boldsymbol{y}_{k+\Delta}$ . From (4.1),

$$\widetilde{\boldsymbol{y}}_{k} = \boldsymbol{A}_{k+\Delta} \boldsymbol{D}^{\Delta} \boldsymbol{x}_{k} + \boldsymbol{A}_{k+\Delta} \sum_{i=0}^{\Delta-1} \boldsymbol{D}^{i} \boldsymbol{z}_{k+\Delta-i} + \boldsymbol{w}_{k+\Delta}$$
$$= \widetilde{\boldsymbol{A}}_{k} \boldsymbol{x}_{k} + \widetilde{\boldsymbol{w}}_{k}$$
(4.17)

where  $\widetilde{\boldsymbol{A}}_{k} \triangleq \boldsymbol{A}_{k+\Delta} \boldsymbol{D}^{\Delta}$  and  $\widetilde{\boldsymbol{w}}_{k} \sim \mathcal{N}\left(0, \widetilde{\boldsymbol{R}}_{k}\right)$ . Since the covariance of  $\boldsymbol{z}_{k+\Delta-i}$  is  $(\boldsymbol{I} - \boldsymbol{D}^{2})\boldsymbol{\Gamma}$ , it is easy to show that

$$\widetilde{\boldsymbol{R}}_{k} = \boldsymbol{A}_{k+\Delta} \left( \boldsymbol{I} - \boldsymbol{D}^{2\Delta} \right) \boldsymbol{\Gamma} \boldsymbol{A}_{k+\Delta}^{\mathsf{T}} + \boldsymbol{R}_{k+\Delta}.$$
(4.18)

The new state space model is given by (4.1) and (4.17). The Kalman filter equations for the new system are given below:

$$\widehat{\boldsymbol{x}}_{k|k+\Delta-1} = \boldsymbol{D}\widehat{\boldsymbol{x}}_{k-1|k+\Delta-1} \tag{4.19}$$

$$\boldsymbol{P}_{k|k+\Delta-1} = \boldsymbol{D}\boldsymbol{P}_{k-1|k+\Delta-1}\boldsymbol{D} + \left(\boldsymbol{I} - \boldsymbol{D}^2\right)\boldsymbol{\Gamma}$$
(4.20)

$$\boldsymbol{J}_{k} = \boldsymbol{P}_{k|k+\Delta-1} \widetilde{\boldsymbol{A}}_{k}^{\mathsf{T}} \left( \widetilde{\boldsymbol{A}}_{k} \boldsymbol{P}_{k|k+\Delta-1} \widetilde{\boldsymbol{A}}_{k}^{\mathsf{T}} + \widetilde{\boldsymbol{R}}_{k} \right)^{-1}$$
(4.21)

$$\widehat{\boldsymbol{x}}_{k|k+\Delta} = (\boldsymbol{I} - \boldsymbol{J}_k \widetilde{\boldsymbol{A}}_k) \widehat{\boldsymbol{x}}_{k|k+\Delta-1} + \boldsymbol{J}_k \boldsymbol{y}_{k+\Delta}$$
(4.22)

$$\boldsymbol{P}_{k|k+\Delta} = (\boldsymbol{I} - \boldsymbol{J}_k \boldsymbol{\tilde{A}}_k) \boldsymbol{P}_{k|k+\Delta-1}$$
(4.23)

$$\boldsymbol{P}_{k,k-1|k+\Delta} = (\boldsymbol{I} - \boldsymbol{J}_k \widetilde{\boldsymbol{A}}_k) \boldsymbol{D} \boldsymbol{P}_{k-1|k+\Delta-1}.$$
(4.24)

As every measurement vector  $\boldsymbol{y}_{k+\Delta}$  arrives, the algorithm updates  $\boldsymbol{\gamma}$  using (4.16). Then, the online estimate of  $\boldsymbol{x}_k$  can be computed using forward and backward recursions of a fixed interval Kalman smoother on the block of data of size  $\Delta+1$ , at times  $t = k, k+1, \ldots, k+\Delta$ , as described in Section 4.2.1.

*Remark:* The above scheme is not applicable when D = 0 and  $\Delta > 0$ , because  $y_{k+\Delta}$  is independent of  $x_k$  in this case. Also, the fixed lag scheme only uses the latest measurement vector to update  $\gamma$ , while one can achieve better performance by using all the available measurements in a window around the time instant of interest. In the following subsection, we present a sawtooth lag scheme that addresses the above issues.

#### Sawtooth Lag Scheme

In this scheme, we update  $\gamma$  as every data block of size  $\overline{\Delta} \leq \Delta + 1$  arrives. Consider  $k \in [k_l + 1, k_l + \overline{\Delta}]$  where  $k_l \triangleq (l-1)\overline{\Delta}$  for the update index  $l = 1, 2, \ldots$ . We replace the fixed lag variables  $\widehat{\boldsymbol{x}}_{k|k+\Delta}$ ,  $\boldsymbol{P}_{k|k+\Delta}$ , and  $\boldsymbol{P}_{k,k-1|k+\Delta}$  with variables  $\widehat{\boldsymbol{x}}_{k|\tilde{k}_l}$ ,  $\boldsymbol{P}_{k|\tilde{k}_l}$ , and  $\boldsymbol{P}_{k,k-1|\tilde{k}_l}$ ,

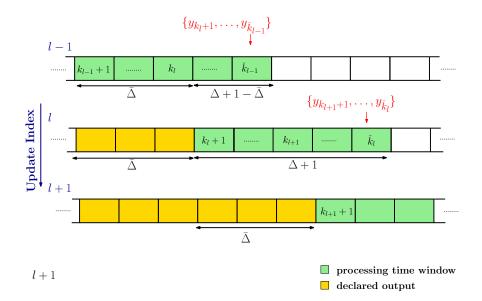


Figure 4.1: The sawtooth lag processing scheme

respectively, where  $\check{k}_l \triangleq k_l + \Delta + 1$ . We compute these variables using the estimate of  $\gamma$  obtained in the previous update,  $\gamma_{l-1}$ . For the  $l^{\text{th}}$  update, (4.15) modifies to

$$\gamma_{l} = \frac{1}{k_{l+1}} \operatorname{Diag} \left\{ \left( \boldsymbol{I} - \boldsymbol{D}^{2} \right)^{-1} \sum_{i=1}^{l} \sum_{\substack{t=k_{i}+1, \\ t \neq 1}}^{k_{i+1}} \boldsymbol{T}_{t|\check{k}_{i}} + \boldsymbol{C}_{1|\Delta} \right\}$$
$$= \gamma_{l-1} + \frac{1}{k_{l+1}} \sum_{\substack{t=k_{l}+1 \\ t=k_{l}+1}}^{k_{l+1}} \operatorname{Diag} \left\{ \left( \boldsymbol{I} - \boldsymbol{D}^{2} \right)^{-1} \boldsymbol{T}_{t|\check{k}_{l}} - \boldsymbol{\Gamma}_{l-1} \right\}.$$
(4.25)

To compute  $T_{t|\check{k}_l}$ , we run the fixed interval Kalman smoothing algorithm on overlapping blocks of data of size  $\Delta + 1$ , and discard the last  $\Delta + 1 - \bar{\Delta}$  values of every block (this is referred to as sawtooth lag smoothing [116]). The processing window is shifted by  $\bar{\Delta}$  after every update. The update equations are comprised of forward recursions and backward recursions. In the forward recursions, we estimate  $\hat{x}_{t|t}$  and  $P_{t|t}$  for  $t = k_l + 1, k_l + 2, \dots, \check{k}_l$  using a Kalman filter as given below:

$$\widehat{\boldsymbol{x}}_{t|t-1} = \boldsymbol{D}\widehat{\boldsymbol{x}}_{t-1|t-1} \tag{4.26}$$

$$\boldsymbol{P}_{t|t-1} = \boldsymbol{D}\boldsymbol{P}_{t-1|t-1}\boldsymbol{D} + (\boldsymbol{I} - \boldsymbol{D}^2)\boldsymbol{\Gamma}$$
(4.27)

$$\boldsymbol{J}_{t} = \boldsymbol{P}_{t|t-1} \boldsymbol{A}_{t}^{\mathsf{T}} \left( \boldsymbol{A}_{t} \boldsymbol{P}_{t|t-1} \boldsymbol{A}_{t}^{\mathsf{T}} + \boldsymbol{R}_{t} \right)^{-1}$$
(4.28)

$$\hat{\boldsymbol{x}}_{t|t} = (\boldsymbol{I} - \boldsymbol{J}_t \boldsymbol{A}_t) \hat{\boldsymbol{x}}_{t|t-1} + \boldsymbol{J}_t \boldsymbol{y}_t$$
(4.29)

$$\boldsymbol{P}_{t|t} = (\boldsymbol{I} - \boldsymbol{J}_t \boldsymbol{A}_t) \boldsymbol{P}_{t|t-1}$$
(4.30)

$$\boldsymbol{P}_{\check{k}_l,\check{k}_l-1|\check{k}_l} = \left(\boldsymbol{I} - \boldsymbol{J}_{\check{k}_l}\boldsymbol{A}_{\check{k}_l}\right)\boldsymbol{D}\boldsymbol{P}_{\check{k}_l-1|\check{k}_l-1}.$$
(4.31)

In the backward recursions, we estimate  $\hat{x}_{t|\check{k}_l}$ ,  $P_{t|\check{k}_l}$  and  $P_{t,t-1|\check{k}_l}$  in the reverse order. For  $t = \check{k}_l, \check{k}_l - 1, \dots, k_l + 2$  we get the following smoothing equations:

$$G_{t-1} = P_{t-1|t-1} D P_{t|t-1}^{-1}$$
(4.32)

$$\hat{\boldsymbol{x}}_{t-1|\check{k}_l} = \hat{\boldsymbol{x}}_{t-1|t-1} + \boldsymbol{G}_{t-1}(\hat{\boldsymbol{x}}_{t|\check{k}_l} - \hat{\boldsymbol{x}}_{t|t-1})$$
(4.33)

$$\boldsymbol{P}_{t-1|\check{k}_l} = \boldsymbol{P}_{t-1|t-1} + \boldsymbol{G}_{t-1}(\boldsymbol{P}_{t|\check{k}_l} - \boldsymbol{P}_{t|t-1})\boldsymbol{G}_{t-1}^{\mathsf{T}}$$
(4.34)

For  $t \neq \check{k}_l$ 

$$\boldsymbol{P}_{t,t-1|\check{k}_{l}} = \boldsymbol{P}_{t|t}\boldsymbol{G}_{t-1}^{\mathsf{T}} + \boldsymbol{G}_{t}\left(\boldsymbol{P}_{t+1,t|\check{k}_{l}} - \boldsymbol{D}\boldsymbol{P}_{t|t}\right)\boldsymbol{G}_{t-1}^{\mathsf{T}}.$$
(4.35)

The scheme is illustrated in Section 4.4.1. In the figure, each box represents a time (sampling) instant with which it is indexed, and each row corresponds to an update index, with the index indicated in blue. The set of  $\boldsymbol{y}$  in red represents the new measurement set processed in each update. A green box (with indices  $k_l + 1 = (l-1)\bar{\Delta} + 1$  to  $\check{k}_l = (l-1)\bar{\Delta} + \Delta + 1$ ) indicates that the state statistics corresponding to the index on box are updated, a yellow box (with indices  $k \leq k_l = (l-1)\bar{\Delta}$ ) indicates that the state statistics

are not updated, and a white box (with indices  $k \ge \tilde{k}_l = (l-1)\bar{\Delta} + \Delta + 1$ ) indicates that the state statistics have not been computed yet. The processing window indicated by green is shifted by  $\bar{\Delta}$  after every update. The average latency of the fixed lag scheme is  $\Delta$ , whereas that of the sawtooth lag scheme is  $\Delta - (\bar{\Delta} - 1)/2$ . In the sawtooth lag scheme,  $\bar{\Delta}$  also controls the frequency of update of  $\gamma$ . If  $\bar{\Delta}$  is large, the average latency decreases, but the  $\gamma$  gets updated more slowly. So, there is a tradeoff between the accuracy and the latency in selecting  $\bar{\Delta}$ .

Next, we discuss the special case of D = 0. We refer to this algorithm as the online *M-SBL algorithm*, as there is no role for Kalman filtering when D = 0.

Online M-SBL: When the sparse vectors are uncorrelated, i.e.,  $\mathbf{D} = \mathbf{0}$ , (4.25) simplifies to the following recursion:

$$\boldsymbol{\gamma}_{l} = \boldsymbol{\gamma}_{l-1} + \frac{1}{k_{l+1}} \sum_{t=k_{l}+1}^{k_{l+1}} \operatorname{Diag} \left\{ \boldsymbol{P}_{t}(\boldsymbol{\gamma}_{l-1}) + \widehat{\boldsymbol{x}}_{t}(\boldsymbol{y}_{t}, \boldsymbol{\gamma}_{l-1}) \widehat{\boldsymbol{x}}_{t}(\boldsymbol{y}_{t}, \boldsymbol{\gamma}_{l-1})^{\mathsf{T}} - \boldsymbol{\Gamma}_{l-1} \right\}.$$
(4.36)

where

$$\boldsymbol{P}_{t}(\boldsymbol{\gamma}) \triangleq \boldsymbol{\Gamma} - \boldsymbol{\Gamma} \boldsymbol{A}_{t}^{\mathsf{T}} \left( \boldsymbol{A}_{t} \boldsymbol{\Gamma} \boldsymbol{A}_{t}^{\mathsf{T}} + \boldsymbol{R}_{t} \right)^{-1} \boldsymbol{A}_{t} \boldsymbol{\Gamma}$$

$$(4.37)$$

$$\widehat{\boldsymbol{x}}_t(\boldsymbol{y},\boldsymbol{\gamma}) \triangleq \boldsymbol{P}_t(\boldsymbol{\gamma}) \boldsymbol{A}_t^{\mathsf{T}} \boldsymbol{R}_t^{-1} \boldsymbol{y}.$$
(4.38)

We note that this implementation depends only on  $\overline{\Delta}$ , and not on  $\Delta$ , because the set of measurement vectors  $\{\boldsymbol{y}_{t}, t = k_{l+1} + 1, k_{l+1} + 2, \dots, \check{k}_{l}\}$  and the set of sparse vectors  $\{\boldsymbol{x}_{t}, t = k_{l} + 1, k_{l} + 2, \dots, k_{l+1}\}$  are independent.

To summarize, we have presented a fixed lag scheme and a sawtooth lag scheme, for computing  $T_{k|k+\Delta}$  recursively using the data in batches. We next discuss the initialization of the algorithm and several interesting special cases.

#### 4.4.2 Discussion

#### Initialization

The initial estimate of  $\gamma$  can be obtained from the first  $\Delta + 1$  input measurements vectors using the offline KM-SBL algorithm. The one round of the offline KM-SBL algorithm can be interpreted as an estimation step, and the recursive update of  $\gamma$  using (4.16) can be interpreted as a tracking process. In fact, if  $\gamma$  is slowly varying over time, the recursive update step (4.16) can track its temporal variations.

#### Special Cases

We make a few interesting observations about the algorithm in the following special cases:

- (a) When D = 0, the sparse vectors are uncorrelated and thus  $\hat{x}_{t|K} = \hat{x}_{t|t+\Delta}$ ,  $P_{t|K} = P_{t|t+\Delta}$ , and  $P_{t,t-1|K} = P_{t,t-1|t+\Delta}$ . Hence, there is no approximation in (4.14). On the other hand, as the correlation coefficient increases, the approximation in (4.14) becomes loose.
- (b) When D = 0 and  $\Delta = 0$ , the fixed lag and the sawtooth lag schemes become identical.
- (c) When  $\Delta = 0$ , the filter for the modified state space reduces to the original Kalman filter equations [2].
- (d) When Δ = 1, the latency of the sawtooth lag scheme equals Δ for all sparse vectors, similar to the fixed lag scheme. Nonetheless, the two schemes are different, because of the forward and backward recursions in the sawtooth lag scheme.

#### 4.4.3 Refinements

#### **Different Learning Rates**

Instead of 1/k in (4.16), any sequence of positive numbers  $b_k$  can be used in the recursive algorithm as long as the following conditions are satisfied:

$$0 \le b_k \le 1 \qquad \sum_{k=1}^{\infty} b_k = \infty \qquad \sum_{k=1}^{\infty} b_k^2 < \infty.$$

$$(4.39)$$

The modified algorithm is given by

$$\boldsymbol{\gamma}_{k} = \boldsymbol{\gamma}_{k-1} + b_{k} \mathsf{Diag} \left\{ \left( \boldsymbol{I} - \boldsymbol{D}^{2} \right)^{-1} \boldsymbol{T}_{k|k+\Delta} - \boldsymbol{\Gamma}_{k-1} \right\}.$$
(4.40)

A good choice for the sequence is  $b_k = 1/k^{\alpha}$ ,  $1/2 < \alpha \leq 1$ , since  $\sum_{k=1}^{\infty} 1/k^{\alpha}$  converges if  $\alpha > 1$  and diverges otherwise. In Section 5.3, we empirically show that the modified algorithm converges faster than the original version (see Figure 5.1).

#### Improved Online M-SBL

Notice that the online M-SBL algorithm in (4.36) does not use the observations  $\boldsymbol{y}_t$ ,  $t = k_{l+1} + 1, k_{l+1} + 2, \ldots, \check{k}_l$ , even though they are available at time  $k_{l+1}$ . Hence, we modify the update step in (4.36) to update  $\boldsymbol{\gamma}$  using all the available measurement vectors  $\boldsymbol{y}^{\check{k}_l}$ , and then estimate the sparse vectors  $\hat{\boldsymbol{x}}_{k_{l+1}}$  to  $\hat{\boldsymbol{x}}_{k_{(l+1)}}$ , as follows:

$$\boldsymbol{\gamma}_{l} = \boldsymbol{\gamma}_{l-1} + \frac{1}{\check{k}_{l}} \sum_{t=\check{k}_{l}-\bar{\Delta}+1}^{\check{k}_{l}} \operatorname{Diag} \left\{ \boldsymbol{P}_{t}(\boldsymbol{\gamma}_{l-1}) + \widehat{\boldsymbol{x}}_{t}(\boldsymbol{y}_{t},\boldsymbol{\gamma}_{l-1}) \widehat{\boldsymbol{x}}_{t}(\boldsymbol{y}_{t},\boldsymbol{\gamma}_{l-1})^{\mathsf{T}} - \boldsymbol{\Gamma}_{l-1} \right\}.$$
(4.41)

Thus, for each update, we use only the latest available block of size  $\overline{\Delta}$ , and not the past values which have already been used. Hence, in this case, we need not store any of the past measurements or the sparse vector estimates.

Scheme		Computational	Memory demand	
		$\cos t$		
$ \begin{array}{c} \text{KM-SBL} \\ (\boldsymbol{D} \neq \boldsymbol{0}) \end{array} $	Offline	$\mathcal{O}(KN^3)$	$\mathcal{O}\left(KN^2 ight)$	
	Fixed lag	$\mathcal{O}\left(KN^2m\right)$	$\mathcal{O}\left(\Delta N^2 ight)$	
	Sawtooth lag	$\mathcal{O}(KN^3)$	$\mathcal{O}\left(\Delta N^2 ight)$	
M-SBL	Offline	$\mathcal{O}(KN^2m)$	$\mathcal{O}\left(Km+N^2\right)$	
(D = 0)	Online	$\mathcal{O}(KN^2m)$	$\mathcal{O}\left(\Delta m + N^2\right)$	

Table 4.2: Comparison of the online schemes with the offline scheme when K observations are available.

#### 4.4.4 Complexity Analysis

We now briefly discuss the computational complexity and memory requirements of our algorithms. We note that Table 4.1 summarized the computational demands of every iteration of the iterative algorithms whereas here in Table 4.2 we summarize the total computational requirements of the non-iterative algorithm as every measurement block arrive.

#### **Computational Cost**

We assume that the multiplication of a  $p \times q$  matrix with a  $q \times r$  matrix requires  $\mathcal{O}(pqr)$ floating-point operations (flops), and the inversion of a  $p \times p$  positive definite matrix requires  $\mathcal{O}(p^3)$  flops [117].

We note that the computational cost per update of  $\gamma$  in the online scheme depends only on  $\Delta$  (which is  $\ll K$ ), although the overall computational complexity does depend on the number of sparse vectors K. However, simulation results show that the overall run time of our online algorithms grow slowly with K when compared to their offline counterparts (see Figure 4.2a). The order-wise complexity of the online M-SBL algorithm (4.36) is similar to the online KM-SBL fixed-lag scheme, but its run time is much smaller than KM-SBL since it does not involve Kalman filtering or smoothing. Note that, the computational cost of the offline algorithms correspond to the complexity of a single iteration, while that of the online algorithms correspond to the overall complexity, as they are non-iterative in nature.

# 4.5 Numerical Results: Non-iterative Algorithm

We use the following setup to evaluate the performance of the algorithm and corroborate the theoretical results. We generate sparse signals of length N = 60, each with s = 6nonzero entries. The locations of nonzero coefficients are chosen uniformly at random, and the nonzero entries are independent and identically distributed with zero mean and unit variance. The length of measurement vector is chosen as m = 20. The measurement matrices  $A_k$  are generated with independent and Gaussian distributed entries with zero mean, and the columns are normalized to have unit Euclidean norm.

We study the properties of the algorithm for both uncorrelated and highly correlated cases in the following subsections. For the uncorrelated case, we consider the improved online algorithm given by (4.41). We evaluate the performance of our algorithm using the same three metrics used in the last section. We consider two methods to initialize the hyperparameter vector  $\gamma$  for the online schemes, which we term *proper* initialization and *fixed* initialization. Proper initialization refers to initializing  $\gamma$  with its estimate obtained from the first  $\Delta+1$  measurements using the offline KM-SBL algorithm. Fixed initialization refers to initialization  $\gamma$  with a fixed vector (which we take as  $4 \cdot 1$ ).

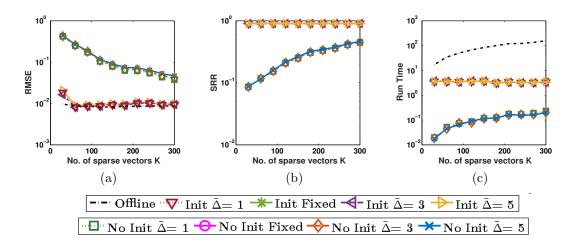


Figure 4.2: Performance of our algorithms relative to the offline algorithm for D = 0 (uncorrelated case, where we use the M-SBL based algorithm). Other parameters are  $\Delta = 5$  and SNR = 20 dB.

#### Uncorrelated Case

Figures 4.2a-4.3c show the performance of the different schemes when D = 0. The curves labeled Offline correspond to the performance of the offline M-SBL algorithm, which is our benchmark, and all other curves correspond to the improved online sawtooth lag scheme discussed in Section 4.4.3. The curves labeled Init  $\overline{\Delta} = 1$ , Init  $\overline{\Delta} = 3$  and Init  $\overline{\Delta} = 5$  correspond to the online algorithm with proper initialization, while the curves labeled No Init  $\overline{\Delta} = 1$ , No Init  $\overline{\Delta} = 3$  and No Init  $\overline{\Delta} = 5$  correspond to the online algorithm with fixed initialization. Our observations from the results are as follows:

• *Initialization:* The online scheme with proper initialization closely matches with the offline scheme in terms of the recovery performance. On the other hand, the online scheme with fixed initialization requires significantly smaller time for execution, but

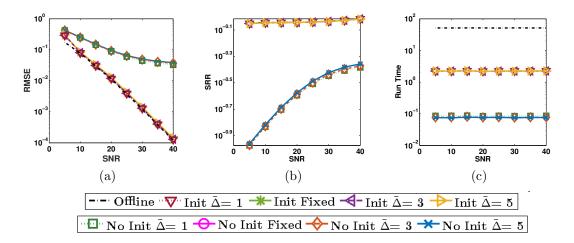


Figure 4.3: Performance of our algorithms relative to the offline algorithm for D = 0 (uncorrelated case, where we use the M-SBL based algorithm). Other parameters are  $\Delta = 5$  and K = 120.

the convergence is slower.

- Number of sparse vectors K: As K increases, the quality of the covariance estimate improves (as seen in Section 5.3.1), and this, in turn, leads to better recovery performance; see Figures 4.2a and 4.2b. From Figure 4.2c, we see that the run time increases almost linearly with K for the offline scheme and the online scheme with fixed initialization. With proper initialization, the run time is roughly constant with K, as most of execution time is spent in computing the initialization of γ.
- SNR: The recovery performance of all algorithms improve with increase in SNR, see Figures 4.3a and 4.3b. Also, the gap between the online scheme with proper initialization and the offline scheme virtually closes beyond an SNR of 10 dB. From Figure 4.3c, the run time remains almost constant with SNR, even though the offline scheme and the online scheme with proper initialization use an iterative step to estimate γ.

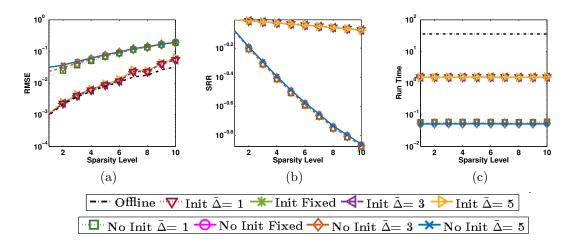


Figure 4.4: Performance of our algorithms relative to the offline algorithm for D = 0 (uncorrelated case, where we use the M-SBL based algorithm). Other parameters are K = 150,  $\Delta = 5$  and SNR = 20 dB.

- Sparsity level: The recovery performance of all algorithms degrade with increase in sparsity level (number of non-zero entries), see Figures 4.4a and 4.4b. However, the SRR performance of the algorithm with fixed initialization degrades significantly with the increase in the sparsity level. From Figure 4.4c, the run time remains almost constant with sparsity level, since the complexity does not depend on the sparsity level.
- Output batch-size Δ: The performance of online schemes do not vary much with Δ, as can be seen from Figures 4.2a-4.3c. However, the recovery performance is slightly better and the run time is slightly worse for smaller values of Δ, as γ is updated more frequently.
- Maximum delay Δ: The performance of the algorithm with varying maximum delay Δ is similar to that of the highly correlated case as shown in Figure 4.5a-Figure 4.5c, and hence omitted. The performance of the online schemes improve as Δ increases,

and the proper initialization can greatly improve the recovery performance compared to fixed initialization. The run time of the online scheme with proper initialization increases with  $\Delta$ , because the number of measurement vectors used to initialize  $\gamma$  increases. However, the behavior the run time of the online schemes for the uncorrelated case is different from that of the highly correlated case, as discussed in Section 4.4.4. This is because the online algorithms use Kalman smoothing in the correlated case, and the complexity of Kalman smoothing increases with  $\Delta$ . In the uncorrelated case, the complexity is independent of  $\Delta$ , thus the run time remains constant for all values of  $\Delta$ .

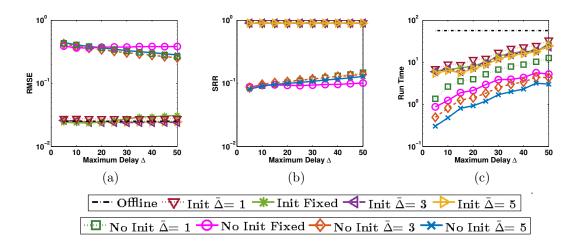


Figure 4.5: Performance of our algorithms relative to the offline algorithm for D = 0.9I (correlated case, where we use the KM-SBL algorithm). Other parameters are  $\Delta = 5$  and SNR = 20 dB.

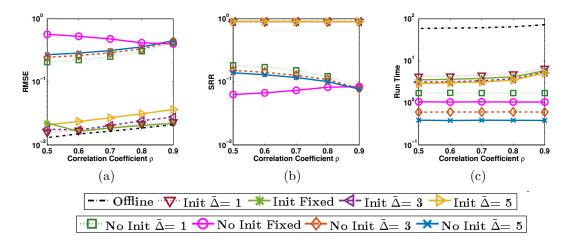


Figure 4.6: Performance of our algorithms relative to the offline algorithm for  $D = \rho I$  (correlated case, where we use the KM-SBL algorithm). Other parameters are K = 150,  $\Delta = 5$  and SNR = 20 dB.

#### Highly Correlated Case

Figures 4.5a-4.6a show the performance of the different algorithms when the sparse vectors are highly correlated ( $D \neq 0$ ). The curves labeled Init Fixed and No Init Fixed correspond to the fixed lag scheme with proper and fixed initialization, respectively, while the other labels are as in the previous plots. Our observations from the results are as follows:

Implementation schemes: As discussed in Section 4.4.1, for the same output batchsize of Δ
= 1, the sawtooth lag scheme outperforms the fixed lag scheme, at the cost of a higher run time. This is because the sawtooth lag scheme uses all the available measurements for updating the hyperparameters, while the fixed lag scheme uses only the latest available measurement. Comparing the fixed lag scheme with the sawtooth lag scheme with higher output batch-sizes (Δ
= 3 and 5), the fixed lag scheme is slower but more accurate, as it updates the hyperparameters more frequently. • Correlation coefficient  $\rho$ : The performance of the algorithms with varying correlation coefficient  $\rho$  (recall  $D = \rho I$ ) is shown in Figures 4.6a-4.6c. As  $\rho$  increases, the recovery performance of the sawtooth lag scheme decreases, while that of the fixed lag scheme improves. This seemingly counterintuitive behavior can be explained as follows. In the offline case, an increase in  $\rho$  can worsen the support recovery of the sparse vectors, but helps the estimation of the amplitude of the nonzero entries. A combination of these effects determine the overall performance of the algorithm, and we see that the recovery performance slightly degrades as the  $\rho$  increases. A similar trend was observed in the SRR for the temporal M-SBL algorithm for recovering correlated sparse vectors [118, Figure 2]. In case of the sawtooth lag scheme, in addition to the above, an increase in  $\rho$  also makes the approximation in (4.14) loose. Due to this, the degradation in the recovery performance of the sawtooth lag scheme is large compared to the offline algorithm. In case of the fixed lag scheme, apart from the effects discussed above, an increase in  $\rho$  also improves  $\rho^{\Delta}$ , the correlation between the state and the observation in the new state space model (described by (4.1) and (4.17)). This improves the quality of the estimate output by the Kalman filter, and in turn helps the recovery. The overall effect of these is an improvement in the recovery performance of the fixed lag scheme. A more rigorous study of the effect of  $\rho$  an interesting topic for future work.

The run time of the algorithm remains the same for all values of  $\rho$  for the fixed initialization case, as its complexity is independent of  $\rho$ . However, the run time of the online schemes with proper initialization is higher in the highly correlated case. This is because, when data is highly correlated, the initialization phase using the offline scheme takes more iterations to converge. We can see a similar slight increase in the run time of the offline scheme in the highly correlated case.

- Maximum delay  $\Delta$ : As the delay increases, the recovery performance of the online schemes increases for both methods of initialization. The change is more evident for the fixed initialization case, as the recovery performance of with proper initialization is very close to that of the offline scheme. We also observe that the improvement in recovery performance is small for the fixed lag scheme compared to the sawtooth lag scheme. This is because of the reduced correlation ( $D^{\Delta}$ ) between the state and the observation of the new state space model given by (4.1) and (4.17). Also as pointed out earlier, the run time of the online schemes increases with  $\Delta$ .
- Output batch-size  $\bar{\Delta}$ : The performance of the online algorithms remains constant with  $\bar{\Delta}$  for both the correlated and uncorrelated case. However, the gap between the run time curves is wider for the correlated case. This is because each update of  $\gamma$ is computationally more expensive due to the Kalman smoothing in the correlated case.

The performance of the online algorithms with K and SNR in the highly correlated case is similar to that observed in the uncorrelated case, and hence omitted.

In the next subsection, we compare the performance of our scheme with other existing online algorithms found in the literature as mentioned in Section 4.1.

#### 4.5.1 Comparison with Existing Algorithms

In Figure 4.7a-Figure 4.7c, we compare our algorithm, labeled Non-iterative KMSBL, with the following algorithms (labels in brackets):

- (i) Offline KM-SBL [3] (Offline KMSBL)
- (ii) Reweighted  $l_1$  dynamic filtering [111] (RL1-DF)
- (iii) Iterative online KM-SBL (Iterative KMSBL)
- (iv) Standard  $l_1$  norm based algorithm on each measurement vector [119] (Regular  $l_1$  Norm)
- (v) Kalman compressed sensing [112] (KF-CS)

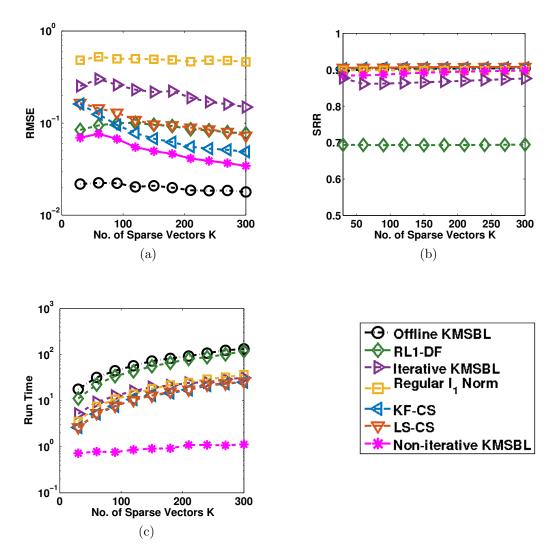


Figure 4.7: Comparison of RMSE, SRR and run time of our algorithm with the existing online schemes.

Here, we choose  $\Delta = 0$ , as the other online schemes except the iterative online KM-SBL algorithm are not designed for  $\Delta > 0$ . We also note that we extended the Kalman compressed sensing algorithm in [112] to handle a first-order AR process with correlation matrix  $\mathbf{D} \in [0, 1]^N$ , while the original algorithm only considers  $\mathbf{D} = \mathbf{I}$ . The recovery performance of our scheme is comparable with the other online schemes algorithms, and approaches the offline performance as K increases. However, the run time of our scheme is significantly lower than all the other schemes. Moreover, the rate of increase of the run time of our scheme with K is much smaller than the other schemes. The significant reduction in the run time is primarily due to the non-iterative nature of our scheme. Since all other algorithms are iterative in nature, their complexity and hence run time depends linearly on the number of iterations which, in turn, depends on N, m, K, the threshold used for stopping the iterations, etc. This brings out the major difference between the other algorithms and our online non-iterative schemes. Thus, our scheme is both fast and accurate, as promised in Section 4.1.

#### 4.5.2 Sparse OFDM Channel Estimation

In this subsection, we consider the sparse OFDM channel estimation problem as an application of our algorithm [3]. We list the simulation parameters in Table 4.3. The sparse channel is of length N = 59, which taken as the length of the cyclic prefixing (CP), with s = 6 nonzero entries for each channel instantiation (PedB channel model [120]). In each OFDM symbol, m = 20 pilot symbols are placed uniformly, and the number of OFDM symbols K is taken as 150. We assume that the algorithms estimate the channel once in every OFDM slot, which gives  $\Delta = 6$ . We consider both coded<sup>3</sup> and uncoded scenarios and three metrics for the performance comparison: BER, MSE in channel estimation, and run time per channel vector estimation. We estimate the channel using the pilot symbols, and decode the data using the channel estimate (for details, refer to [3]). In Figure 4.8a-Figure 4.8c, we compare the performance of our algorithm, labeled Online Non-iterative, with the following three schemes (labels in brackets):

- (i) Offline KM-SBL [3] (Offline)
- (ii) Iterative online KM-SBL (Online Iterative)

(iii) Rece	eiver with perfect knowledge of channel (Gen		e) Value	
		Transmission bandwidth	2.5 MHz	
		Sub-frame duration	$0.5 \mathrm{ms}$	
	OFDM	Subcarrier spacing	$15 \mathrm{~kHz}$	
	(3GPP/LTE)	Sampling frequency	3.84 MHz	
	broadband	FFT size	256	
	standard $[122]$ )	No. of data subcarriers	200	
		OFDM symbol/slot	6	
		CP length	$16.67~\mu \mathrm{s}$	
		Environment	Pedestrian B [120]	
	Channel	Model	Jakes model [123]	
		Norm. Doppler freq.	$10^{-3}$	
	Coding and modulation		rate $1/2$ Turbo code	
			and QPSK	
	Pulse shaping		Raised cosine	
			with rolloff	
			factor= $0.5 [124]$	

Table 4.3: Simulation parameters for OFDM channel estimation

<sup>&</sup>lt;sup>3</sup>For the Turbo code generation, we use the publicly available software [121].

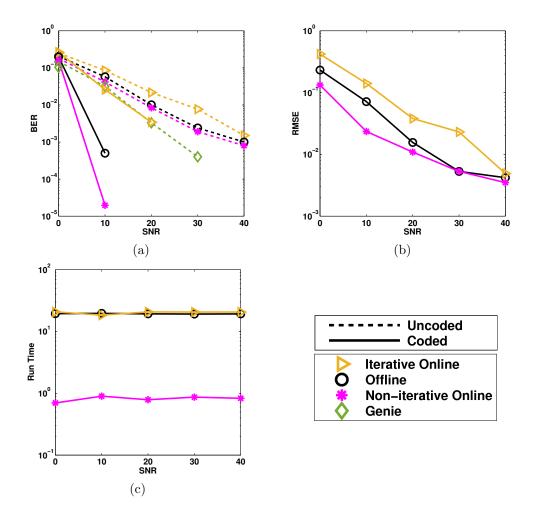


Figure 4.8: Comparison of the BER, RMSE and run time of our algorithm with existing schemes

As mentioned earlier, the other online schemes are not applicable here, as we take  $\Delta > 0$ . From the figure, we infer that the BER and the MSE performance of our algorithm is better than the offline algorithm which was originally proposed for the channel estimation problem [3]. This is because the offline algorithm processes the data in blocks of size 6, and does not reuse the past measurements blocks, whereas our algorithm uses information from all past measurement blocks to estimate the channel vectors for the current block. Moreover, our algorithm has an added advantage of significantly reduced run time.

# 4.6 Summary

The chapter presented algorithms that are particularly useful in scenarios where noisy undetermined linear measurements of sparse state vectors arrive sequentially, and when one wants to exploit structure in the signal beyond (simultaneous) sparsity, specifically, the correlation introduced by the LDS. We developed two algorithms, namely, iterative and non-iterative, by combining the sequential EM procedure and the SBL framework, and presented two schemes for implementation: the fixed lag and sawtooth lag schemes. Our algorithms do not require any parameter tuning. Simulations showed that the performance of our algorithm is close to that of the offline algorithm, but it demands less memory and computational resources, both when the sparse vectors are uncorrelated and highly correlated.

# Chapter 5

# Convergence Analysis of Online M-SBL Algorithm

#### Answering problem Q3 for model version SM2

In this chapter, we continue from the previous chapter and discuss the uniqueness of the solution obtained using the non-iterative online KM-SBL algorithm. The estimator for sparse unknowns is just the MAP estimator, which enjoys the properties of being linear, unbiased and minimum variance. If the hyperparamter iterates converge to the true value, then the accuracy of estimating sparse unknowns is asymptotically be equal to that of an oracle estimator that knows the variances of the entries of the unknowns. Thus, the convergence analysis has implications on the accuracy of the estimates, and thus, provide insights to the uniqueness of the solution. Hence, we establish the uniqueness by examining the convergence properties of the presented algorithms and compute the limit point of the sequence of common hyperparameter iterates generated by the algorithm. However, the analysis for the arbitrary correlation case is hard, because the evolution of the every value in the sequence is a complicated function of the previous value. Therefore, we consider the convergence results for the tow extreme value of correlation: one, uncorrelated case, and two, perfectly correlated case. Intuitively, the algorithm should converge for all intermediate values of the correlation also. Further, this is corroborated by our empirical results.

# 5.1 Uncorrelated Case

In the section, we study the convergence properties of the non-iterative online algorithm under the following assumptions:

- (A1) The measurement matrices are identical, i.e.,  $A_k = A$ ,  $\forall k$ , and without loss of generality, Rank  $\{A\} = m$ .
- (A2) The noise covariance matrix is the same for all measurements, i.e.,  $\mathbf{R}_k = \mathbf{R}, \forall k$ .
- (A3) The sparse vectors are uncorrelated, i.e., D = 0.

The above assumptions are standard in the MMV literature, and are referred to as the joint sparsity model-2 (JSM-2) [37, 39–41]. The assumptions simplify the recursive algorithm, and make the analysis tractable. Since D = 0, the fixed lag scheme discussed in Section 4.4.1 is not applicable, and we focus our analysis on the sawtooth lag implementation. We start with the case when  $\bar{\Delta} = 1$ . A similar analysis follows for  $\bar{\Delta} > 1$ , and we discuss this case later in the sequel.

When  $A_k = A$  and  $R_k = R$ , (4.36)-(4.38) simplify to the following recursion:

$$\boldsymbol{\gamma}_{k} = \boldsymbol{\gamma}_{k-1} + \frac{1}{k} \mathsf{Diag} \left\{ \boldsymbol{P}(\boldsymbol{\gamma}_{k-1}) \right\} + \frac{1}{k} \mathsf{Diag} \left\{ \widehat{\boldsymbol{x}}(\boldsymbol{y}_{k}, \boldsymbol{\gamma}_{k-1}) \widehat{\boldsymbol{x}}(\boldsymbol{y}_{k}, \boldsymbol{\gamma}_{k-1})^{\mathsf{T}} - \boldsymbol{\Gamma}_{k-1} \right\}$$
(5.1)

where  $P(\gamma)$  and  $\hat{x}(y, \gamma)$  are as defined in (4.37) and (4.38), with  $A_t$  and  $R_t$  replaced by A and R, respectively. We can rewrite (5.1) as a stochastic approximation recursion as follows:

$$\boldsymbol{\gamma}_{k} = \boldsymbol{\gamma}_{k-1} + \frac{1}{k} \boldsymbol{f}(\boldsymbol{\gamma}_{k-1}) + \frac{1}{k} \boldsymbol{e}_{k}.$$
(5.2)

Here,  $\boldsymbol{f}(\boldsymbol{\gamma})$  is the mean field function, given by

$$\boldsymbol{f}(\boldsymbol{\gamma}) \triangleq \operatorname{Diag} \left\{ \boldsymbol{P}(\boldsymbol{\gamma}) + \boldsymbol{P}(\boldsymbol{\gamma}) \boldsymbol{A}^{\mathsf{T}} \boldsymbol{R}^{-1} \mathbb{E} \left\{ \boldsymbol{y} \boldsymbol{y}^{\mathsf{T}} \right\} \boldsymbol{R}^{-1} \boldsymbol{A} \boldsymbol{P}(\boldsymbol{\gamma}) \right\} - \boldsymbol{\gamma},$$
(5.3)

where the expectation is over the distribution of  $\boldsymbol{y}$ , and  $\boldsymbol{e}_k$  is given by

$$\boldsymbol{e}_{k} \triangleq \mathsf{Diag}\left\{\boldsymbol{P}(\boldsymbol{\gamma}_{k-1}) + \widehat{\boldsymbol{x}}(\boldsymbol{y}_{k}, \boldsymbol{\gamma}_{k-1})\widehat{\boldsymbol{x}}(\boldsymbol{y}_{k}, \boldsymbol{\gamma}_{k-1})^{\mathsf{T}}\right\} - \boldsymbol{\gamma}_{k-1} - \boldsymbol{f}(\boldsymbol{\gamma}_{k-1}).$$
(5.4)

Further, using  $\boldsymbol{P}(\boldsymbol{\gamma})$  from (4.37),

$$\boldsymbol{P}(\boldsymbol{\gamma}) - \boldsymbol{\Gamma} = -\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} \left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right)^{-1} \boldsymbol{A}\boldsymbol{\Gamma}$$
(5.5)

$$\boldsymbol{P}(\boldsymbol{\gamma})\boldsymbol{A}^{\mathsf{T}}\boldsymbol{R}^{-1} = \boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}}\left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right)^{-1}.$$
(5.6)

Thus, we get the following:

$$f(\boldsymbol{\gamma}) = \operatorname{Diag} \left\{ \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} \left( \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right)^{-1} \left( \mathbb{E} \left\{ \boldsymbol{y} \boldsymbol{y}^{\mathsf{T}} \right\} - \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} - \boldsymbol{R} \right) \left( \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right)^{-1} \boldsymbol{A} \boldsymbol{\Gamma} \right\}$$

$$(5.7)$$

$$\boldsymbol{e}_{k} = \operatorname{Diag} \left\{ \boldsymbol{\Gamma}_{k-1} \boldsymbol{A}^{\mathsf{T}} \left( \boldsymbol{A} \boldsymbol{\Gamma}_{k-1} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right)^{-1} \left( \boldsymbol{y}_{k} \boldsymbol{y}_{k}^{\mathsf{T}} - \mathbb{E} \left\{ \boldsymbol{y} \boldsymbol{y}^{\mathsf{T}} \right\} \right) \left( \boldsymbol{A} \boldsymbol{\Gamma}_{k-1} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right)^{-1} \boldsymbol{A} \boldsymbol{\Gamma}_{k-1} \right\}.$$

$$(5.8)$$

We next present the convergence results of the algorithm. We begin with a proposition which shows that the sequence of  $\gamma_k$  generated by the algorithm is bounded. **Proposition 5.1.** If  $\gamma_0$  is a nonnegative vector, the sequence  $\gamma_k$  generated by (5.1) remains in a compact subset of  $\mathbb{R}^N_+$  almost surely (a.s.).

*Proof.* See Appendix C.1.

The next question to be answered is about the values to which the sequence  $\gamma_k$  could converge. The following theorem characterizes the asymptotic behavior of the algorithm.

**Theorem 5.1.** Assume that the nonzero entries of  $\boldsymbol{x}$  are orthogonal, and the diagonal matrix  $\Gamma_{opt} \triangleq \mathbb{E} \{\boldsymbol{x}\boldsymbol{x}^{\mathsf{T}}\}$ . If  $\boldsymbol{\gamma}_0$  is a nonnegative vector, then the sequence  $\boldsymbol{\gamma}_k$  of the online M-SBL algorithm given by (5.1) converges to an element in the union set:  $\{\boldsymbol{0}\} \cup \{\boldsymbol{\gamma} \in \mathbb{R}^N_+ : \boldsymbol{A}(\boldsymbol{\Gamma} - \boldsymbol{\Gamma}_{opt}) \boldsymbol{A}^{\mathsf{T}} = \boldsymbol{0}\}$  a.s. Further, if Rank  $\{\boldsymbol{A} \odot \boldsymbol{A}\} = N$ , the sequence  $\boldsymbol{\gamma}_k$  converges to a point in the two-element set  $\{\boldsymbol{0}, \boldsymbol{\gamma}_{opt}\}$  a.s.

*Proof.* See Appendix C.2.

We make the following observations from Theorem 5.1.

- The results are independent of the following parameters:
  - (a) sparsity level of the unknown vectors
  - (b) initialization of the algorithm (however,  $\gamma_0 \in \mathbb{R}^N_+$ )
  - (c) distribution of the sparse vectors (even though the algorithm is designed assuming a Gaussian distribution), as long as the entries are orthogonal
  - (d) properties of A, such as its restricted isometry constant or mutual coherence
  - (e) construction of A, i.e., it can be deterministic or random, with normalized or unnormalized columns.

- The convergence guarantee of the original M-SBL algorithm in [41] holds only in the noiseless case. However, our generalized result applies whether noise is present or not. Hence, the result is practically more useful.
- The condition that the nonzero entries of *x* should be orthogonal is similar to the orthogonality condition required for the convergence guarantee of the original M-SBL algorithm in the noiseless case [41]. In fact, the orthogonality condition in [41] is hard to achieve since the number of sparse vectors to be estimated is finite. In that sense, ours is a more reasonable assumption.
- The M-SBL cost function [41] is defined as

$$V_{\text{M-SBL}}(\boldsymbol{\gamma}) = \lim_{k \to \infty} \left[ \frac{1}{k} \sum_{t=1}^{k} \boldsymbol{y}_{t}^{\mathsf{T}} \left( \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right)^{-1} \boldsymbol{y}_{t} + \log |\boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}| \right]$$
  
=  $\mathsf{Tr} \left\{ \left( \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right)^{-1} \left( \boldsymbol{A} \boldsymbol{\Gamma}_{\mathsf{opt}} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right) \right\} - \log \left| \left( \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right)^{-1} \right|.$   
(5.9)

We note that  $V_{\text{M-SBL}}(\boldsymbol{\gamma}) - \log |\boldsymbol{A}\boldsymbol{\Gamma}_{\text{opt}}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}| - m$  is the Kullback-Leibler (KL) divergence between two distributions:  $\mathcal{N}(\boldsymbol{0}, \boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R})$  and  $\mathcal{N}(\boldsymbol{0}, \boldsymbol{A}\boldsymbol{\Gamma}_{\text{opt}}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R})$ . The global minimum of  $V_{\text{M-SBL}}(\boldsymbol{\gamma})$  is therefore achieved at  $\{\boldsymbol{\gamma} \in \mathbb{R}^{N}_{+} : \boldsymbol{A} (\boldsymbol{\Gamma} - \boldsymbol{\Gamma}_{\text{opt}}) \boldsymbol{A}^{\mathsf{T}} = \boldsymbol{0}\}$ . Hence, the set to which our algorithm converges contains all the points achieving the global minimum of  $V_{\text{M-SBL}}(\boldsymbol{\gamma})$ .

- Since  $V_{\text{M-SBL}}(\boldsymbol{\gamma})$  is a function of  $\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}}$ , the smallest set to which M-SBL can converge is  $\{\boldsymbol{\gamma} \in \mathbb{R}^{N}_{+} : \boldsymbol{A} (\boldsymbol{\Gamma} \boldsymbol{\Gamma}_{\text{opt}}) \boldsymbol{A}^{\mathsf{T}} = \boldsymbol{0}\}$ . The  $\boldsymbol{\gamma}_{k}$  output by our algorithm converges to the union of this set with  $\boldsymbol{0}$ .
- It can be shown that the algorithm is guaranteed to converge to a sparse solution,

where, by sparse solution, we mean one with no more than m nonzero entries. Given any *s*-sparse vector  $\boldsymbol{\gamma}_{opt}$  and sensing matrix  $\boldsymbol{A}$ , we can always construct a pair  $(\boldsymbol{x}_c, \boldsymbol{y}_c)$ such that  $\boldsymbol{y}_c = \boldsymbol{A}\boldsymbol{x}_c$  and  $\boldsymbol{x}_c = \Gamma_{opt}^{1/2} (\boldsymbol{A}\Gamma_{opt}^{1/2})^{\dagger} \boldsymbol{y}_c$ . By [23, Theorem 1],  $\boldsymbol{\gamma}_{opt}$  is the global minimizer of the SBL cost function constructed under a noiseless measurement model using  $\boldsymbol{y}_c$  and  $\boldsymbol{A}$ . Further, from [23, Theorem 2], it is known that every local minimum of the SBL cost function is achieved at a sparse solution (even in the presence of noise). Now, the SBL cost is a function of  $\boldsymbol{\Gamma}$  only through  $\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}}$ . Hence, the set  $\{\boldsymbol{\gamma} \in \mathbb{R}^N_+ : \boldsymbol{A} (\boldsymbol{\Gamma} - \boldsymbol{\Gamma}_{opt}) \boldsymbol{A}^{\mathsf{T}} = \boldsymbol{0}\}$  consists of local minima of this SBL cost function, which implies that the elements of the set are all sparse. Therefore, the algorithm is guaranteed to converge to a sparse solution.

We can extend the above convergence results to the refined algorithm given by (4.40) using the following corollary.

Corollary 5.1. Consider the modified online M-SBL algorithm given by (4.40) and having learning rates satisfying (4.39). Under the assumptions of Theorem 5.1, the sequence  $\gamma_k$ converges to a point in the set  $\{0\} \cup \{\gamma \in \mathbb{R}^N_+ : \mathbf{A}(\Gamma - \Gamma_{opt}) \mathbf{A}^T = \mathbf{0}\}$  a.s. Further, if Rank  $\{\mathbf{A} \odot \mathbf{A}\} = N$ , the sequence  $\gamma_k$  converges to a point in the set  $\{\mathbf{0}, \gamma_{opt}\}$  a.s.

The proof of the above is similar to that of Theorem 5.1 because the only properties of the sequence 1/k (in (4.16)) that are used in Theorem 5.1 are the ones listed in (4.39).

We now consider the more general case where  $\overline{\Delta} \geq 1$ . As in the previous case, the algorithm can be rewritten as a stochastic approximation recursion as follows:

$$\boldsymbol{\gamma}_{l} = \boldsymbol{\gamma}_{l-1} + \frac{1}{l} \boldsymbol{f}(\boldsymbol{\gamma}_{l-1}) + \frac{1}{l} \tilde{\boldsymbol{e}}_{l}, \qquad (5.10)$$

where  $f(\gamma)$  is as defined in (5.3), and

$$\tilde{\boldsymbol{e}}_{l} \triangleq -\boldsymbol{f}(\boldsymbol{\gamma}_{l-1}) + \frac{1}{\bar{\Delta}} \sum_{t=k_{l}+1}^{k_{l}+\bar{\Delta}} \operatorname{Diag}\left\{\boldsymbol{P}(\boldsymbol{\gamma}_{l-1}) + \hat{\boldsymbol{x}}(\boldsymbol{y}_{t},\boldsymbol{\gamma}_{l-1}) \hat{\boldsymbol{x}}(\boldsymbol{y}_{t},\boldsymbol{\gamma}_{l-1})^{\mathsf{T}}\right\}.$$
(5.11)

The following theorem characterizes the asymptotic behavior of the above algorithm. Using the theorem, we can also derive a corollary similar to Corollary 5.1. However, we omit the statement to avoid repetition.

**Theorem 5.2.** Under the assumptions of Theorem 5.1, the sequence  $\gamma_l$  output by the online M-SBL algorithm given by (5.10) converges to a point in the set  $\{\mathbf{0}\} \cup \{\boldsymbol{\gamma} \in \mathbb{R}^N_+ : \mathbf{A} (\boldsymbol{\Gamma} - \boldsymbol{\Gamma}_{opt}) \mathbf{A}^{\mathsf{T}} = \mathbf{0}\}$  a.s. Further, if Rank  $\{\mathbf{A} \odot \mathbf{A}\} = N$ , the sequence  $\gamma_l$  converges to a point in the set  $\{\mathbf{0}, \gamma_{opt}\}$  a.s.

*Proof.* The algorithm given by (5.10) differs from the algorithm given by (5.2) only in the last term. The only place where this term plays a role in the proof in Appendix C.2 is via Lemma C.1. Hence, it suffices to show that  $\lim_{l\to\infty} \sum_{i=1}^{l} \frac{1}{i} \tilde{e}_i$  exists and is finite. From (5.11), we get

$$ilde{oldsymbol{e}}_{l} = ext{Diag} igg\{ oldsymbol{\Gamma}_{l-1} oldsymbol{A}^{\mathsf{T}} + oldsymbol{R} ig)^{-1} \ \left( \mathbb{E} \left\{ oldsymbol{y} oldsymbol{y}^{\mathsf{T}} 
ight\} - rac{1}{ar{\Delta}} \sum_{t=k_{i}+1}^{k_{i+1}} oldsymbol{y}_{t} oldsymbol{y}_{t}^{\mathsf{T}} igg) igl(oldsymbol{A} oldsymbol{\Gamma}_{l-1} oldsymbol{A}^{\mathsf{T}} + oldsymbol{R} igr)^{-1} oldsymbol{A} oldsymbol{\Gamma}_{l-1} iggr\}.$$

Now the result follows by replacing  $e_k$  in the proof of Lemma C.1 with  $\tilde{e}_l$ .

We can also get similar convergence results for the improved M-SBL algorithm given by (4.41), as follows.

**Corollary 5.2.** Under the assumptions of Theorem 5.1, the sequence  $\gamma_l$  output by the improved online M-SBL algorithm given by (4.41) converges to a point in the set  $\{\mathbf{0}\} \cup \{\gamma \in \mathcal{I}\}$ 

 $\mathbb{R}^{N}_{+}: \boldsymbol{A}\left(\boldsymbol{\Gamma} - \boldsymbol{\Gamma}_{\mathsf{opt}}\right) \boldsymbol{A}^{\mathsf{T}} = \boldsymbol{0} \} \text{ a.s. Further, if } \mathsf{Rank}\left\{\boldsymbol{A} \odot \boldsymbol{A}\right\} = N, \text{ the sequence } \boldsymbol{\gamma}_{l} \text{ converges}$ to a point in the set  $\{\boldsymbol{0}, \boldsymbol{\gamma}_{\mathsf{opt}}\}$  a.s.

Proof. Under the assumptions of Theorem 5.1, the improved online algorithm given by (4.41) is equivalent to the original algorithm given by (4.36) except that it uses  $\overline{\Delta}$  measurement vectors  $\{\boldsymbol{y}_t, t = \check{k}_l - \overline{\Delta} + 1, \check{k}_l - \overline{\Delta} + 2, \dots, \check{k}_l\}$  instead of  $\overline{\Delta}$  measurement vectors  $\{\boldsymbol{y}_t, t = k_l + 1, k_l + 2, \dots, k_{l+1}\}$  used by the original version. Since the measurement vectors are independent and identically distributed, the rest of the proof is the same as that of Theorem 5.1.

# 5.2 Perfectly Correlated Case

In this section, we consider the convergence results for the other extreme value of correlation, i.e., D = I. We note that when D = I, from (4.1), we get  $x_k = x_1 \triangleq x$  for all values of k. This is because the covariance of  $z_k$  is assumed to be I - D = 0. Further, the Kalman filtering/smoothing equations for estimating the sparse vectors (4.19)-(4.24) and (4.26)-(4.31) become independent of  $\gamma$ . Thus, when  $\Delta = 1$ , the algorithm simplifies to

$$\boldsymbol{J}_{k} = \boldsymbol{P}_{k-1}\boldsymbol{A}_{k}^{\mathsf{T}} \left( \boldsymbol{A}_{k}\boldsymbol{P}_{k-1}\boldsymbol{A}_{k}^{\mathsf{T}} + \boldsymbol{R}_{k} \right)^{-1}$$
(5.12)

$$\widehat{\boldsymbol{x}}_{k} = (\boldsymbol{I} - \boldsymbol{J}_{k}\boldsymbol{A}_{k})\widehat{\boldsymbol{x}}_{k-1} + \boldsymbol{J}_{k}\boldsymbol{y}_{k}$$
(5.13)

$$\boldsymbol{P}_{k} = (\boldsymbol{I} - \boldsymbol{J}_{k} \boldsymbol{A}_{k}) \boldsymbol{P}_{k-1}, \qquad (5.14)$$

where  $\boldsymbol{x}_k$  is the estimate of  $\boldsymbol{x}$  at time k, and  $\boldsymbol{P}_k$  is the corresponding covariance estimate of  $\boldsymbol{x}$ .

When D = I, (4.27) becomes  $P_{k|k-1} = P_{k-1}$ , and hence, here we analyze the converge

of  $P_k$ . Further, when D = 0, we showed that  $A\Gamma_k A^{\mathsf{T}}$  converges to  $A\Gamma_{\mathsf{opt}} A^{\mathsf{T}}$ . Similarly, here we show the convergence of  $AP_kA^{\mathsf{T}}$ .

As before, we first prove that the algorithm converges and then characterize the limit point.

**Proposition 5.2.** Under assumptions A1 and A2, the algorithm given by (5.12)-(5.14)converges as  $k \to \infty$ , provided  $P_0$ , the initialization of the covariance matrix, has full rank.

*Proof.* See Appendix C.3.

Now that we know the algorithm converges, and the next important question is whether the algorithm converges to the right solution. This is addressed in the following theorem.

**Theorem 5.3.** Under assumptions A1 and A2, the sequence  $\hat{x}_k$  of the algorithm given by (5.12)-(5.14) converges to the true solution almost surely, if the initialization  $P_0$  is a full rank matrix.

```
Proof. See Appendix C.4.
```

*Remark 1:* Theorem 5.3 is very general, and holds under a variety of settings. In particular, it is independent of:

- the sparsity level of the unknown vector  $\boldsymbol{x}$
- initializations of  $\widehat{\boldsymbol{x}}_0$  and  $\boldsymbol{P}_0$ , provided  $\boldsymbol{P}_0$  has full rank
- measurement noise level and noise correlation, i.e.,  $\boldsymbol{R}$

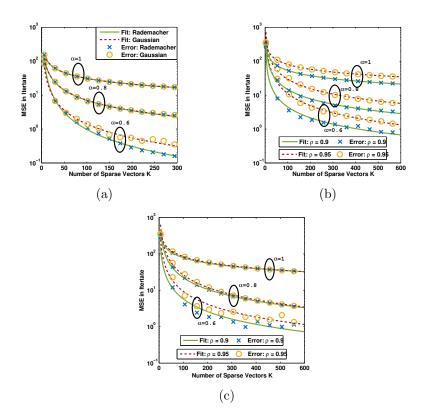


Figure 5.1: Convergence of the hyperparameters to the true value.

# 5.3 Simulation Results

We use the following setup to evaluate the performance of the algorithm and corroborate the theoretical results. We generate sparse signals of length N = 60, each with s = 6nonzero entries. The locations of nonzero coefficients are chosen uniformly at random, and the nonzero entries are independent and identically distributed with zero mean and unit variance. The length of measurement vector is chosen as m = 20. The measurement matrices  $A_k$  are generated with independent and Gaussian distributed entries with zero mean, and the columns are normalized to have unit Euclidean norm.

We study the properties of the algorithm for both uncorrelated and highly correlated cases in the following subsections. For the uncorrelated case, we consider the improved online algorithm given by (4.41).

Algo.	Rademacher Dist.		Gaussian Dist.			
	$\bar{\Delta} = 1$	$\bar{\Delta} = 3$	$\bar{\Delta} = 5$	$\bar{\Delta} = 1$	$\bar{\Delta} = 3$	$\bar{\Delta} = 5$
$\alpha = 0.6$	1.69	1.30	1.17	1.18	1.09	0.96
$\alpha = 0.8$	0.87	0.79	0.72	0.86	0.78	0.71
$\alpha = 1.0$	0.49	0.47	0.43	0.49	0.47	0.43

Table 5.1: Value of error-fit power function parameter p when D = 0.

#### 5.3.1 Convergence

We consider three different learning rates  $b_k = 1/k^{\alpha}$ :  $\alpha = 0.6, 0.8$  and 1. The maximum delay between the measurement and estimation is taken as  $\Delta = 5$ . To highlight the convergence behavior, we initialize the hyperparameters with a fixed value 4.1, irrespective of the measurements. The SNR is chosen as 20 dB for all the results in this subsection.

#### **Uncorrelated Case**

We generate the sparse vectors from two distributions: Gaussian and Rademacher distribution. The mean squared error (MSE) in the estimated hyperparameters when  $\bar{\Delta} = 3$  are plotted in Figure 5.1a. The curves labeled Fit are the fitted curves on the error using the function:  $f(x) = ax^{-p}$  where a and p are parameters. The result for other values of  $\bar{\Delta}$  is similar, and we summarize the values of p in Table 5.1. Our observations from the results are as follows:

- Convergence: The algorithm converges to the true  $\gamma$ , and not to the other equilibrium point,  $\gamma = 0$ , in all cases. This happens even if we initialize the algorithm with very small values such as  $10^{-2} \cdot 1$ .
- Sparse vector distribution: The algorithm works equally well for both Gaussian

(which is continuous) and Rademacher distribution (which is discrete), as guaranteed by Theorem 5.2. In particular, it works for the Rademacher distribution even though it was developed by imposing a Gaussian distribution on the nonzero coefficients of the sparse vectors.

- Learning rate: The smaller the  $\alpha$ , the larger the learning rate  $b_k$ , and hence the larger the weightage given to the update term  $\text{Diag} \{ (I - D^2) T_{k|k+\Delta} - \Gamma_{k-1} \}$  in (4.40), leading to faster convergence. Since  $1/2 < \alpha \leq 1$  is required for theoretical convergence guarantee, a value of  $\alpha$  close to 1/2 ensures the fastest convergence. However, we have also observed from our experiments that  $\alpha \leq 1/2$  leads to even faster convergence. Hence, in practice, one could try using  $\alpha \leq 1/2$ , but the convergence would not be guaranteed by our analysis.
- Value of  $\overline{\Delta}$ : As  $\overline{\Delta}$  increases, the exponent p slightly decreases. This is because when  $\overline{\Delta}$  increases, the hyperparameter  $\gamma$  gets updated less frequently. Hence, a lower  $\overline{\Delta}$  improves the convergence rate and estimation accuracy, but at the cost of higher average latency and computational complexity. This is further illustrated in the following subsections.

#### Highly Correlated Case

Next, we study the convergence of our algorithm in the highly correlated case. Figures 5.1b and 5.1c show the MSE in the hyperparameter estimates when  $\overline{\Delta} = 3$ , for the fixed lag and sawtooth lag schemes, respectively. A few interesting observations from the figures are as follows:

- Correlation coefficient: As the correlation coefficient increases, the convergence becomes slower. This is because the approximation in (4.14) becomes loose as the correlation increases, as discussed in Section 4.4.2.
- Implementation scheme: We see that the convergence behavior of the fixed lag and sawtooth lag schemes are similar. However, the gap between the curves when the correlation coefficient is 0.9 and 0.95 is smaller for the fixed lag scheme compared to the sawtooth lag scheme. Further discussion about this is provided in Section 4.5.
- Learning rates: As observed in the uncorrelated case, the convergence is faster for small values of α. However, the gap between the curves for the two correlation coefficients is wider for smaller values of α. This is because as α decreases, the weightage given to the update term in (4.40) increases, and thus, it becomes more sensitive to the approximation in (4.14).

## 5.4 Summary

In this chapter, we provided a rigorous convergence analysis of the algorithm presented in Chapter 4. It is established via the simulations that the algorithm output converges to the true value, for the regimes which are not covered by the theoretical results. Hence, the chapter throughly answered the question on uniqueness of sparse solution for the model and algorithms under consideration. With this we completed the mathematical theory for the second model in this thesis. In the next chapter, we advance to the last sparsity model considered in the thesis, model SM3.

# Chapter 6

# Bayesian Learning Algorithm for Sparse Control Inputs and Observation Matrix

Answering problem Q2 and Q3 for model version SM3

In this chapter, we consider at the most sophisticated sparsity model where the initial state, sparse inputs and the measurement matrix are unknown (model SM3). The unknown measurement matrices are assume be identical for all the measurements. Further, we also assume that the transition matrix is an all zero matrix, i.e., the states are independent of each other. For this setting, we learn a measurement matrix or a dictionary from a finite set of noisy measurement vectors, such that the measurement vectors admit a sparse representation over the dictionary. This problem is refereed to as the dictionary learning (DL) problem. While several solutions are available in the literature, relatively little is known about their convergence and optimality properties. We make progress on this problem by analyzing a Bayesian algorithm for DL. Specifically, we cast the DL problem into the sparse Bayesian learning framework by imposing a hierarchical Gaussian

prior on the sparse vectors. This allows us to simultaneously learn the dictionary as well as the parameters of the prior on the sparse vectors using the expectation-maximization algorithm. The dictionary update step turns out to be a nonconvex optimization problem, and we present two solutions, namely, an alternating minimization (AM) procedure and an Armijo line search (ALS) method. We rigorously analyze the convergence properties of the solution, and show that the ALS procedure is globally convergent. We also analyze the stability of the solution by characterizing its limit points. Further, we show that the minima of the cost function of the overall algorithm are achieved at sparse solutions. As a concrete example, we consider the application of the SBL-based DL algorithm to image denoising, and demonstrate the efficacy of the algorithm relative to existing DL algorithms.

## 6.1 Background

In sparse coding, the signal of interest is represented as a linear combination of a relatively small number of columns of a properly chosen over-complete *dictionary*. The dictionary can be of two types: first, non-adaptive or predefined dictionaries like Fourier, Gabor, discrete cosine transform and wavelet [125]; and second, an adaptive or learned dictionary that is specific to the given class of signals. The use of adaptive dictionaries often leads to more compact representations and better performance in many signal processing applications ranging from image denoising [126–128], audio processing [129,130], and classification tasks [131–137], to name a few. Therefore, we are interested in the dictionary learning problem, where the objective is to find a dictionary over which a set of training signals admits a sparse representation.

Several dictionary learning algorithms for sparse coding have been proposed in the literature such as method of optimal directions (MOD) [138], K-singular value decomposition (K-SVD) [139], dictionary learning with the majorization method (DL-MM) [140], simultaneous codeword optimization (SimCO) [141], parallel atom-updating dictionary learning (PAU-DL) [142], and sequential generalization of K-means (SGK) [143]. Most of the algorithms involve an iterative procedure, alternately updating the dictionary and the sparse representation, and differ in the cost function used in the dictionary update step. To update the sparse representation, an existing standard sparse signal recovery algorithm is used.

Although the aforementioned algorithms achieve good performance, they require the knowledge of the sparsity level of the system and hand-tuning of various sensitive algorithm parameters. These limitations are handled to some extent by Bayesian algorithms [144–146]. Bayesian algorithms come with an added advantage of not requiring the knowledge of the measurement noise variance. However, the posterior distributions proposed in [144, 146] cannot be derived analytically, and a Gibbs sampler is used for Bayesian inference. The Gibbs sampling based algorithms are computationally demanding as they involve ensemble learning. To overcome this difficulty, [146] also proposes a variational Bayes' based algorithm for dictionary learning by imposing a Gaussian prior on the dictionary elements. The Gaussian prior intuitively models the boundedness of the dictionary elements and helps to obtain closed form expressions for the dictionary update. The closed form expressions results in faster convergence than the Gibbs sampling based Bayesian algorithms. Nonetheless, imposing a Gaussian prior (on a dictionary with no special structure) results in low accuracy and requires a large number of iterations to converge. Therefore, the choice of Gaussian prior still leaves room for improvement. This motivates us to develop an improved Bayesian dictionary learning algorithm which does not require the knowledge of the sparsity level, or fine-tuning of parameters, while at the same time improving on the recovery performance.

Our proposed dictionary learning algorithm is based on the sparse Bayesian learning (SBL) framework [22, 23]. In the context of sparse signal recovery, SBL is known to offer superior performance compared to algorithms based on convex relation and greedy approaches, and does not require one to tune the algorithm parameters. The basic idea of SBL is to incorporate a parameterized prior on the unknown sparse vectors that encourages sparsity. Specifically, a fictitious Gaussian prior is imposed on the sparse vectors, and the so-called hyperparameters of the Gaussian distribution are determined using Type-II maximum likelihood (ML) estimation. Our approach is different from other Bayesian dictionary learning algorithms as we impose no prior on the dictionary elements. Instead, we estimate the dictionary as a deterministic matrix with unit norm columns. The estimation method uses the expectation-maximization (EM) algorithm to simultaneously learn the parameters of the prior and the sparsifying dictionary. The dictionary update step in the EM algorithm turns out to be a quadratic optimization problem with unit norm constraints, which is a nonconvex problem because of the constraint. Since a closed form solution is not available, we propose to employ the alternating minimization (AM) procedure or Armijo line search (ALS) to solve it.

# 6.2 SBL Based Dictionary Learning

We consider a special case of LDS presented in Chapter 1, where  $\boldsymbol{D} = \boldsymbol{0}$ . Hence, we have a set of K training signals  $\boldsymbol{y}^{K} = \{\boldsymbol{y}_{k} \in \mathbb{R}^{m}\}_{k=1}^{K}$  such that  $\boldsymbol{y}^{K}$  admits a sparse representation  $\boldsymbol{x}^{K} = \{\boldsymbol{x}_{k} \in \mathbb{R}^{N}\}_{k=1}^{K}$  over an unknown dictionary  $\boldsymbol{A} \in \mathbb{R}^{m \times N}$  and is corrupted by noise, i.e.,

$$\boldsymbol{y}_k = \boldsymbol{A}\boldsymbol{x}_k + \boldsymbol{w}_k, \tag{6.1}$$

where the noise term  $\boldsymbol{w}_k \sim \mathcal{N}(\boldsymbol{0}, \sigma^2 \boldsymbol{I})$ . Our goal is to estimate the K sparse vectors and the measurement matrix  $\boldsymbol{A}$ , using the knowledge of N. To resolve the ambiguity in amplitude, we assume  $\boldsymbol{A}$  has unit norm columns. That is,  $\boldsymbol{A} \in \mathbb{O}$ , where

$$\mathbb{O} \triangleq \left\{ \boldsymbol{A} \in \mathbb{R}^{m \times N} : \boldsymbol{A}_i^{\mathsf{T}} \boldsymbol{A}_i = 1, i = 1, 2, \dots, N \right\}.$$
(6.2)

Motivated by the SBL framework [22, 23], we impose a Gaussian prior on the unknown sparse vectors  $\boldsymbol{x}_k \sim \mathcal{N}(\boldsymbol{0}, \mathsf{Diag}\{\boldsymbol{\gamma}_k\})$ , where  $\boldsymbol{\gamma}_k \in \mathbb{R}^N_+$ . Using this hierarchical model, we first compute the ML estimates  $\hat{\boldsymbol{\gamma}}_k$  and  $\hat{\boldsymbol{A}}$  of  $\boldsymbol{\gamma}_k$  and  $\boldsymbol{A}$ , respectively. These estimates, in turn, can be used to estimate the sparse vectors as  $\hat{\boldsymbol{x}}_k = \mathbb{E}\left\{\boldsymbol{x}_k | \boldsymbol{y}_k, \hat{\boldsymbol{\gamma}}_k, \hat{\boldsymbol{A}}\right\}$ .

We do not assume any structure in  $\boldsymbol{A}$  apart from the unit norm columns, and thus we do not impose any prior on  $\boldsymbol{A}$ . To obtain the ML estimates  $\hat{\boldsymbol{\gamma}}_k$  and  $\hat{\boldsymbol{A}}$ , we need to maximize  $p(\boldsymbol{y}^K; \boldsymbol{\Lambda})$ , where  $\boldsymbol{\Lambda} = \{\boldsymbol{A}, \boldsymbol{\gamma}_k; k = 1, 2, ..., K\} \in \mathbb{O} \times \mathbb{R}^{NK}_+$  is the tuple of unknown parameters.

We now develop an EM procedure to solve the ML estimation problem, equivalently, for minimizing the negative log likelihood  $-\log p(\boldsymbol{y}^{K}; \boldsymbol{\Lambda})$ . Thus, the optimization problem to be solved is  $\underset{\mathbf{\Lambda}\in\mathbb{O}\times\mathbb{R}_{+}^{NK}}{\arg\min}T\left(\mathbf{\Lambda}\right)$ , where the cost function<sup>1</sup> is

$$T(\boldsymbol{\Lambda}) \triangleq \sum_{k=1}^{K} \log \left| \sigma^{2} \boldsymbol{I} + \boldsymbol{A} \boldsymbol{\Gamma}_{k} \boldsymbol{A}^{\mathsf{T}} \right| + \boldsymbol{y}_{k}^{\mathsf{T}} \left( \sigma^{2} \boldsymbol{I} + \boldsymbol{A} \boldsymbol{\Gamma}_{k} \boldsymbol{A}^{\mathsf{T}} \right)^{-1} \boldsymbol{y}_{k}.$$
(6.3)

The EM algorithm treats the unknowns  $\boldsymbol{x}^k$  as the hidden data and the observations  $\boldsymbol{y}^K$ as known data. It is an iterative procedure with two steps: an expectation step (E-step) and a maximization step (M-step). Let  $\boldsymbol{\Lambda}^{(r)}$  be the estimate of  $\boldsymbol{\Lambda}$  at the  $r^{\text{th}}$  iteration. The E-step computes the marginal log-likelihood of the observed data  $Q^{(r-1)}$ , and the M-step computes the parameter tuple  $\boldsymbol{\Lambda}$  that maximizes  $Q^{(r-1)}$ .

**E-step:** 
$$Q\left(\mathbf{\Lambda};\mathbf{\Lambda}^{(r-1)}\right) = \mathbb{E}_{\boldsymbol{x}^{K}|\boldsymbol{y}^{K};\mathbf{\Lambda}^{(r-1)}}\left\{\log p\left(\boldsymbol{y}^{K},\boldsymbol{x}^{K};\mathbf{\Lambda}\right)\right\}$$
  
**M-step:**  $\mathbf{\Lambda}^{(r)} = \underset{\mathbf{\Lambda}\in\mathbb{O}\times\mathbb{R}_{+}^{NK}}{\operatorname{arg\,max}}Q\left(\mathbf{\Lambda};\mathbf{\Lambda}^{(r-1)}\right).$  (6.4)

Simplifying  $Q\left(\mathbf{\Lambda}, \mathbf{\Lambda}^{(r-1)}\right)$  we get,

$$Q\left(\boldsymbol{\Lambda};\boldsymbol{\Lambda}^{(r-1)}\right) = c_{K} - \frac{1}{2} \sum_{k=1}^{K} \left[\log|\boldsymbol{\Gamma}_{k}| + \operatorname{Tr}\left\{\boldsymbol{\Gamma}_{k}^{-1}\mathbb{E}\left\{\boldsymbol{x}_{k}\boldsymbol{x}_{k}^{\mathsf{T}}|\boldsymbol{y}^{K};\boldsymbol{\Lambda}^{(r-1)}\right\}\right\}\right] - \frac{1}{2\sigma^{2}} \sum_{k=1}^{K} \mathbb{E}\left\{\left(\boldsymbol{y}_{k} - \boldsymbol{A}\boldsymbol{x}_{k}\right)^{\mathsf{T}}\left(\boldsymbol{y}_{k} - \boldsymbol{A}\boldsymbol{x}_{k}\right)|\boldsymbol{y}^{K};\boldsymbol{\Lambda}^{(r-1)}\right\}, \quad (6.5)$$

where  $c_K$  is a constant independent of  $\Lambda$ . We notice that the optimization in the M-step is separable in its variables  $\Gamma_k$  and A. We get the update of  $\gamma_k$  in the M-step as follows (see [22, 23] for the detailed derivation):

$$\boldsymbol{\gamma}_{k}^{(r)} = \mathsf{Diag}\left\{\boldsymbol{\mu}_{k}\boldsymbol{\mu}_{k}^{\mathsf{T}} + \boldsymbol{\Sigma}_{(k)}\right\},\tag{6.6}$$

<sup>&</sup>lt;sup>1</sup>With a slight abuse of notation, we define  $\Gamma_k = \mathsf{Diag} \{\gamma_k\}$ , and not the  $k^{\mathsf{th}}$  column of the matrix  $\Gamma$ .

where we define the following quantities:

$$\boldsymbol{\mu}_{k} \triangleq \mathbb{E}\left\{\boldsymbol{x}_{k} | \boldsymbol{y}_{k}; \boldsymbol{\Lambda}^{(r-1)}\right\} \in \mathbb{R}^{N}$$
(6.7)

$$\boldsymbol{\Sigma}_{(k)} \triangleq \mathbb{E}\left\{ \left( \boldsymbol{x}_{k} - \boldsymbol{\mu}_{k} \right) \left( \boldsymbol{x}_{k} - \boldsymbol{\mu}_{k} \right)^{\mathsf{T}} | \boldsymbol{y}_{k}; \boldsymbol{\Lambda}^{(r-1)} \right\} \in \mathbb{R}^{N \times N}$$
(6.8)

The optimization problem corresponding the dictionary update reduces to

$$\underset{\boldsymbol{A}\in\mathbb{O}}{\operatorname{arg\,min}} \sum_{k=1}^{K} \mathbb{E}\left\{\left.\left(\boldsymbol{y}_{k}-\boldsymbol{A}\boldsymbol{x}_{k}\right)^{\mathsf{T}}\left(\boldsymbol{y}_{k}-\boldsymbol{A}\boldsymbol{x}_{k}\right)\right|\boldsymbol{y}_{k};\boldsymbol{\Lambda}^{(r-1)}\right\}.$$
(6.9)

The objective function above can be equivalently written as

$$g(\boldsymbol{A}) = -\operatorname{Tr}\left\{\boldsymbol{M}\boldsymbol{Y}^{\mathsf{T}}\boldsymbol{A}\right\} + \frac{1}{2}\operatorname{Tr}\left\{\boldsymbol{A}\left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right)\boldsymbol{A}^{\mathsf{T}}\right\}, \qquad (6.10)$$

where  $\boldsymbol{M} \in \mathbb{R}^{N \times K}$  has  $\boldsymbol{\mu}_k$  as its  $k^{\text{th}}$  column,  $\boldsymbol{Y} \in \mathbb{R}^{m \times K}$  has  $\boldsymbol{y}_k$  as its  $k^{\text{th}}$  column, and  $\boldsymbol{\Sigma} \triangleq \sum_{k=1}^{K} \left( \boldsymbol{\Sigma}_{(k)} + \boldsymbol{\mu}_k \boldsymbol{\mu}_k^{\mathsf{T}} \right) \in \mathbb{R}^{N \times N}$ . We note that there is no closed form solution to the quadratic optimization with the unit norm column constraints in (6.9). Therefore, we solve the optimization problem using two iterative schemes: AM and ALS.

### 6.2.1 Alternating Minimization (AM)

The AM procedure updates one column of  $\boldsymbol{A}$  at a time, keeping the other columns fixed. If we fix all columns of  $\boldsymbol{A}$  except the  $i^{\text{th}}$  column, the optimization problem reduces to

$$\underset{\boldsymbol{A}_{i}:\boldsymbol{A}_{i}^{\mathsf{T}}\boldsymbol{A}_{i}=1}{\operatorname{arg\,min}}\left(\sum_{k=1}^{K}-\boldsymbol{\mu}_{k}[i]\boldsymbol{y}_{k}+\sum_{j=1;j\neq i}^{N}\boldsymbol{\Sigma}[i,j]\boldsymbol{A}_{j}\right)^{\mathsf{T}}\boldsymbol{A}_{i}.$$
(6.11)

Interestingly, the above reduced optimization problem admits a unique closed form solution provided  $\sum_{k=1}^{K} \boldsymbol{\mu}_{k}[i]\boldsymbol{y}_{k} - \sum_{j=1; j \neq i}^{N} \boldsymbol{\Sigma}[i, j]\boldsymbol{A}_{j} \neq \boldsymbol{0}$ . If otherwise, we skip the update of that particular column and continue with the update of the next column. Therefore,

Algorithm 3 Dictionary Learning via SBL using AM

**Input:**  $\boldsymbol{Y} = \boldsymbol{y}^{K}$ , N and  $\sigma^{2}$ **Parameters:**  $\epsilon_1$  and  $\epsilon_2$  (stopping thresholds) Initialize:  $r = 0, A^{(0)} = 1, \gamma_k^{(0)} = 1, k = 1, 2, ..., K$ repeat for k = 1, 2, ..., K do #E-Step:  $\tilde{oldsymbol{\Phi}} = \left( \sigma^2 oldsymbol{I} + oldsymbol{A}^{(r)} oldsymbol{\Gamma}_k^{(r)} oldsymbol{A}^{(r)\mathsf{T}} 
ight)^{-1}$  $\boldsymbol{\Sigma}_{(k)} = \boldsymbol{\Gamma}_{k}^{(r)} - \boldsymbol{\Gamma}_{k}^{(r)} \boldsymbol{A}^{(r)\mathrm{T}} \tilde{\boldsymbol{\Phi}} \boldsymbol{A}^{(r)} \boldsymbol{\Gamma}_{k}^{(r)}$  $\boldsymbol{\mu}_{k} = \sigma^{-2} \boldsymbol{\Sigma}_{(k)} \boldsymbol{A}^{(r)\mathrm{T}} \boldsymbol{y}_{k}$  $r \leftarrow r + 1$  $\begin{array}{l} \#M\text{-}Step: \\ \boldsymbol{\gamma}_k^{(r)} = \mathsf{Diag}\left\{\boldsymbol{\mu}_k \boldsymbol{\mu}_k^\mathsf{T} + \boldsymbol{\Sigma}_{(k)}\right\} \end{array}$ end for #Update of A (also part of the M-Step) Initialize AM:  $u = 0, A^{(r,0)} = A^{(r-1)}$  $\mathbf{\Sigma} = \sum_{k=1}^{K} \left( \mathbf{\Sigma}_{(k)} + oldsymbol{\mu}_k oldsymbol{\mu}_k^{\mathsf{T}} 
ight), \, oldsymbol{M} = [oldsymbol{\mu}_1, oldsymbol{\mu}_2, \dots, oldsymbol{\mu}_K]$ repeat  $u \leftarrow u + 1$  $\boldsymbol{u} \leftarrow \boldsymbol{u} + \boldsymbol{1}$ for  $i = 1, 2, \dots, N$  do  $\boldsymbol{v}_{i}^{(r,u)} = \left(\boldsymbol{Y}\boldsymbol{M}^{\mathsf{T}}\right)_{i} - \sum_{j=1}^{i-1} \boldsymbol{\Sigma}[i, j]\boldsymbol{A}_{j}^{(r,u)}$  $-\sum_{j=i+1}^{N} \boldsymbol{\Sigma}[i, j]\boldsymbol{A}_{j}^{(r,u-1)}$  $\boldsymbol{A}_{i}^{(r,u)} = \begin{cases} \frac{1}{\|\boldsymbol{v}_{i}^{(r,u)}\|} \boldsymbol{v}_{i}^{(r,u)} & \text{if } \boldsymbol{v}_{i}^{(r,u)} \neq \boldsymbol{0} \\ \boldsymbol{A}_{i}^{(r,u-1)} & \text{otherwise.} \end{cases}$ end for until  $\|\boldsymbol{A}^{(r,u)} - \boldsymbol{A}^{(r,u-1)}\| < \epsilon_2$  $\boldsymbol{A}^{(r)} = \boldsymbol{A}^{(r,u)}$ **until**  $\| \boldsymbol{A}^{(r)} - \boldsymbol{A}^{(r-1)} \| + \sum_{k=1}^{K} \| \boldsymbol{\gamma}_{k}^{(r)} - \boldsymbol{\gamma}_{k}^{(r-1)} \| < \epsilon_{1}$ **Output:**  $\{\mu_k, k = 1, 2, ..., K\}$  and  $A^{(r)}$ 

the dictionary update in the  $r^{\text{th}}$  iteration of the EM algorithm reduces to the following recursions for i = 1, 2, ..., N:

$$\boldsymbol{v}_{i}^{(r,u)} \triangleq \sum_{k=1}^{K} \boldsymbol{\mu}_{k}[i] \boldsymbol{y}_{k} - \sum_{j=1}^{i-1} \boldsymbol{\Sigma}[i,j] \boldsymbol{A}_{j}^{(r,u)} - \sum_{j=i+1}^{N} \boldsymbol{\Sigma}[i,j] \boldsymbol{A}_{j}^{(r,u-1)}$$
(6.12)

$$\boldsymbol{A}_{i}^{(r,u)} = \begin{cases} \frac{1}{\|\boldsymbol{v}_{i}^{(r,u)}\|} \boldsymbol{v}_{i}^{(r,u)} & \text{if } \boldsymbol{v}_{i}^{(r,u)} \neq \boldsymbol{0} \\ \boldsymbol{A}_{i}^{(r,u-1)} & \text{otherwise.} \end{cases}$$
(6.13)

where u denotes the AM procedure iteration index. We stop the AM iterations when  $A^{(r,u)}$  converges, i.e., its change in successive iterations is small. The pseudo-code for this algorithm, which we call *dictionary learning via SBL (DL-SBL) using AM*, is provided in Algorithm 3.

*Remark:* For the special case when  $\Sigma$  is a diagonal matrix and  $\mathbf{Y}\mathbf{M}^{\mathsf{T}} \neq \mathbf{0}$ , the optimization problem (6.9) is separable in the columns of  $\mathbf{A}$ . Then, the AM procedure returns the global minimum of (6.10) in one iteration.

### 6.2.2 Armijo Line Search (ALS)

The ALS procedure updates the entire matrix A in every iteration instead of updating one column at a time [147–149]. The idea here is to translate the constrained optimization problem into an unconstrained convex optimization problem using Riemannian geometry. The algorithm continuously translates a test point in the direction of a tangent vector at the point, while staying on the manifold, until a reasonable decrease in objective function is obtained, and finally reaches a stationary point. Such a mapping is called a *retraction*,

### Algorithm 4 Dictionary Learning SBL using ALS

**Input:**  $\boldsymbol{Y} = \boldsymbol{y}^{K}$ , N and  $\sigma^{2}$ 

**Parameters:**  $\epsilon_1$  and  $\epsilon_2$  (stopping thresholds) Initialize:  $r = 0, A^{(0)} = 1, \gamma_k^{(0)} = 1, k = 1, 2, ..., K$ repeat for k = 1, 2, ..., K do #E-Step:  $ilde{\mathbf{\Phi}} = \left(\sigma^2 \boldsymbol{I} + \boldsymbol{A}^{(r)} \boldsymbol{\Gamma}_k^{(r)} \boldsymbol{A}^{(r)\mathsf{T}}
ight)^{-1}$  $\boldsymbol{\Sigma}_{(k)} = \boldsymbol{\Gamma}_{k}^{(r)} - \boldsymbol{\Gamma}_{k}^{(r)} \boldsymbol{A}^{(r)\mathrm{T}} \tilde{\boldsymbol{\Phi}} \boldsymbol{A}^{(r)} \boldsymbol{\Gamma}_{k}^{(r)}$  $\boldsymbol{\mu}_{k} = \sigma^{-2} \boldsymbol{\Sigma}_{(k)} \boldsymbol{A}^{(r)\mathrm{T}} \boldsymbol{y}_{k}$  $r \leftarrow r + 1$ #M-Step:  $\overset{''}{oldsymbol{\gamma}_k^{(r)}} = \mathsf{Diag}\left\{oldsymbol{\mu}_koldsymbol{\mu}_k^\mathsf{T} + oldsymbol{\Sigma}_{(k)}
ight\}$ end for #Update of A (also part of the M-Step) Initialize ALS:  $u = 0, A^{(r,0)} = A^{(r-1)}$  $oldsymbol{\Sigma} = \sum_{k=1}^{K} \left( oldsymbol{\Sigma}_{(k)} + oldsymbol{\mu}_k oldsymbol{\mu}_k^{\mathsf{T}} 
ight), \, oldsymbol{M} = \left[oldsymbol{\mu}_1, oldsymbol{\mu}_2, \dots, oldsymbol{\mu}_K 
ight]$ repeat  $u \leftarrow u + 1$  $\boldsymbol{Z}^{(r,u-1)} = P_{\boldsymbol{A}^{(r,u-1)}} \left( \boldsymbol{Y} \boldsymbol{M}^{\mathsf{T}} - \boldsymbol{A}^{(r,u-1)} \boldsymbol{\Sigma} \right)$ Compute the smallest integer p > 0 such that  $g\left(R_{\boldsymbol{A}^{(r,u-1)}}\left(\beta^{p}\alpha\boldsymbol{Z}^{(r,u-1)}\right)\right) - g\left(\boldsymbol{A}^{(r,u-1)}\right) \leq -c\beta^{p}\alpha\left\|\boldsymbol{Z}^{(r,u-1)}\right\|^{2}$  $\boldsymbol{A}^{(r,u)} = R_{\boldsymbol{A}^{(r,u-1)}} \left( \beta^p \alpha \boldsymbol{Z}^{(r,u-1)} \right)$ until  $\|\boldsymbol{A}^{(r,u)} - \boldsymbol{A}^{(r,u-1)}\| < \epsilon_2$  $\boldsymbol{A}^{(r)} = \boldsymbol{A}^{(r,u)}$  $\mathbf{until} \quad \|\boldsymbol{A}^{(r)} - \boldsymbol{A}^{(r-1)}\| + \sum_{k=1}^{K} \|\boldsymbol{\gamma}_{k}^{(r)} - \boldsymbol{\gamma}_{k}^{(r-1)}\| < \epsilon_{1}$ Output:  $\{\boldsymbol{\mu}_k, k = 1, 2, \dots, K\}$  and  $\boldsymbol{A}^{(r)}$ 

is denoted by  $R_A$ . For Riemannian manifolds, the line search method takes the form

$$\boldsymbol{A}^{(r,u)} = R_{\boldsymbol{A}^{(r,u-1)}} \left( \beta^p \alpha \boldsymbol{Z}^{(r,u-1)} \right), \qquad (6.14)$$

where  $\mathbf{Z}^{(r,u-1)}$  is the tangent direction of the cost function at  $\mathbf{A}^{(r,u-1)}$  and  $\beta^p \alpha$  is the Armijo step size. The constants  $\beta$  and  $\alpha$  are the parameters of the algorithm. The step size is chosen so that p is the smallest nonnegative integer that satisfies

$$g\left(R_{\boldsymbol{A}^{(r,u-1)}}\left(\beta^{p}\alpha\boldsymbol{Z}^{(r,u-1)}\right)\right) - g\left(\boldsymbol{A}^{(r,u-1)}\right) \leq -c\beta^{p}\alpha\left\|\boldsymbol{Z}^{(r,u-1)}\right\|^{2}, \quad (6.15)$$

where the scalar parameter  $c \in (0, 1)$ . The interested readers are referred to [147] for more details on ALS procedure.

We first note that the feasible set  $\mathbb{O}$  is the Cartesian product of N unit spheres in  $\mathbb{R}^m$  which are submanifolds of the Euclidean space  $\mathbb{R}^m$ . Since the Cartesian product of Riemmanian manifolds is a Riemmanian manifold,  $\mathbb{O}$  is a Riemmanian manifold. We define the Riemannian metric for  $\mathbb{O}$  as  $\langle \boldsymbol{A}, \boldsymbol{B} \rangle = \text{Tr} \{ \boldsymbol{A}^{\mathsf{T}} \boldsymbol{B} \}$  for  $\boldsymbol{A}, \boldsymbol{B} \in \mathbb{O}$ . The gradient of the objective function g in the Euclidean space is as follows:

$$\nabla g\left(\boldsymbol{A}\right) = -\boldsymbol{Y}\boldsymbol{M}^{\mathsf{T}} + \boldsymbol{A}\left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right). \tag{6.16}$$

The tangent space of the Cartesian product of manifolds is the Cartesian product of the tangent spaces. Therefore, we get the tangent space as

$$T_{\boldsymbol{A}} = \left\{ \boldsymbol{B} : \boldsymbol{A}_{i}^{\mathsf{T}} \boldsymbol{B}_{i} = 0, \forall i \right\}.$$

$$(6.17)$$

The  $i^{\text{th}}$  column of the orthogonal projection onto the tangent space is

$$P_{\boldsymbol{A}}\left(\boldsymbol{Z}\right)_{i} = \left(\boldsymbol{I} - \boldsymbol{A}_{i}\boldsymbol{A}_{i}^{\mathsf{T}}\right)\boldsymbol{Z}_{i} \tag{6.18}$$

Thus, the gradient of the restriction of g to  $\mathbb{O}$  is  $P_{\mathbf{A}}(\nabla g(\mathbf{A}))$ , and we can choose the  $i^{\mathsf{th}}$  column of the retraction as

$$R_{\boldsymbol{A}}\left(\boldsymbol{Z}\right)_{i} = \frac{\boldsymbol{A}_{i} + \boldsymbol{Z}_{i}}{\|\boldsymbol{A}_{i} + \boldsymbol{Z}_{i}\|}.$$
(6.19)

We note that the denominator  $\|\mathbf{A}_i + \mathbf{Z}_i\| \neq 0$  when  $\mathbf{Z}_i$  is the orthogonal projection onto the tangent space from (6.18). We call this algorithm DL-SBL using ALS, and summarize its pseudo-code in Algorithm 4.

### 6.2.3 Comparison of the two optimization procedures

In this subsection, we compare the AM and the ALS procedures to get insights on how to choose between them.

• Computational complexity: We assume that the multiplication of a  $p \times q$  matrix with a  $q \times r$  matrix requires  $\mathcal{O}(pqr)$  flops [117]. Each iteration of the AM procedure has a complexity  $\mathcal{O}(mKN + mN^2)$ . Typically,  $K \gg N$  for accurate estimation, and therefore the complexity order is  $\mathcal{O}(mKN)$ . Thus, the complexity is linear in m, Nand K. On the other hand, the computational complexity of the ALS procedure is also of the order  $\mathcal{O}(mKN)$ , except for the computation of the step-size parameter m. The complexity of this step depends on  $c, \beta$  and  $\alpha$ , and it is hard to determine the precise dependence. However, we have observed in our simulations that the ALS algorithm requires a larger number of iterations and a longer run time to converge compared to the AM procedure for the same initialization. Hence, the AM procedure is faster than the ALS procedure.

- Memory Requirements: Both AM and ALS procedures require  $\mathcal{O}(N^2)$  sized memory, as the largest matrix we keep track of has size  $N \times N$ .
- Parameter tuning: The AM procedure does not require tuning of any sensitive parameters. However, the ALS procedure has scalar parameters  $c, \beta$  and  $\alpha$  which determine the rate of convergence, but these parameters do not affect the recovery performance of the overall algorithm. Hence, the tuning of the parameters of ALS is not very critical.

Thus, for practical applications, we prefer AM to ALS as it is computationally less expensive and does not require tuning of any parameters. However, ALS has better theoretical convergence guarantees compared to AM algorithm, which we discuss in Section 6.3.

### 6.2.4 Comparison with other Bayesian techniques

The main differences between our algorithm and the other Bayesian algorithms in the literature are as follows:

- 1. Our algorithm does not use Gibbs sampling, unlike the algorithms in [144, 146]. Instead, we use a variational evidence framework which obviates the need for generating posterior samples, and thus our algorithm is faster. Moreover, the ensemble learning based algorithms come with no convergence guarantees. We provide rigorous convergence guarantees for our algorithm in Section 6.3.
- 2. Our algorithm is similar to the Sparse Bayesian dictionary learning with a Gaussian

hierarchical model proposed in [146] except for the prior on the dictionary. The algorithm in [146] uses a Gaussian prior on the dictionary elements to obtain a closed form expression for the EM updates. However, the choice of Gaussian prior was heuristically motivated by the fact that the entries of the dictionary are bounded. Since the dictionary is an arbitrary matrix with unit norm columns, the ideal choice of prior on the dictionary columns is a uniform distribution on the unit m-dimensional sphere. Hence, we propose to use no prior (which is equivalent to a uniform prior) on the dictionary and learn the dictionary as a deterministic unknown. Due to the better prior model used, our algorithm outperforms the one in [146] in terms of the reconstruction accuracy. The cost paid for this approach is the extra iterative procedure that is nested within the EM algorithm. Using an optimization procedure within the EM framework may appear to be more computationally demanding than an approach with closed form expressions. Nonetheless, from our simulations, we find that our algorithm requires far fewer number of iterations compared to the algorithm in [146]. Hence, the overall run time of the algorithm is much smaller.<sup>2</sup> In other words, the algorithm in this chapter is an improved version of Gaussian hierarchical model based SBL algorithm with reduced run time and higher accuracy. We corroborate these arguments through numerical simulations in Section 6.5.2 (See Figure 6.2c).

<sup>&</sup>lt;sup>2</sup>A similar observation can be found, in the context of sparse signal recovery, in [150]. Iterative reweighted  $\ell_2$  algorithms are typically slower than iterative reweighted  $\ell_1$  algorithms, even though the former admits closed form expressions in the iterations.

# 6.3 Convergence Analysis

In this section, we discuss the convergence properties of the AM and ALS procedures proposed to solve (6.9).

**Proposition 6.1** (Function value convergence). The sequences of cost function values  $\left\{g\left(\mathbf{A}^{(r,u)}\right)\right\}_{u\in\mathbb{N}}$  generated by the AM and the ALS procedures are non-increasing and convergent.

*Proof.* See Appendix D.1.

While above proposition guarantees that the cost function converges, it does not establish the convergence of the iterates. Hence, we study the convergence behavior of the iterates in the next subsections. We start with the following definition.

**Definition 6.1** (Nash equilibrium). The matrix  $\boldsymbol{A}$  with unit norm columns is said to be a Nash equilibrium point of (6.9) if

$$g(\boldsymbol{A}) \leq g\left([\boldsymbol{A}_1, \dots, \boldsymbol{A}_{i-1}, \boldsymbol{a}, \boldsymbol{A}_{i+1}, \dots, \boldsymbol{A}_N]\right), \qquad (6.20)$$

for any unit-norm vector  $\boldsymbol{a}$  and for  $i = 1, 2, \ldots, N$ .

Every column of a Nash equilibrium is optimal when other columns of the dictionary are held fixed, that is, one cannot unilaterally improve the cost function in (6.9) by updating any single column. We now proceed with our analysis of the convergence of the AM procedure in the next subsection.

### 6.3.1 AM Procedure

The iterative AM procedure can be viewed as a fixed point iteration with the update mapping dictated by the function whose stationary point is sought. The following result shows that the fixed points of the updates generated by the AM procedure are Nash equilibria of (6.9).

**Theorem 6.1** (Nash Equilibrium). Let  $G : \mathbb{O} \to \mathbb{O}$  be the update mapping of AM procedure, i.e.,  $\mathbf{A}^{(r,u+1)} = G(\mathbf{A}^{(r,u)})$ . Then, a matrix  $\mathbf{A}^*$  is a fixed point of G if and only if  $\mathbf{A}^*$  is a Nash equilibrium point of (6.9).

```
Proof. See Appendix D.2.
```

We note that the update mapping of the AM procedure does not have a closed form expression owing to the sequential, column-wise update of the dictionary. Due to this, although the above theorem characterizes its fixed points, it is hard to establish the convergence of the iterates. On the other hand, it is possible to show several interesting convergence properties of the iterates in the ALS procedure. We discuss this next.

### 6.3.2 ALS Procedure

We begin by noting that establishing convergence guarantees for the ALS procedure is challenging because the optimization problem in (6.9) is nonconvex in A. In particular, since A is constrained to lie in the set  $\mathbb{O}$ , the set of all matrices with unit-norm columns, establishing convergence requires analyzing the convergence behavior over Riemann manifolds. Existing results in this direction, e.g., [151–155], consider convex optimization problems, and very few results are known for the nonconvex case. In [156], the authors studied

the convergence of the so-called proximal algorithm applied to nonsmooth functions that satisfy the Lojasiewicz inequality around their generalized critical points. Based on this, convergence of iterative solvers for quadratic optimization of a matrix valued variable over the space of orthogonal matrices was shown in [157]. In [158], quadratic optimization over the space of unit norm vectors was studied. These results, when extended to a matrix setting, lead to a unit norm constraint on the rows of the matrix, and hence are not applicable in our case. Thus, the convergence of an ALS type procedure for a quadratic optimization problem under *unit-norm column* constraints has not been studied in the literature, and requires new analysis.

To discuss the convergence properties of the ALS procedures, we consider an equivalent unconstrained version of the optimization problem in (6.9) as follows:

$$\underset{\boldsymbol{A}}{\operatorname{arg\,min}} \operatorname{Tr} \left\{ -\boldsymbol{M}\boldsymbol{Y}^{\mathsf{T}}\boldsymbol{A} + \frac{1}{2} \left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right) \boldsymbol{A}^{\mathsf{T}}\boldsymbol{A} \right\} + \delta_{\operatorname{norm}}(\boldsymbol{A}).$$
(6.21)

Here, we define  $\delta_{\text{norm}}$  as a barrier function corresponding to the feasible region of (6.9):

$$\delta_{\text{norm}}(\boldsymbol{A}) \triangleq \begin{cases} 0, & \text{if } \boldsymbol{A} \in \mathbb{O} \\ \infty, & \text{otherwise.} \end{cases}$$
(6.22)

Also, let  $\tilde{g} : \mathbb{R}^{m \times N} \to \mathbb{R}$  denote the objective function of (6.21). The critical points of (6.9) are the points where the subgradient of  $\tilde{g}$  vanishes.

**Theorem 6.2** (Convergence of iterates). The sequence output by the ALS procedure,  $\{\mathbf{A}^{(r,u)}\}_{u\in\mathbb{N}}$ , is globally convergent.

Proof. See Appendix D.3.

The above result guarantees that the iterates of the ALS procedure converge irrespective

of the initial point. However, it does not ensure that the algorithm converges to the same point irrespective of the initialization. Such a guarantee exists only if the cost function has only one limit point. Hence, we next characterize the properties of the limits points of the sequence of iterates.

**Proposition 6.2** (Characterization of limits). The limit  $\mathbf{A}^{(r)}$  of the sequence  $\left\{\mathbf{A}^{(r,u)}\right\}_{u\in\mathbb{N}}$  generated by the ALS procedure satisfies the relation:

$$\boldsymbol{Y}\boldsymbol{M}^{\mathsf{T}} - \boldsymbol{A}^{(r)}\left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right) = \boldsymbol{A}^{(r)}\boldsymbol{L},\tag{6.23}$$

for some diagonal matrix L. Moreover,

- 1.  $\mathbf{A}^{(r)}$  is a Nash equilibrium point of (6.9) if and only if  $\mathbf{L}$  is a positive semidefinite matrix.
- 2.  $\mathbf{A}^{(r)}$  is a local minimum if and only if  $\mathbf{L} + \mathbf{\Sigma} \mathcal{D} \{\mathbf{\Sigma}\}$  is a positive semidefinite matrix. Further,  $\mathbf{A}^{(r)}$  is a strict local minimum if and only if  $\mathbf{L} + \mathbf{\Sigma} \mathcal{D} \{\mathbf{\Sigma}\}$  is a positive definite matrix.

*Proof.* See Appendix D.4.

We make the following observations from the above results:

- As in the case of the AM procedure, the update mapping of ALS is not available in closed form because of the step size selection process. However, the results characterize the fixed points of the mapping.
- The initialization  $A^{(r,0)}$  need not be a feasible point of (6.9). Because of the retraction step which projects the iterates to the feasible set, the algorithm can be initialized from any bounded matrix.

- The results are independent of the estimates from the outer iteration loop of the EM algorithm and the dimension of the dictionary. Thus, the results are applicable to any quadratic cost function of the form (6.9).
- Given A<sup>(r)</sup>, M, Y and Σ, the conditions for the Nash equilibrium and local minimum are easily verifiable.

Now, for any first order method such as the ALS procedure, the best guarantees one can obtain are that it converges to a stationary point. Further, we can determine whether the stationary point is a local minimum using the test in step 2 of Proposition 6.2. Beyond this, the only guarantee one can provide for first order methods is that of stability of the limit points. Stability implies that the algorithm converges to a limit point whenever it is initialized close enough to it. Formally, we define stability as follows:

**Definition 6.2** (Stability). Let  $G : \mathbb{O} \to \mathbb{O}$  be the update mapping of an iterative algorithm, i.e.,  $\mathbf{A}^{(r,u+1)} = G(\mathbf{A}^{(r,u)})$ . Also, we let  $G^{(u)}(\cdot)$  denote the result of u applications of G:

$$G^{(1)}(\mathbf{A}) = G(\mathbf{A}); \quad G^{(u+1)}(\mathbf{A}) = G(G^{(u)}(\mathbf{A})).$$
 (6.24)

The matrix  $\mathbf{A}^*$  said to be a stable point of the iterative algorithm if, for every neighborhood  $\mathcal{U}$  of  $\mathbf{A}^*$ , there exists a neighborhood  $\mathcal{V}$  of  $\mathbf{A}^*$  such that, for all  $\mathbf{A} \in \mathcal{V}$  and any positive integer u, it holds that  $G^{(u)}(\mathbf{A}) \in \mathcal{U}$ .

We have the following characterization of the stability of the fixed points of the ALS procedure, based on whether the fixed point is a local minimum or not.

**Theorem 6.3** (Stability). Let  $\mathbf{A}^{(r)}$  be a limit point of the sequence  $\left\{\mathbf{A}^{(r,u)}\right\}_{u\in\mathbb{N}}$  generated by the ALS procedure. Then,

- (i) If  $\mathbf{A}^{(r)}$  is not a local minimum of  $\tilde{g}$ , then  $\mathbf{A}^{(r)}$  is not a stable point of the ALS procedure.
- (ii) If  $\mathbf{A}^{(r)}$  is a strict local minimum of  $\tilde{g}$ , then the algorithm converges to  $\mathbf{A}^{(r)}$  if the initial point  $\mathbf{A}^{(r,0)}$  is sufficiently close to  $\mathbf{A}^{(r)}$ .
- Proof. See Appendix D.5.

An implication of Theorem 6.3 is that the ALS procedure converges to a local minimum of the cost function, except when the initial condition is carefully constructed to be adversarial in nature. Also, as in the previous case, the results are independent of the estimates from the outer iteration loop of the EM algorithm and the dimension of the dictionary. Thus, Theorem 6.3 is applicable to any optimization of the form (6.9).

In this section, we have analyzed the convergence properties of the inner loop in the M-step of EM algorithm. Our analysis guarantees that the optimization procedure has good converge properties. As a consequence, and by virtue of the well-known properties of the EM algorithm, DL-SBL is globally convergent.<sup>3</sup> Next, we address the question of how good the final solution is, by analyzing the global minima of the DL-SBL cost function given by (6.3).

# 6.4 DL-SBL Cost: Analysis of Minima

First, note that, in the context dictionary learning, uniqueness of the solution is defined up to an unavoidable permutation of the unit-norm columns of A and rows of X, where  $X \in \mathbb{R}^{N \times K}$  is the matrix obtained by stacking the sparse vectors  $x_k$ . We now present

<sup>&</sup>lt;sup>3</sup>Each EM iteration is guaranteed to reduce the overall cost function until a stationary point is reached.

necessary conditions for the uniqueness of the solution:

**Proposition 6.3.** Under noiseless condition,  $\sigma = 0$ , the solution to the DL problem is unique only if the following conditions are satisfied:

$$\mathsf{Rank}\left\{\boldsymbol{X}\right\} = N \tag{6.25}$$

$$Spark\{\boldsymbol{A}\} > 2 \max_{k=1,2,\dots,K} \|\boldsymbol{x}_k\|_0, \qquad (6.26)$$

where  $Spark \{ \mathbf{A} \} \triangleq \min_{\mathbf{x} \in \mathbb{R}^N, \mathbf{x} \neq \mathbf{0}} \| \mathbf{x} \|_0$  s. t.  $\mathbf{A}\mathbf{x} = \mathbf{0}$ . Further, for the special case of 1-sparse vectors,  $\max_{k=1,2,\dots,K} \| \mathbf{x}_k \|_0 = 1$ , the conditions are sufficient.

*Proof.* See Appendix D.6.  $\Box$ 

Next, we establish that the cost function in (6.3), when minimized, has the desired global minima.

**Theorem 6.4.** Suppose the tuple  $(\mathbf{A}^*, \mathbf{X}^*)$  satisfies the necessary conditions (6.25) and (6.26). Also, let  $\{\mathbf{\Gamma}_k^* \in \mathbb{R}^{N \times N}\}_{k=1}^K$  be a set of nonnegative diagonal matrices denoting the covariance matrix of the sparse vectors such that

$$\boldsymbol{x}_{k}^{*} = \boldsymbol{\Gamma}_{k}^{*1/2} \left( \boldsymbol{A}^{*} \boldsymbol{\Gamma}_{k}^{*1/2} \right)^{\dagger} \boldsymbol{y}_{k}$$
(6.27)

and 
$$0 < c < \min_{k=1,2,\dots,K} \gamma_k^*$$
 (6.28)

where  $\gamma_k^*$  is the smallest nonzero entry of  $\Gamma_k^*$  and c is a universal constant. Then, as the noise variance  $\sigma^2 \to 0$ , the global minimum of (6.3) is achieved at  $\left( \mathbf{A}^* \mathbf{P}, \{\mathbf{P} \Gamma_k^* \mathbf{P}\}_{k=1}^K \right)$  where  $\mathbf{P}$  is a signed permutation matrix.

Proof. See Appendix D.7.

We note that the sparsest solution of (6.3) is  $(\mathbf{A}^*, \mathbf{X}^*)$  due to (6.26). Although we assume that the necessary conditions (6.25) and (6.26) hold, the theorem holds true under the mild condition that

$$\max_{k=1,2,\dots,K} \|\boldsymbol{x}_k\|_0 < m.$$
(6.29)

However, under the above condition, uniqueness is not guaranteed, i.e., solutions with suboptimal sparsity may also globally minimize the cost function.

We know that the DL problem is NP-hard [159]. Thus, it is not surprising that the cost function obtained using SBL framework may have multiple local minima. Nonetheless, extending the results of the original SBL algorithm on sparse recovery [23], we can show that all the local minima of the function are achieved at sparse solutions.

**Theorem 6.5.** Every  $\gamma_k$  corresponding to the local minimum of the DL-SBL cost function (6.3) is at most m-sparse, regardless of the value of noise variance  $\sigma^2$ .

*Proof.* See Appendix D.8.

To sum up, in this section, we proved a set of results related to the minima of the optimization problem solved by the DL-SBL algorithm, complementing the results we presented in the previous section. Combining the two sets of results, it seems reasonable that the algorithm may converge to the maximally sparse solution and the global minimum. Our empirical results in the next section support this intuition.

# 6.5 Simulation Results

We use the following simulation setup to evaluate the performance of the algorithms and validate the theoretical convergence results in Section 6.5.1 and Section 6.5.2. We generate

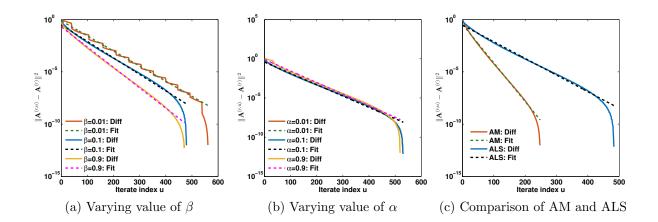


Figure 6.1: Convergence of ALS procedure ((a), (b)) and comparison with AM (c), with K = 1000, m = 20, N = 60, s = 6, and SNR = 20 dB, for the first iteration of EM algorithm.

sparse signals of length N = 60, each with s = 6 nonzero entries. The locations of nonzero coefficients are chosen uniformly at random, and the nonzero entries are independent and identically Gaussian distributed with zero mean and unit variance. The length of measurement vector is chosen as m = 20, and SNR = 20 dB. The columns of dictionary matrix A are drawn uniformly from the surface of the *m*-dimensional unit hypersphere [160].

### 6.5.1 Convergence

To study the convergence of the AM procedure, we take size of training data set as K = 1000. We look at the first iteration (r = 1) of the EM algorithm because that requires the maximum number of inner iterations to converge, and thus illustrates the convergence behavior well.

Figure 6.1 shows the  $\ell_2$  squared norm of the difference between the iterates and the limit point, given by  $\|\mathbf{A}^{(1,u)} - \mathbf{A}^{(1)}\|^2$ , of the AM and ALS procedures under different settings. We set  $\beta = 0.1$  for Figure 6.1 and  $\beta = \alpha = 0.1$  for Figure 6.1c. The curves labeled Diff

Setting		Fit parameters		no. of	run
		a	b	iterations	time (s)
$\alpha = 0.1$	$\beta = 0.01$	-0.034	-0.093	565.04	1.33
	$\beta = 0.1$	-0.036	-1.102	490.09	1.5
	$\beta = 0.9$	-0.044	-1.554	480.63	13.68
$\beta = 0.1$	$\alpha = 0.01$	-0.036	-1.118	494.26	1.55
	$\alpha = 0.1$	-0.036	-1.102	490.09	1.50
	$\alpha = 0.9$	-0.037	-0.226	486.60	1.51

Table 6.1: Comparison of ALS convergence behavior with varying step size parameters  $\beta$  and  $\alpha$ 

Algo.	Fit parameters		no. of	run
	a	b	iterations	time (s)
AM	-0.0427	-0.4603	248.95	0.5828
ALS	-0.0361	-1.1022	490.09	1.5020

Table 6.2: Comparison of ALS and AM convergence behavior and Fit correspond to the curves obtained via numerical experiments and by fitting the function  $f(u) = \exp(au + b)$  on the values, respectively, where a < 0 and b are parameters of the curve. The values of the parameters averaged over 100 experiments are listed in Table 6.1 and Table 6.2. Our observations are as follows:

- Rate of convergence: From Figure 6.1, we see that the curve is well approximated using an exponential function for moderate values of iteration number. Further, the tail of the curve exhibits a faster-than-exponential decay. Interestingly, both AM and ALS procedures exhibit the same behavior for all choices of β and α.
- Step size parameter  $\beta$ : In the backtracking step of the ALS procedure, we need to evaluate the optimal step size by searching over the step sizes  $\beta \alpha, \beta^2 \alpha, \beta^3 \alpha, \ldots$  For smaller  $\beta$ , the number of iterations to reach the stopping threshold decreases, as

the search domain is larger. However, as  $\beta$  increases, the optimal value of p also increases, which results in a higher run time as it requires p function evaluations and comparisons. Thus, using  $\beta \leq 0.1$  strikes a good balance between run time and the number of iterations.

- Step size parameter α: For the ALS procedure, α does not have any effect on rate of convergence or run time. This is because, the size of the discrete search in the backtracking step does not depend on the value of α.
- Comparison of AM and ALS: From Figure 6.1c and Table 6.2, we see that the AM algorithm converges faster than ALS and requires fewer number of iterations for the same stopping threshold. Therefore, the AM procedure is computationally more attractive than the ALS in practice, although the ALS procedure comes with stronger theoretical convergence guarantees.

*Remark:* The number of iterations required by the procedure dramatically reduces as r increases. All the plots shown here correspond to r = 1. However, for r > 10, only about 2-4 iterations are required for the inner optimization, making it computationally very efficient.

### 6.5.2 Performance of the Algorithms

In this subsection, we compare the performance of our algorithms with other popular algorithms in literature. Here, we do not show separate curves for DL-SBL using the ALS and AM algorithms, as their performances are virtually identical.

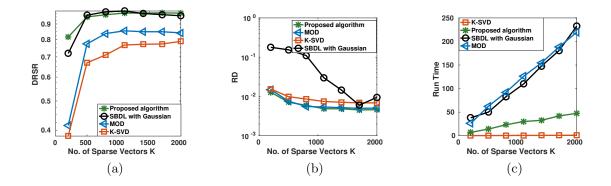


Figure 6.2: Comparison of DL-SBL with KSVD, MOD and Gaussian hierarchical model based SBL algorithms when the number of input vectors is varied. The performance of DL-SBL is superior to the other three algorithms.

### Synthetic Data

We first use the same setup as that of Section 6.5.1. We let  $\hat{x}_k$  and  $x_k$  denote the estimate and true value of the sparse vector, respectively and  $\hat{A}$  and A denote the estimate and true value of the dictionary, respectively. We use the following metrics evaluating the performance.

- (i) Dictionary recovery success rate (DRSR) [139], which is the fraction of successfully recovered columns of the dictionary. A column is said to be successfully recovered if the magnitude inner product between a column in the true dictionary and the estimated column exceeds 0.99.
- (ii) Relative distortion (RD) [141], defined as:

$$\mathrm{RD} \triangleq \frac{\sum_{k=1}^{K} \|\hat{\boldsymbol{A}}\hat{\boldsymbol{x}}_{k} - \boldsymbol{A}\boldsymbol{x}_{k}\|^{2}}{\sum_{k=1}^{K} \|\boldsymbol{A}\boldsymbol{x}_{k}\|^{2}}.$$
(6.30)

(iii) Run time, which is the time required to complete the computations. It measures the computational complexity.



(a) Original image



(d) SimCO, PSNR = (e) DL-MM, PSNR =28.64 dB, run time = 58.7 s



(b) Corrupted image, PSNR = $20 \mathrm{~dB}$ 



(c) DL-SBL, PSNR = 28.96 dB, run time = 105.7 s



28.54 dB, run time = 98.7 s



(f) KSVD, PSNR =28.34 dB, run time = 76.7 s



(g) SBDL with Gaussian, PSNR = 28.24 dB,run time = 225.95 s



(h) BiG-AMP, PSNR = 28.02 dB,run time = 330.4 s



(i) SGK, PSNR =27.44 dB, run time = 82.5 s



(j) PAU, PSNR =27.44 dB, run time = 84.5 s



(k) MOD, PSNR =27.42 dB, run time = 78.90 s

Figure 6.3: An example of image denoising using different dictionary learning algorithms with K = 1000, m = 64, and N = 256. We see that all algorithms improve the visual quality of the noisy image.

We refer to the DRSR and RD metrics jointly as the recovery performance of the algorithm. These two metrics are equally important due to non-uniqueness of the solution. Any solution of the form  $\{AP, Px_k, k = 1, 2, ..., K\}$ , where P is a signed permutation matrix<sup>4</sup> is a solution to the dictionary learning problem. Thus, the error metric  $\frac{\|A-\hat{A}\|^2}{\|A\|^2}$  does not account for the inherent non-uniqueness of the solution. Hence, we use DRSR as a measure of how well the dictionary is recovered. Now, suppose that the algorithm outputs a matrix whose all columns are the same as the first column of the true dictionary. Then, the corresponding DRSR = 1 even though the recovery is poor. Therefore, we also look at RD which is a measure of how well the recovery performance depends on both DRSR and RD.

Figure 6.2 compares the proposed algorithm with KSVD [139], MOD [138] and Gaussian hierarchical model based SBL algorithms [146] (labeled as SBDL with Gaussian).<sup>5</sup> Clearly, the performance of all the algorithms improve with K, as more information about the dictionary is available to the algorithm. Our proposed algorithm outperforms the other algorithms in terms of both DRSR and RD. However, the run time demanded by our algorithm is larger than K-SVD, but it is lower than the other two algorithms.

Comparing DL-SBL with the Gaussian hierarchical model based SBL [146], we see that the DRSR performance of both algorithms is similar for K > 400. However, the RD of Gaussian hierarchical model based SBL is worse than that of our algorithm. Moreover, the run time of Gaussian hierarchical model based SBL is much higher than our algorithm for the same convergence criteria. This observation agrees with the intuitive explanation

 $<sup>{}^{4}</sup>$ A matrix is said to be a signed permutation matrix if it has exactly one nonzero entry which is either 1 or -1 in each row and each column.

<sup>&</sup>lt;sup>5</sup>The other algorithms are highly sensitive to parameter-tuning, and hence we found it hard to pick the right parameters to get performance comparable to the ones presented.

presented in Section 6.2.4.

### **Image Denoising**

We next consider the application of DL to the problem of image denoising. Here, the goal is to remove zero-mean white and homogeneous Gaussian additive noise from a given image [126]. We use the  $512 \times 512$  image "Barbara" shown in Figure 6.3a for all the experiments. We learn the dictionary using K = 1000 uniformly randomly chosen blocks of size  $m = 8 \times 8 = 64$  pixels out of 255025 corrupted image blocks. We note that the number of patches we use is an order of magnitude smaller than  $K = 6.2 \times 10^4$  used in [126]. The length of the sparse vectors N is taken as 256. For the competing algorithms, once the dictionary is learned, the complete image is reconstructed using the orthogonal matching pursuit algorithm with the corrupted image and the learned dictionary as inputs. The corrupted image with peak SNR, PSNR = 20 dB is shown in Figure 6.3b, and the images reconstructed using the learned dictionary by the following algorithms are shown in Figure 6.3c-Figure 6.3k: Simultaneous codeword optimization (SimCo) [141]; Dictionary learning with the majorization-minimization method (DL-MM) [140]; K-singular value decomposition (K-SVD) [139]; Gaussian hierarchal model based SBL (SBDL with Gaussian) [146]; Bilinear generalized approximate message passing (BiG-AMP) [145]; Sequential generalization of K-means (SGK) [143]; Parallel atom-updating dictionary learning (PAU-DL) [142] Method of optimal directions (MOD) [138]. The PSNR values of reconstructed image and run time required are also given in the caption of each reconstructed image. We see that all dictionary learning algorithm improve the image PSNR, but the largest gain is obtained for our DL-SBL algorithm, at the cost of slightly higher run time. Also, we see that our algorithm outperforms the Gaussian hierarchal model based SBL both in terms of accuracy and run time.

### 6.6 Summary

In this chapter, we analyzed a Bayesian algorithm for jointly recovering a dictionary matrix and a set of sparse vectors from a noisy linear underdetermined training set. We developed the algorithm using the SBL framework, and implemented it using the EM algorithm, with the dictionary matrix and the variances of the entries of the sparse vectors as unknown parameters. The EM algorithm requires one to solve a non-convex optimization problem in the M-step, which we tackled using an AM or ALS procedure. We compared the AM and ALS procedures in terms of their computational complexity and memory requirements. We also provided a rigorous convergence analysis of the proposed optimization procedures. Further, by direct analysis of the cost function involved, we showed that the DL-SBL algorithm is likely to output the sparsest representation of the input vectors. We empirically showed the efficacy of our algorithm compared to existing algorithms, when applied to the image denoising problem. The chapter dealt with the second and the third questions Q2 and Q3 on algorithm development and recovery guranteens for the third level of sparsity model SM3. So, we have looked at all questions posed in Chapter 1 in great depth of technical detail in Chapter 2 to Chapter 6. Before, we conclude the main idea of the thesis, in the next chapter, we shift gears and provide a new application of structured sparse signal recovery in the context of structural health monitoring.

# Chapter 7

# Anomaly Imaging for Structural Health Monitoring

### Exploiting clustered sparsity

In this chapter, we take a small deviation from the main theme of the thesis and present a structured sparse signal recovery algorithm to a signal model which is not coupled with LDS. The goal here is to develop an algorithm for anomaly imaging for structural health monitoring. Under this model, we are given a set of sensor measurements which are linearly related to an unknown anomaly map. We use a Bayesian framework to explicitly account for both sparsity and cluster pattern structures that are typical of structural anomalies. Hence, the algorithm of this chapter provides excellent reconstruction accuracy by incorporating the available prior information on the anomaly map. Experimental results on a unidirectional composite plate confirms that the algorithm of this chapter outperforms two competing methods in terms of reconstruction accuracy.

# 7.1 Background

Many critical infrastructures like aircraft, load bearing walls and oil pipelines use fiber reinforced laminate composite materials. Although composite materials are lightweight, strong, and possess excellent fatigue and corrosion resistance, many inter-laminar defects may show no visible evidence [161, 162]. To ensure the integrity of the structure for safe and efficient operation, it is desirable to embed an inspection system within the structure [163]. An active structural health monitoring (SHM) system consists of an array of transducers that can excite and sense wave propagation within the thickness of the structure. The system periodically excites the structure using the transducers sequentially. The resulting waveforms are collected at the remaining transducers which act as sensors. The collected signal is compared to a set of baseline measurements acquired from the structure prior to deployment. The differences between the two signals are characterized using an anomaly metric. The anomaly metric for all actuator-sensor pairs are used to detect and characterize structural damage.

Several algorithms for anomaly mapping have been presented in the literature. Malyarenko and Hinders [164] described a tomography-based approach to image flaws and corrosion on metallic structures. This approach employs the time difference of arrival of the wave between an actuator and a sensor as the measure of the average properties of the actuator-sensor path. A similar approach is studied for composite plates in [165]. Later, Prasad *et al.* [166] successfully located holes on crossply and quasi-isotropic plates using an iterative algebraic reconstruction technique (ART). The algorithm uses the rootmean-square value between the sensor signals and the corresponding baseline signals as the basis for the reconstruction algorithm. Gao *et al.* [167] proposed the reconstruction algorithm for probabilistic inspection of damage (RAPID) for damage mapping. Although low in complexity, the algorithm design does not consider any particular signal structure associated with the anomaly map. Recently, Zoubi and Mathews [168, 169] developed an anomaly mapping algorithm that uses the sparse nature of damage distribution on structures. However, the sparsity model in the anomaly map is significantly more structured and exhibits a two-dimensional clustered pattern. Therefore, we present a new solution exploiting the two-dimensional clustered sparsity pattern as a prior information to the anomaly map reconstruction problem.

This chapter presents a new algorithm for anomaly imaging, based on ART and the two-dimensional pattern coupled sparse Bayesian learning algorithm. The algorithm takes a set of Lamb wave measurements collected on the structure as input, and outputs an anomaly map from which one can estimate the boundaries of damage on the structure. To the best of our knowledge, this chapter is the first to exploit the two-dimensional clustered sparsity pattern of the anomaly map to improve damage mapping accuracy. Experimental results on a unidirectional composite plate show that the method of this chapter provides more accurate estimates of the damage boundaries than two competing algorithms.

# 7.2 System Model

We consider an SHM system that uses a set of m transducers where each transducer can act as a wave sensor or a wave actuator, as needed. The structure is excited sequentially by the transducers to obtain an anomaly metric (index) for each signal path connecting the actuator-sensor pairs. The goal is to estimate the anomaly map of the structure, using the K = m(m-1) damage indices thus obtained.

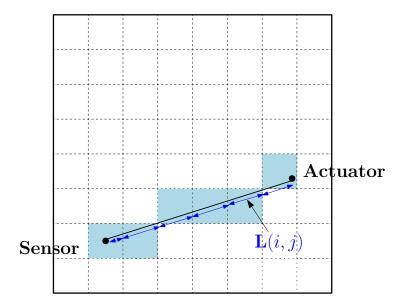


Figure 7.1: The figure shows  $i^{\text{th}}$  sensor-actuator pair and the direct path between them. The pixels in blue correspond to the nonzero entries of  $j^{\text{th}}$  row of L, and the non-zero value equals the length of the path overlapping the pixel.

To reconstruct the anomaly map, we employ a grid architecture where the spatially continuous map is discretized into N cells or pixels using a grid as shown in Figure 7.1. The damage value associated with each pixel indicates the state of the corresponding part of the structure. Thus, our goal reduces to computing the map value at each pixel using the measured anomaly indices. The mathematical model relating anomaly map values and the damage indices is adopted from the ART framework. Here, the damage indices are assumed to be a linear combination of the pixel values weighted with the length of the direct path between the actuator-sensor pair that crosses the pixel [170]. For instance, in Figure 7.1, the damage index for the signal path between the sensor-actuator pair depends on the pixels which are marked in blue. This linear relationship can be written as

$$\boldsymbol{y} = \boldsymbol{L}\boldsymbol{x},\tag{7.1}$$

where  $\boldsymbol{y} \in \mathbb{R}^{K}$  is obtained by stacking the damage metrics into a column vector. The column vector  $\boldsymbol{x} \in \mathbb{R}^{N}$  is the vector of the pixel values or the vectorized version of the discretized anomaly map. The  $(i, j)^{\text{th}}$  entry of the matrix  $\boldsymbol{L} \in \mathbb{R}^{K \times N}$  is the length of  $i^{\text{th}}$  line segment that overlaps pixel j, as illustrated using Figure 7.1. Hence, the map recovery problem is equivalent to the recovery of  $\boldsymbol{x}$  from (7.1) when  $\boldsymbol{y}$  and  $\boldsymbol{L}$  are known. In the next section, we present the algorithm to recover the discretized anomaly map which utilizes the sparse and clustered structures associated with the unknown map.

# 7.3 Map Recovery Algorithm

We recover the unknown  $\boldsymbol{x}$  by exploiting two underlying structures in the signal:

- 1. Anomaly areas on the structure are usually small compared to its overall size, which makes  $\boldsymbol{x}$  naturally sparse.
- The anomaly areas occupy a small continuous region of the structure. Therefore, the anomaly map exhibits two-dimensional cluster patterns, also known as blocksparsity.

Several recovery algorithms that exploit block-sparsity have been proposed in the literature. Examples include block-OMP [171], mixed  $\ell_2/\ell_1$  norm-minimization [172], group LASSO [173] and block-sparse Bayesian learning [174]. These algorithms require *apriori* knowledge of the block boundaries. However, in our case, the two-dimensional cluster pattern is not known as it depends on the unknown location and shape of the anomaly area. Recently, a new approach has been proposed to tackle the difficulty of unknown block boundaries using the sparse Bayesian learning (SBL) framework [175, 176]. Moreover, the SBL-based algorithms are known to have superior performance compared to convex relaxation or greedy approaches. Hence, we use the *pattern-coupled (PC)* SBL algorithm to exploit the two-dimensional block-sparse structure.

In the SBL framework, we use a fictitious prior on the unknown signal which promotes the underlying signal structures. To account for the two-dimensional block-sparse structure, a pattern-coupled Gaussian hierarchical prior is imposed on  $\boldsymbol{x}$ . The use of hyper-parameters associated with each entry of  $\boldsymbol{x}$  in the hierarchical Gaussian prior is known to promote sparsity. In addition, two-dimensional block-sparse structure is captured by imposing dependency between the hyper-parameters associated with each entry and that of its neighboring entries:  $\boldsymbol{x} \sim \mathcal{N}(\boldsymbol{x}; \boldsymbol{0}, \boldsymbol{\Gamma})$ , where  $\boldsymbol{\Gamma} \in \mathbb{R}^{N \times N}$  is a diagonal matrix with diagonal entries:

$$\boldsymbol{\gamma}_i^{-1} = \boldsymbol{\alpha}_i + \beta \sum_{j \in \mathcal{B}(i)} \boldsymbol{\alpha}_j, \tag{7.2}$$

Here,  $\boldsymbol{\alpha} \in \mathbb{R}^N$  is a vector of non-negative hyperparameters,  $\beta \in [0, 1]$  is the coupling parameter, and  $\mathcal{B}(i)$  is the set of neighboring entries of  $\boldsymbol{x}_i$  in the two-dimensional signal. Due to the interdependence on the priors, the entry  $\boldsymbol{x}_i$  is driven to zero if  $\alpha_i$  or any of its neighboring hyperparameters goes to infinity. The shared hyperparameters enables the prior to flexibly model any block-sparse structure, without pre-specifying the block boundaries.

Using the model in (7.2), as in conventional SBL, we use type II maximum likelihood estimation for  $\boldsymbol{x}$ . In other words, we first estimate the hyperparameters of the imposed prior which in turn yields an estimate of the sparse  $\boldsymbol{x}$ . The hyperparameters are obtained using the expectation-maximization (EM) algorithm, where the sparse vectors are treated as hidden variables. We summarize the pseudo-code for anomaly mapping in Algorithm 5. For detailed derivation of the PC-SBL algorithm, please refer to [175, 176].

Algorithm 5 The PC-SBL Recovery Algorithm

Input:  $\boldsymbol{y}$  and  $\boldsymbol{L}$ Parameters: Coupling coefficient  $\beta$ , Tolerance  $\epsilon$ Initialize:  $\boldsymbol{\alpha}^{(0)}, \sigma^{2(0)}, c = d = 10^{-4}$ while  $\|\boldsymbol{\alpha}^{(r)} - \boldsymbol{\alpha}^{(r-1)}\| \le \epsilon$  and  $|\sigma^{2(r)} - \sigma^{2(r-1)}| \le \epsilon$  do for r = 1, 2, ... do  $\boldsymbol{\gamma}_i = \left(\boldsymbol{\alpha}_i^{(r-1)} + \beta \sum_{j \in \mathcal{B}(i)} \boldsymbol{\alpha}_j^{(r-1)}\right)^{-1}, i = 1, 2, ..., N$   $\boldsymbol{\Sigma} = (\sigma^{-2(r-1)} \boldsymbol{L} \boldsymbol{L}^{\mathsf{T}} + \text{Diag} \{\boldsymbol{\gamma}\})^{-1}$   $\boldsymbol{\mu} = \sigma^{-2(r-1)} \boldsymbol{\Sigma} \boldsymbol{L}^{\mathsf{T}} \boldsymbol{y}$   $\sigma^{2(r)} = K + 2c \left(2d + \|\boldsymbol{y} - \boldsymbol{L}\boldsymbol{\mu}\|^2 + \text{Tr} \{(\} \boldsymbol{L} \boldsymbol{L}^{\mathsf{T}} \boldsymbol{\Sigma})\right)^{-1}$ for i = 1, 2, ..., N do  $\boldsymbol{\alpha}_i^{(r)} = 2 \left(\boldsymbol{\mu}_i^2 + \boldsymbol{\Sigma}_{ii} + \beta \sum_{j \in \mathcal{B}(i)} \boldsymbol{\mu}_j^2 + \boldsymbol{\Sigma}_{jj}\right)^{-1}$ end for end for end while Output:  $\boldsymbol{x} = \boldsymbol{\mu}$ 

Although (7.1) does not assume any model mismatch, PC-SBL can handle noisy measurements. The PC-SBL based reconstruction can also be applied to other ART-based tomographic imaging methods such as MRI, for cancer detection.

### 7.4 Experimental Results

The experiments described here were conducted on a 41'' wide, 40'' long and 0.1'' thick, unidirectional composite panel made out of 8 IM7/8552 carbon fiber plies. Thirty two piezoelectric transducers were attached to the plate covering the middle  $33'' \times 32''$  region of the plate. The excitation signal used was a linear chirp with bandwidth [150,300 kHz] and the resulting waveforms were acquired with a  $2 \times 10^6$  samples/second sampling rate. First, the baseline signals were collected before impact damage was introduced into the structure. Then, we impacted the structure on different locations to create damage, and the test signals were acquired after each impact experiment. Other computational details are as follows:

*Choice of damage index:* We first applied a mode decomposition algorithm based on cross Wigner-Ville-distribution of the received signal. The anomaly indices were computed using the extracted first arriving mode of the measured signal and the baseline signal, as proposed in [168, 169, 177].

*Multi-grid architecture:* The virtual grid on the structure is assumed to be rectangular with  $22 \times 22$  pixels. Since the choice of the grid structure is arbitrary, we used the multi-grid architecture to improve the reconstruction accuracy. We reconstructed the map using 20 different grids on the structure, then, interpolated them to obtain a map on a high-resolution grid. The interpolated grids were  $200 \times 200$  pixels. The final estimate of the anomaly map was obtained by averaging over these 20 maps. Further details on the multi-grid averaging approach can be found in [178].

Algorithm tuning: From our experiments, we have seen that the choice of parameters  $\beta$ and  $\epsilon$  of Algorithm 5 is not critical. For the results presented here, we choose  $\beta = 1$  and  $\epsilon = 10^{-6}$ . Also, in the algorithm, we adopt a pruning operation for faster convergence. At each iteration, we pruned those small coefficients associated with hyperparameters  $\alpha_i$ greater than  $10^{11}$  times the minimum value.

*Estimation of anomaly boundaries:* The damage area was estimated as the locations on the structure where the estimated map value was greater than some threshold, and the threshold is calculated using training data.

Recovery accuracy metric: We used the Sørenson-Dice index (also known as F1 score), which computes the correlation between two data sets A and B as  $\frac{2|A \cap B|}{|A|+|B|}$ . Here, the anomaly map obtained using an A-scan device (manual non-destructive evaluation technique) was used as the ground truth.

To illustrate the performance of our algorithm, we compare the algorithm of this chapter based on PC-SBL with two state-of-the-art algorithms: a least-squares (LS) based damage mapping algorithm [178] and a LASSO based damage mapping algorithm [168, 169]. Figure 7.2 shows the reconstructed map of the composite plate obtained using three algorithms after impact experiments. Each row corresponds to an experiment, and each column corresponds to an algorithm. The blue outlines in the maps represent the boundaries of the anomaly estimated using A-scan. The extent of the anomaly estimated by each algorithm is shown in red. We also provide the Sørenson-Dice similarity index of the estimated boundaries in the caption of each figure.

From Figure 7.2, we see that, compared to the LS-based method (first column), the algorithm of this chapter (last column) has fewer false alarms. Also, compared to the LASSO based method (middle column), the algorithm of this chapter gives better estimate of the damage boundaries, which is evident from the Sørenson-Dice index. Overall, the results clearly indicate that the map reconstructed by the PC-SBL algorithm identifies the true anomalies in the structure more closely compared to the other approaches.

### 7.5 Summary

This chapter presented a new algorithm for anomaly map reconstruction for health monitoring of composite structures. We utilized the two-dimensional clustered sparse structure associated with structural damage to present a new map reconstruction algorithm. Using a data set obtained from impact experiments, we demonstrated the superiority of our

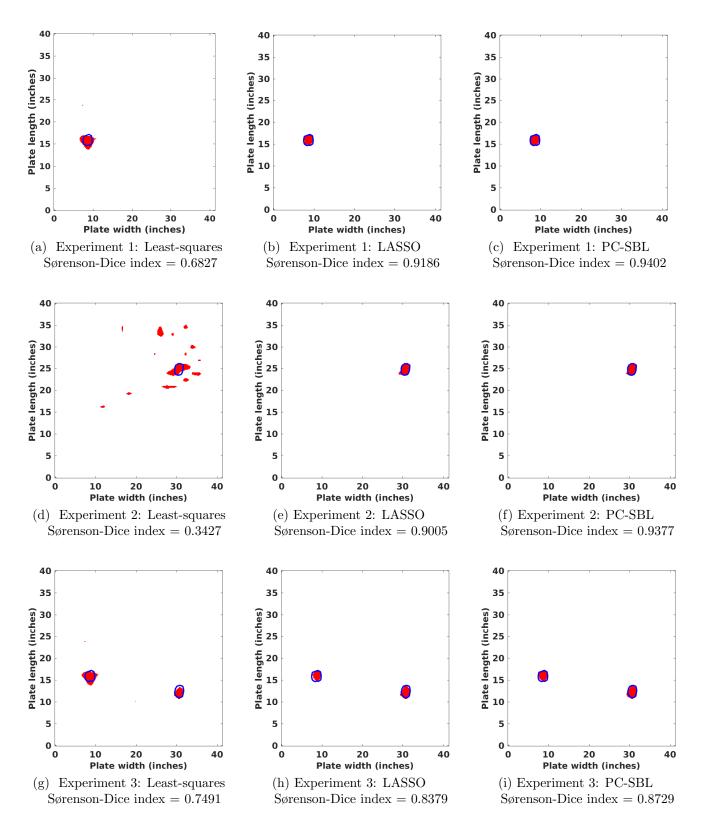


Figure 7.2: Comparison of the damage outlines estimated by three different algorithms along with corresponding Sørenson-Dice similary index. The method of this chapter provides the best results out of the three methods.

algorithm compared to two competing algorithms available in the literature. The results showed that exploiting any underlying structure of the damage improves the map reconstruction accuracy. Hence, the chapter covered a problem which is not related to LDS, but connected the sparsity property of the a linear system, which is the central theme of the thesis. Finally, in the next chapter, we summarize the main points and discuss some broader implications of the research presented in the thesis.

# Chapter 8

# Conclusions

#### Summarizing the memorable takeaways and looking ahead

The thesis presented new theoretical results and algorithms concerning the estimation of state vectors in LDS with sparsity constraints. This final chapter summarizes all the findings presented so far, and the new insights the thesis has contributed. We also provide some exiting questions that the results raise and certain paths that seem to be promising for future work.

## 8.1 Summary of Contributions

We studied the sparse signal recovery problem under three different models associated with LDS. For each model, we investigated on three important aspects: conditions for the existence of a solution, low-complexity recovery algorithm development, recovery guarantees. We list the specific contributions associated with each model in the following subsections.

#### 8.1.1 SM1: Known inputs and observation matrix

Under this model, we considered the recovery of a sparse initial state using the knowledge of measurements and inputs and other system matrices. The recovery problem, in this case, is similar to a standard compressive sensing problem, except that the measurement matrix has an additional structure that arises due to the underlying LDS model. Hence, none of the known standard recovery guarantees are directly applicable to this problem. Moreover, the existing literature on this problem is applicable only for the special case when the system transfer matrix is a unitary matrix. Therefore, we derived guarantees on recoverability of the sparse initial state of a linear dynamical system under a stochastic setting for two cases: (i) the observation matrices at different time instants are independent and identically distributed (i.i.d.) subgaussian random matrices; (ii) the observation matrices at all time instants are identical, and equal to a subgaussian random matrix. We do not impose any restriction on the system matrices, and in particular, the system transfer matrix can be any arbitrary matrix. Our result reveals that when the system transfer matrix is arbitrary, the measurement bound for recovery depends on the inverse of the condition number of the matrix. This relationship was not known earlier because when the transfer matrix is unitary, the condition number is unity. Furthermore, for the regime where our results are comparable to the existing result, our measurement bounds are tighter. In essence, our results which bring out the dependence of condition number of transfer matrix, is more general and stronger than the existing results.

#### 8.1.2 SM2: Unknown inputs and Known observation matrix

Under this model, we looked at the recovery of a set of sparse control inputs using the knowledge of the measurements and system transfer matrices. We first consider the conditions on the system for the existence of a solution. This problem has already been studied in the context of controllability of actuator scheduling in the control theory literature, and extension of Kalman rank test gives the necessary and sufficient conditions for the existence of solution. However, the test can not be used directly due to its combinatorial in nature, and some approximations of this test are used in practice. Hence, we developed an alternate test for checking the existence of the solution which is non-combinatorial and can be completed in polynomial time. Our procedure is equivalent to the Kalman rank based conditions, but it comes with the added advantage of low complexity. Moreover, we do not impose any assumptions on the structure of the system transfer and on the input matrices. Hence, our results are quite general. It is interesting to note that such a non-combinational test is not available for a canonical sparse recovery problem, and the special structure in the measurement induced by the LDS made it possible to develop this simple test. We have also derived bounds on the minimum number of input vectors required to ensure the existence of a solution, and extension of Kalman decomposition algorithm for sparse inputs. On the whole, the key contribution from this part of the work is the first-ever low-complexity controllability test for LDS with sparse inputs.

Next, we addressed at the recovery of jointly sparse control inputs using the SBL framework. The choice is motivated by the superior performance and parameter hand-tuning free nature of SBL approach. Moreover, it naturally extends to incorporate the temporal correlation structure in the signal model that arises in LDS. However, its complexity and memory requirements increase with the number of measurements to be processed, which limits its practical application. Our algorithm overcomes this drawback and is computationally efficient while retaining the good performance of SBL. We present a non-iterative online algorithm for recovering temporally correlated sparse vectors, which resulted in low computational complexity and memory requirements. We presented two schemes for implementation: a fixed lag scheme and a sawtooth lag scheme, and discuss an efficient method to initialize the algorithm. Further, we demonstrate the efficacy of the algorithm using OFDM wireless channel estimation.

Even though the proposed algorithm performed better than the existing algorithms in terms of recovery performance and run time, the key novelty of the work is in the analysis of the algorithm. To the best of our knowledge, none of the online algorithms for the sparse signal recovery comes with theoretical guarantees. Moreover, the offline-counter part of our algorithm, KM-SBL is known to have theoretical guarantees only for the special cases when the sparse vectors are the uncorrelated and perfectly correlated. We analyzed the proposed algorithms for these two special cases and established strong convergence guarantees. However, there is an important difference between the offline and online algorithm guarantees: offline algorithm analysis does not consider computational or memory limitations, and establishes the recovery guarantees for a given finite set of measurements; online algorithm analysis takes care of the computational or memory limitations of the system, and establishes the asymptotic guarantees describing limiting behavior as the number of measurements gets larger. Therefore, although both type of algorithms have guarantees for the same special cases, the flavor of results and the mathematical machinery used are completely different. In a few words, our algorithm stands out from the array of existing online sparse recovery algorithms owing to its strong theoretical guarantees.

#### 8.1.3 SM3: Unknown inputs and observation matrix

Under this model, we considered the dictionary learning problem where the goal is to recover both the set of sparse control inputs and the observation matrix from the noisy measurements. We tackled the problem using SBL framework by estimating the dictionary as a deterministic matrix with unit norm columns. Due to this specific hierarchical model, our algorithm outperforms existing Bayesian algorithms which use a prior on the dictionary elements, both in terms of the reconstruction accuracy and run time. The estimation method uses the expectation-maximization (EM) algorithm to simultaneously learn the parameters of the prior and the sparsifying dictionary. The dictionary update step in the EM algorithm turns out to be a quadratic optimization problem with unit norm constraints, which is a nonconvex problem because of the constraint. Since a closed form solution is not available, we propose to employ the alternating minimization (AM) procedure or Armijo line search (ALS) to solve it. We illustrate the performance of the algorithms by comparing with the other popular algorithms when applied to the image denoising problem.

Apart from the superior recovery performance, the main highlight of our dictionary learning algorithm is the associated theoretical guarantees. We have formulated the underlying cost function of the algorithm such that it converges to a sparse representation. Our analysis shows why the DL-SBL algorithm is likely to converge to the sparsest possible representation of the measurement vectors. Further, we derive convergence guarantees of the dictionary update step using AM and ALS optimization procedures and establish the stability of the limit points of the ALS procedure. Thus, the remarkable characteristic of the algorithm compared to other dictionary algorithms is the powerful recovery guarantees.

Next, we summarize the key results in the sparse signal recovery in the context of anomaly imaging.

#### 8.1.4 Anomaly Imaging Exploiting Clustered Sparsity

We present a new algorithm for anomaly imaging, based on ART and the two-dimensional pattern coupled sparse Bayesian learning algorithm.

- *Exploiting Clustered Sparsity:* We exploit the two-dimensional clustered sparsity pattern of the anomaly map to improve damage mapping accuracy.
- *Experimental Validation:* Experimental results on a unidirectional composite plate show that the method of this paper provides more accurate estimates of the damage boundaries than two competing algorithms.

The major takeaway of this work is that exploiting any known structure in the map significantly improves the reconstruction accuracy of the anomaly map.

Overall, the thesis revolves around one special structure, namely sparsity that appears in the linear systems. We looked at three different models of sparsity, and thoroughly examined some of the basic aspects related to sparse signal recovery in the context of LDS. The theoretical analysis presented here is deep-rooted in the rich and elegant mathematical theory of linear algebra, optimization, probability theory (in particular, concentration inequalities and random matrix theory), stochastic approximation, Riemannian matrix manifold, etc. The mathematically significant works presented in the thesis can lead to some promising path of research which we present in the next section.

## 8.2 Future Work

Future work concerns new algorithm development for different but related sparsity models, and a deeper analysis of some problems. Some possible directions for future work are as follows:

- 1. **Stabilizability:** Similar to the analysis of observability and the controllability of LDS presented in Chapter 2 and Chapter 3, a similar theory on the stabilizability aspect of an LDS under sparsity constraint can be delivered.
- 2. Constrained sparse-controllability: Building upon our results in Chapter 3, further studies which impose a constraint on the maximum magnitude of the sparse inputs, can be undertaken in the future.
- 3. Online recovery algorithms: Continuing the online algorithm development in Chapter 4, one can devise online algorithms where the measurements arrive sequentially, for the following scenarios:
  - Single sparse recovery problem (D = I).
  - Sparse input recovery in LDS for arbitrary D and H.
  - Dictionary learning using linear projections of sparse data.
- 4. Dictionary learning for LDS: Following our idea in Chapter 6, a universal algorithm that can learn the observation matrix for a general LDS can be developed.

To conclude, the research presented in the thesis offered new mathematical theory and a bundle of algorithms connecting the areas of control theory, compressed sensing and online learning algorithms. We identified that a sparse structure can arise in LDS in several practical scenarios. The results establish that exploiting the sparsity along with any additional structure is intriguing and fascinating because of the beauty of its theoretical guarantees and the superior performance. Moreover, involving randomness in the measurement step enables one to utilize a minimal number of measurements. These realizations, together with their potential applications, have also triggered some important research questions that need to be looked at next.

# Appendix A

# Appendix to Chapter 2

# A.1 Proof of Proposition 2.1

*Proof.* Using [32, Corollary 7.32] and [32, Theorem 7.30], we can show that, for t > 0,

$$\mathbb{P}\left\{\sum_{l=1}^{m} \left(A_l - \mathbb{E}\left\{A_l\right\}\right) \ge t\right\} \le \exp\left(-\frac{c_2^2 t^2/2}{2c_1 m + c_2 t}\right),\tag{A.1}$$

$$\mathbb{P}\left\{\sum_{l=1}^{m}\left(-A_{l} + \mathbb{E}\left\{A_{l}\right\}\right) \ge t\right\} \le \exp\left(-\frac{c_{2}^{2}t^{2}/2}{2c_{1}m + c_{2}t}\right).$$
(A.2)

Therefore, for  $t > m \max\{a_{\max}, -a_{\min}\},\$ 

$$\mathbb{P}\left\{\sum_{l=1}^{m} A_{l} \ge t\right\} \le \exp\left(-\frac{c_{2}^{2}(t-ma_{\max})^{2}/2}{2c_{1}m+c_{2}(t-ma_{\max})}\right),\\ \mathbb{P}\left\{\sum_{l=1}^{m} -A_{l} \ge t\right\} \le \exp\left(-\frac{c_{2}^{2}(t+ma_{\min})^{2}/2}{2c_{1}m+c_{2}(t+ma_{\min})}\right).$$

We get the desired result by combining the above inequalities using the union bound.  $\Box$ 

### A.2 Proof of Theorem 2.1

*Proof.* First, we note that an overall scaling does not affect the RIP of a matrix. Hence, without loss of generality, we assume that the largest and the smallest singular values of  $D \neq 0$  are 1 and  $\lambda$ , respectively. For any  $\boldsymbol{z} \in \mathbb{R}^N$  such that  $\|\boldsymbol{z}\|^2 = 1$  and  $t \in (0, 1)$ , we have

$$\mathbb{P}\left\{\left|\frac{1}{Km}\left\|\tilde{\boldsymbol{A}}_{(K)}\boldsymbol{z}\right\|^{2}-\left\|\boldsymbol{z}\right\|^{2}\right| \geq t\right\}$$
$$=\mathbb{P}\left\{\left|\sum_{k=0}^{K-1}\sum_{l=1}^{m}\left(a_{k,l}+\left\|\boldsymbol{D}^{k}\boldsymbol{z}\right\|^{2}-\left\|\boldsymbol{z}\right\|^{2}\right)\right| \geq Kmt\right\}, (A.3)$$

where  $a_{k,l} \triangleq |(\mathbf{A}_{(k)}^{\mathsf{T}})_l^{\mathsf{T}} \mathbf{D}^k \mathbf{z}|^2 - ||\mathbf{D}^k \mathbf{z}||^2$ , where  $(\mathbf{A}_{(k)}^{\mathsf{T}})_l^{\mathsf{T}}$  is the  $l^{\text{th}}$  row of the matrix  $\mathbf{A}_{(k)}$ . Here, the term  $(\mathbf{A}_{(k)}^{\mathsf{T}})_l^{\mathsf{T}} \mathbf{D}^k \mathbf{z}$  is the inner product between a row of  $\tilde{\mathbf{A}}_{(K)}$  and  $\mathbf{z}$ . It is easy to see that  $(\mathbf{A}_{(k)}^{\mathsf{T}})_l^{\mathsf{T}} \mathbf{D}^k \mathbf{z}$  is a subgaussian random variable with parameter  $c ||\mathbf{D}^k \mathbf{z}||^2$ . Also, using the independence and unit variance property of the entries of  $(\mathbf{A}_{(k)}^{\mathsf{T}})_l$ , we have  $\mathbb{E} \{a_{k,l}\} = 0$ . Thus, from Lemma 2.1, for  $|\theta| \leq \frac{1}{16c||\mathbf{D}^k \mathbf{z}||^2}$  and hence for  $|\theta| \leq \frac{1}{16c}$ , we have

$$\mathbb{E}\left\{\exp\left(\theta a_{k,l}\right)\right\} \leq \exp\left(128\theta^{2}c^{2} \left\|\boldsymbol{D}^{k}\boldsymbol{z}\right\|^{4}\right) \leq \exp\left(128\theta^{2}c^{2}\right),$$

which follows since the largest singular value of D is 1. Note that this holds true even if D is not invertible. Hence, using the Chernoff bound, for all t > 0,

$$\mathbb{P}\left\{|a_{k,l}| \ge t\right\} \le 2\min_{0 < \theta \le \frac{1}{16c}} \exp\left(128\theta^2 c^2\right) \exp\left(-\theta t\right)$$
(A.4)

$$\leq 2 \exp(1/8) e^{-t/(32c)}$$
 (A.5)

where (A.5) is obtained by setting  $\theta = 1/(32c)$ . Further, independence of the rows of  $A_{(k)}$ for k = 1, 2, ..., K implies that  $a_{k,l}$  are independent. Therefore,  $a_{k,l} + \|\boldsymbol{D}^k \boldsymbol{z}\|^2 - \|\boldsymbol{z}\|^2$  satisfies the conditions required to apply Proposition 2.1. Thus, (A.3), along with the fact  $\lambda^{2(K-1)} - 1 \leq \|\boldsymbol{D}^k \boldsymbol{z}\|^2 - \|\boldsymbol{z}\|^2 \leq 0$  yields, for  $t \in (1 - \lambda^{2(K-1)}, 1)$ 

$$\mathbb{P}\left\{ \left| \frac{1}{Km} \left\| \tilde{\boldsymbol{A}}_{(K)} \boldsymbol{z} \right\|^{2} - \left\| \boldsymbol{z} \right\|^{2} \right| \geq t \right\} \\
\leq \exp\left( -\frac{c_{2}^{2} (Kmt)^{2}/2}{2c_{1}Km + c_{2}Kmt} \right) + \exp\left( -\frac{c_{2}^{2} \left[ Kmt + Km \left( \lambda^{2(K-1)} - 1 \right) \right]^{2}/2}{2c_{1}Km + c_{2}Km \left( t + \lambda^{2(K-1)} - 1 \right)} \right) \quad (A.6)$$

$$\leq \exp\left(-\frac{c_2^2 Kmt^2}{2\left(2c_1+c_2t\right)}\right) + \exp\left(-\frac{c_2^2 Km\left(t+\lambda^{2(K-1)}-1\right)^2}{2\left(2c_1+c_2t\right)}\right)$$
(A.7)

$$\leq 2 \exp\left(-\tilde{c}Km\left(t-1+\lambda^{2(K-1)}\right)^2\right),\tag{A.8}$$

where  $c_1 = 2 \exp(1/8)$ ,  $c_2 = 1/(32c)$  and  $\tilde{c} = \frac{c_2^2}{2(2c_1+c_2)}$ . Also, the last step follows because when  $t \in (1 - \lambda^{2(K-1)}, 1)$ , we have  $t^2 \ge (t + \lambda^{2(K-1)} - 1)^2$ . Now, using the proof technique in [32, Theorem 9.11], we get that if (2.13) holds, the RIC  $\delta_s$  of A satisfies  $\delta_s < \delta$ , for all  $\delta > 1 - \lambda^{2(K-1)}$ , with probability at least  $1 - \epsilon$ . This completes the proof.

## A.3 Proof of Theorem 2.3

Before we prove the theorem, we present a set of mathematical tools used in the proof.

#### A.3.1 Toolbox

Let  $\mathcal{Z} \subset \mathbb{R}^{m \times N}$  be a set of matrices, and the set  $\mathcal{T}_s$  denote the set of *s*-sparse vectors in  $\mathbb{R}^N$ :  $\mathcal{T}_s = \{ \boldsymbol{z} \in \mathbb{R}^N : \|\boldsymbol{z}\| = 1 \text{ and } \|\boldsymbol{z}\|_0 \leq s \}$ . We need the following two definitions to state the results in this subsection.

**Definition A.1** (Admissible sequence). An admissible sequence  $\mathcal{U} = {\mathcal{U}_i}_{i=0}^{\infty}$  on  $\mathcal{Z}$  is an increasing sequence of partitions of  $\mathcal{Z}$  such that  $|\mathcal{U}_i| = 2^{2^i}$  and  $|\mathcal{U}_0| = 1$ . Here, increasing sequence of partitions implies that every set of  $\mathcal{U}_i$  is contained in one of the sets of  $\mathcal{U}_{i-1}$ 

for all *i*. Also,  $\mathcal{U}_0 = \mathcal{Z}$ , and every set of  $\mathcal{U}_i$  is a subset of  $\mathcal{Z}$ . Given a matrix  $\mathbf{Z} \in \mathcal{Z}$ , we denote the unique set of  $\mathcal{U}_i$  that contains  $\mathbf{Z}$  by  $\mathcal{U}_i(\mathbf{Z})$ .

**Definition A.2.** [Functionals on a set of matrices] We define three functionals on  $\mathcal{Z}$  as follows:

$$d_F\left(\mathcal{Z}\right) \triangleq \sup_{\mathbf{Z} \in \mathcal{Z}} \|\mathbf{Z}\|_F \tag{A.9}$$

$$d_2\left(\mathcal{Z}\right) \triangleq \sup_{\mathbf{Z} \in \mathcal{Z}} \|\mathbf{Z}\|_2 \tag{A.10}$$

$$\zeta\left(\mathcal{Z}\right) \triangleq \inf_{\mathcal{U} = \{\mathcal{U}_i\}_{i=0}^{\infty}} \sup_{\mathbf{Z} \in \mathcal{Z}} \sum_{i=0}^{\infty} 2^{i/2} \mathcal{D}(\mathcal{U}_i(\mathbf{Z})),$$
(A.11)

where the inf is over all possible admissible sequences, and the term  $\mathcal{D}$  is defined as follows:

$$\mathcal{D}(\mathcal{U}_{i}(\boldsymbol{Z})) \triangleq \max_{\boldsymbol{U}_{(1)}, \boldsymbol{U}_{(2)} \in \mathcal{U}_{i}(\boldsymbol{Z})} \left\| \boldsymbol{U}_{(1)} - \boldsymbol{U}_{(2)} \right\|.$$
(A.12)

It represents the diameter the set  $\mathcal{U}_i(\mathbf{Z})$ , which is a decreasing function of *i*.

Next, we state a result which is the main ingredient of our proof. It bounds the suprema of a chaos process indexed by the set  $\mathcal{Z}$ .

**Theorem A.1.** [72, Theorem 3.1] Let **u** be a random vector whose entries are independent zero-mean, unit-variance subgaussian random variables with common parameter c. Let

$$F_{1} \triangleq \zeta\left(\mathcal{Z}\right)\left[\zeta\left(\mathcal{Z}\right) + d_{F}\left(\mathcal{Z}\right)\right] + d_{F}\left(\mathcal{Z}\right)d_{2}\left(\mathcal{Z}\right)$$
(A.13)

$$F_2 \triangleq d_2^2\left(\mathcal{Z}\right) \left[\zeta\left(\mathcal{Z}\right) + d_F\left(\mathcal{Z}\right)\right]^2 \tag{A.14}$$

$$F_3 \triangleq d_2^2\left(\mathcal{Z}\right). \tag{A.15}$$

Then, for t > 0, it holds that

$$\mathbb{P}\left\{\sup_{\boldsymbol{Z}\in\mathcal{Z}}\left|\|\boldsymbol{Z}\boldsymbol{u}\|^{2}-\mathbb{E}\left\{\|\boldsymbol{Z}\boldsymbol{u}\|^{2}\right\}\right|>c_{1}F_{1}+t\right\}\leq2\exp\left(-c_{2}\min\left\{\frac{t^{2}}{F_{2}},\frac{t}{F_{3}}\right\}\right),\quad(A.16)$$

where  $c_1, c_2 > 0$  are universal positive constants which depend only on c.

It is difficult to directly apply the above theorem due to the complicated form of the functional  $\zeta(\mathcal{Z})$ . We need a result that bounds the function  $\zeta(\mathcal{Z})$  using the notion of the covering number. The covering number is defined as follows.

**Definition A.3** (Covering number). Given u > 0, the covering number  $CN \{Z, u\}$  is defined as the smallest integer p such that one can find a subset  $Z' \subset Z$  satisfying  $|Z'| \leq p$  and

$$\min_{\mathbf{Z}'\in\mathcal{Z}'}\|\mathbf{Z}-\mathbf{Z}'\|_2 \le u, \forall \mathbf{Z}\in\mathcal{Z}.$$
(A.17)

**Lemma A.1.** [179]. For the functional  $\zeta(\mathcal{Z})$ , it holds that

$$\zeta\left(\mathcal{Z}\right) \le \int_{0}^{\infty} \ln^{1/2} \left(\mathsf{CN}\left\{\mathcal{Z}, u\right\}\right) du. \tag{A.18}$$

The covering number is hard to compute in closed form. Hence, we use the following lemma to further bound the covering number of the set of interest to us.

**Lemma A.2.** [74, Lemma 6] Let a linear map  $\mathbf{A} : \mathbb{R}^N \to \mathbb{R}^m$  be such that

$$\left\|\frac{1}{\sqrt{s}}\boldsymbol{A}(\boldsymbol{z})\right\| \leq \kappa \left\|\boldsymbol{z}\right\|_{1}, \forall \boldsymbol{z} \in \mathbb{R}^{N}, \kappa > 0.$$
(A.19)

Then, for the set  $\mathcal{T}_s$ , it holds that

$$\ln\left(\mathsf{CN}\left\{\boldsymbol{A}(\mathcal{T}_{s}), u\right\}\right) \qquad \begin{cases} \leq Cs \min\left\{\frac{\kappa^{2}}{u^{2}}\ln^{2} N, s \ln N + s \ln\left(1 + \frac{2\kappa}{u}\right)\right\} \\ for \ 0 < u < \kappa \end{cases} \tag{A.20} \\ = 0 \qquad \qquad for \ u > \kappa, \end{cases}$$

where C is a universal positive constant.

We will also need the following result from elementary calculus in the proof to tackle the integral in Lemma A.1.

**Lemma A.3.** For  $a \leq b$ , it holds that

$$\int_{0}^{a} \sqrt{\ln\left(1+\frac{b}{u}\right)} du \le \frac{3}{2}a \sqrt{\ln\left(1+\frac{b}{a}\right)}.$$
(A.21)

*Proof.* See Appendix A.4.

We have now presented all the mathematical tools that are required to prove the theorem. In the next subsection, we formally prove the desired result.

#### A.3.2 Proof

As mentioned in Section 2.2, (2.6) is equivalent to (2.4). Therefore, without loss of genreality, we assume that the largest and the smallest singular values of  $\boldsymbol{D}$  are 1 and  $\lambda$ , respectively. We recall that our goal is to obtain a probabilistic bound on  $\|\widetilde{\boldsymbol{A}}_{(K)}\boldsymbol{z}\|$  for  $\boldsymbol{z} \in \mathcal{T}_s$ , using Theorem A.1. At a high level, there are four main steps to the proof:

- First, we convert *A*<sub>(K)</sub>*z* to the form given in Theorem A.1, i.e., the product of a matrix and a subgaussian vector.
- Second, we bound the three functionals  $d_2$ ,  $d_F$  and  $\zeta$  in Definition A.2.

- Third, using the bounds in the previous step, we bound  $F_1, F_2$  and  $F_3$  in Theorem A.1, since the three quantities are functions of  $d_2$ ,  $d_F$  and  $\zeta$ .
- Fourth, we apply Theorem A.1 with the upper bounds on  $F_1, F_2$  and  $F_3$ . Here, we note that Theorem A.1 holds for upper bounds on  $F_1, F_2$  and  $F_3$ . This yields a concentration inequality bounding the deviation of the random variable  $\left\|\frac{1}{\sqrt{Km}}\widetilde{A}_{(K)}\boldsymbol{z}\right\|^2$  from its mean  $\frac{1}{K}\sum_{k=0}^{K-1} \|\boldsymbol{D}^k\boldsymbol{z}\|^2$ . Finally, we establish the desired result by suitable algebraic manipulation of the concentration inequality.

In the remainder of this section, we provide the details of each of these steps. For the first step, we consider the following:

$$\left\|\frac{1}{\sqrt{Km}}\widetilde{\boldsymbol{A}}_{(K)}\boldsymbol{z}\right\|^{2} = \sum_{k=0}^{K-1} \left\|\frac{1}{\sqrt{Km}}\boldsymbol{A}\boldsymbol{D}^{k}\boldsymbol{z}\right\|^{2}$$
(A.22)

$$= \left\| \frac{1}{\sqrt{Km}} \boldsymbol{A} \boldsymbol{Z}_{(K)}(\boldsymbol{z}) \right\|_{F}^{2}$$
(A.23)

where the matrix function  $\boldsymbol{Z}_{(K)} : \mathcal{T}_s \to \mathbb{R}^{N \times K}$  is defined as follows:

$$\boldsymbol{Z}_{(K)}(\boldsymbol{z}) \triangleq \begin{bmatrix} \boldsymbol{z} & \boldsymbol{D}\boldsymbol{z} & \dots \boldsymbol{D}^{K-1}\boldsymbol{z} \end{bmatrix}, \, \boldsymbol{z} \in \mathcal{T}_s.$$
 (A.24)

Further, we have

$$\left\|\frac{1}{\sqrt{Km}}\widetilde{\boldsymbol{A}}_{(K)}\boldsymbol{z}\right\|^{2} = \left\|\frac{1}{\sqrt{Km}}\boldsymbol{Z}_{(K)}^{\mathsf{T}}(\boldsymbol{z})\boldsymbol{A}^{\mathsf{T}}\right\|_{F}^{2}$$
(A.25)

$$= \left\| \widetilde{\boldsymbol{Z}}_{(K)}(\boldsymbol{z}) \operatorname{vec} \left\{ \boldsymbol{A}^{\mathsf{T}} \right\} \right\|^{2}, \qquad (A.26)$$

where  $\operatorname{vec} \{ \boldsymbol{A}^{\mathsf{T}} \} \in \mathbb{R}^{mN}$  is the vectorized version of the matrix  $\boldsymbol{A}^{\mathsf{T}}$ , which has subgaussian entries with common parameter c. The matrix function  $\widetilde{\boldsymbol{Z}}_{(K)} : \mathcal{T}_s \to \mathbb{R}^{Km \times mN}$  is a block diagonal matrix with  $\frac{1}{\sqrt{Km}} \boldsymbol{Z}_{(K)}^{\mathsf{T}}(\boldsymbol{z})$  as the  $m^{\mathsf{th}}$  block diagonal entries, for all  $\boldsymbol{z} \in \mathcal{T}_s$ . Thus, the first step is complete.

The next step is bound the three terms  $d_2$ ,  $d_F$  and  $\zeta$  using the following lemmas.

**Lemma A.4.** For the set  $\widetilde{\mathcal{Z}}_{(K)}$ ,

$$d_F\left(\widetilde{\mathcal{Z}}_{(K)}\right) \le 1 \tag{A.27}$$

$$d_2\left(\widetilde{\mathcal{Z}}_{(K)}\right) \le \sqrt{\frac{s}{Km}} L\left(\boldsymbol{D}, K\right), \qquad (A.28)$$

where  $L(\mathbf{D}, K)$  is as defined in (2.32).

*Proof.* See Appendix A.5.

**Lemma A.5.** The functional  $\zeta\left(\widetilde{\mathcal{Z}}_{(K)}\right)$  can be bounded as follows:

$$\zeta\left(\widetilde{\mathcal{Z}}_{(K)}\right) \le C\sqrt{\frac{s}{Km}}L\left(\boldsymbol{D},K\right)\ln N\ln s,\tag{A.29}$$

for some C which is universal positive constant that depends only on the subgaussian parameter c.

*Proof.* See Appendix A.6.

Now, we combine the results in the second step to obtain bounds on  $F_1, F_2$ , and  $F_3$ . Further, we note that we need to bound  $\mathbb{P}\left\{ \left\| \frac{1}{\sqrt{Km}} \widetilde{\boldsymbol{A}}_{(K)} \boldsymbol{z} \right\|^2 - \frac{1}{K} \sum_{k=0}^{K-1} \|\boldsymbol{D}^k \boldsymbol{z}\|^2 \right| > \widetilde{\delta} \right\}$ , for some  $0 \leq \widetilde{\delta} < 1$  using Theorem A.1. To this end, we use the assumptions of Theorem 2.3 to further upper bound  $F_1$  to make it a multiple of  $\widetilde{\delta}$ . We summarize the third step in the following lemma.

**Lemma A.6.** Using the definitions in Theorem A.1 and under the assumptions of Theorem 2.3, there exists a constant  $C_1 > 0$  such that

$$F_1 \le \tilde{\delta}/2c_1 \tag{A.30}$$

$$F_2 \le C_1 \frac{s}{Km} L^2 \left( \boldsymbol{D}, K \right) \tag{A.31}$$

$$F_3 \le \frac{s}{Km} L^2(\boldsymbol{D}, K), \qquad (A.32)$$

where  $c_1$  is the same constant as in Theorem A.1 and

$$\widetilde{\delta} \triangleq \delta - 1 + \lambda^{2(K-1)}. \tag{A.33}$$

Proof. See Appendix A.7.

Now, we are ready to go the final step of the proof. We apply Theorem A.1 to (A.26) with  $t = \tilde{\delta}/2$  to get

$$\mathbb{P}\left\{\left\|\left\|\frac{1}{\sqrt{Km}}\widetilde{\boldsymbol{A}}_{(K)}\boldsymbol{z}\right\|^{2}-\frac{1}{K}\sum_{k=0}^{K-1}\left\|\boldsymbol{D}^{k}\boldsymbol{z}\right\|^{2}\right|>\widetilde{\delta}\right\}\leq 2\exp\left(-C\min\left\{\widetilde{\delta}^{2},\widetilde{\delta}\right\}\frac{Km}{sL^{2}\left(\boldsymbol{D},K\right)}\right) \tag{A.34}$$

$$\leq 2\exp\left(-C\frac{Km\widetilde{\delta}^{2}}{sL^{2}\left(\boldsymbol{D},K\right)}\right)\leq\epsilon,\quad(A.35)$$

where the universal positive constant C depends on the subgaussian parameter c, and we use (2.33) of Theorem 2.3 to bound using  $\epsilon$  in the last step.

Thus, for all  $\boldsymbol{z} \in \mathbb{R}^N$  such that  $\|\boldsymbol{z}\| = 1$  and  $\|\boldsymbol{z}\|_0 \leq s$ , with probability at least  $1 - \epsilon$ ,

$$\left\| \left\| \frac{1}{\sqrt{Km}} \widetilde{\boldsymbol{A}}_{(K)} \boldsymbol{z} \right\|^2 - \frac{1}{K} \sum_{k=0}^{K-1} \left\| \boldsymbol{D}^k \boldsymbol{z} \right\|^2 \right\| \le \widetilde{\delta}.$$
(A.36)

#### Appendix A.

Therefore, for  $0 \leq \tilde{\delta} < \lambda^{2(K-1)}$ ,

$$\lambda^{2(K-1)} - \widetilde{\delta} < \frac{1}{Km} \left\| \widetilde{\boldsymbol{A}}_{(K)} \boldsymbol{z} \right\|^2 < 1 + \widetilde{\delta}, \tag{A.37}$$

since  $\lambda^{K-1} \leq \lambda^k \leq \|\boldsymbol{D}^k \boldsymbol{z}\| \leq 1$ . We also use (A.33) to relate  $\delta$  and  $\tilde{\delta}$  as follows:

$$1 - \delta < \frac{1}{Km} \left\| \widetilde{\boldsymbol{A}}_{(K)} \boldsymbol{z} \right\|^2 < 1 + \delta,$$
(A.38)

for  $\delta > 1 - \lambda^{2(K-1)}$ , with probability at least  $1 - \epsilon$ . Hence,  $\frac{1}{Km} \widetilde{A}_{(K)}$  satisfies RIP of order s with RIC as  $\delta$ , with probability at least  $1 - \epsilon$ . Thus, the proof is complete.

# A.4 Proof of Lemma A.3

We have

$$\int_{0}^{a} \sqrt{\ln\left(1+\frac{b}{u}\right)} du = b \int_{\sqrt{\ln(1+b/a)}}^{\infty} td\left(\frac{1}{\exp(t^{2})-1}\right)$$
(A.39)

$$=a\sqrt{\ln\left(1+\frac{b}{a}\right)+b\int_{\sqrt{\ln(1+b/a)}}^{\infty}\frac{1}{\exp(t^2)-1}dt},$$
 (A.40)

where we use the substitution  $t = \sqrt{\ln\left(1 + \frac{b}{u}\right)}$  in (A.39) and integration by parts to get (A.40). Now, the second term simplifies as follows:

$$\int_{\sqrt{\ln(1+b/a)}}^{\infty} \frac{1}{\exp(t^2) - 1} dt \le \int_{\sqrt{\ln(1+b/a)}}^{\infty} \frac{t}{\sqrt{\ln(1+b/a)}} \frac{e^{-t^2}}{1 - e^{-t^2}} dt$$
(A.41)

$$= \frac{1}{2\sqrt{\ln(1+b/a)}} \ln\left(1+\frac{a}{b}\right). \tag{A.42}$$

Therefore, we get

$$\int_{0}^{a} \sqrt{\ln\left(1+\frac{b}{u}\right)} du \le a \sqrt{\ln\left(1+\frac{b}{a}\right)} + \frac{b}{2\sqrt{\ln(1+b/a)}} \ln\left(1+\frac{a}{b}\right) \tag{A.43}$$

$$=a\sqrt{\ln\left(1+\frac{b}{a}\right)\left[1+\frac{b}{2a}\left(1-\frac{\ln\left(\frac{b}{a}\right)}{\ln(1+b/a)}\right)\right]}.$$
 (A.44)

Now, we need to show that  $\frac{b}{a}\left(1-\frac{\ln(\frac{b}{a})}{\ln(1+b/a)}\right) \leq 1$  to complete the proof. So, we consider the function  $h(u) \triangleq u\left(1-\frac{\ln u}{\ln(u+1)}\right)$ , by replacing  $b/a = u \geq 1$ . Further, we note that h(1) = 1, and therefore it suffices to show that  $\frac{d}{du}h(u) \leq 0$ , which then implies that  $h(u) \leq h(1) = 1$ , for all  $u \geq 1$ . We have

$$\frac{d}{du}h(u) = 1 - \frac{\ln u}{\ln(u+1)} - \frac{(u+1)\ln(u+1) - u\ln u}{(u+1)\ln^2(u+1)}$$
(A.45)

$$=\frac{h(u)}{(u+1)\ln^2(u+1)},$$
(A.46)

where we define

$$\widetilde{h}(u) \triangleq (u+1)\ln^2(u+1) - (u+1)\ln(u+1)\ln u - (u+1)\ln(u+1) + u\ln u. \quad (A.47)$$

Now,  $\frac{d}{du}h(u) \leq 0$  if  $\tilde{h}(u) \leq 0$ . Therefore, we show that  $\frac{d}{du}\tilde{h}(u) \leq 0$ , which implies that  $\tilde{h}(u) \leq \tilde{h}(1) = 2\ln^2 2 - 2\ln 2 < 0$ . Then, we get

$$\frac{d}{du}\widetilde{h}(u) = \ln(u+1)\left(\ln(u+1) - \ln u - \frac{u+1}{u}\right)$$
(A.48)

$$= -\ln(u+1)\left(\ln u + 1 - \ln(u+1) + \frac{1}{u}\right).$$
 (A.49)

Using the same technique again, we now consider the function  $\ln u + 1 - \ln(u+1)$ . Since derivative of  $\ln u + 1 - \ln(u+1)$  is  $\frac{1}{u(u+1)} > 0$ , for  $u \ge 1$ ,  $\ln u + 1 - \ln(u+1) \ge 1 - \ln 2 > 0$ . Therefore,  $\frac{d}{du}\tilde{h}(u) \le 0$  because  $\ln u + 1 - \ln(u+1) \ge 0$  and  $\frac{1}{u} \ge 0$ , for  $u \ge 1$ . Hence, we get  $\tilde{h}(u) < 0$ . This implies that  $\frac{d}{du}h(u) < 0$ , and thus,  $h(u) \le h(1) = 1$ , for  $u \ge 1$ . Substituting this in (A.44) completes the proof.

## A.5 Proof of Lemma A.4

To show the first part of the lemma, we have

$$d_F\left(\widetilde{\mathcal{Z}}_{(K)}\right) = \sup_{\boldsymbol{z}\in\mathcal{T}_s} \left\|\widetilde{\boldsymbol{Z}}_{(K)}(\boldsymbol{z})\right\|_F$$
(A.50)

$$= \frac{1}{\sqrt{K}} \sup_{\boldsymbol{z} \in \mathcal{T}_s} \left\| \boldsymbol{Z}_{(K)}(\boldsymbol{z}) \right\|_F$$
(A.51)

$$= \sup_{\boldsymbol{z}\in\mathcal{T}_{s}}\frac{1}{\sqrt{K}}\sqrt{\sum_{k=0}^{K-1}\left\|\boldsymbol{D}^{k}\boldsymbol{z}\right\|^{2}} \leq 1, \qquad (A.52)$$

where the last step follows from the definition of  $\mathcal{T}_s$ , and the fact that the largest singular value of D is unity.

To show the second part of the lemma, we have

$$d_2\left(\widetilde{\mathcal{Z}}_{(K)}\right) = \sup_{\boldsymbol{z}\in\mathcal{T}_s} \left\|\widetilde{\boldsymbol{Z}}_{(K)}(\boldsymbol{z})\right\|_2 = \frac{1}{\sqrt{Km}} \sup_{\boldsymbol{z}\in\mathcal{T}_s} \left\|\boldsymbol{Z}_{(K)}(\boldsymbol{z})\right\|_2$$
(A.53)

$$= \frac{1}{\sqrt{Km}} \sup_{\boldsymbol{z}\in\mathcal{T}_s} \left\| \sum_{i=1}^{N} \widetilde{\boldsymbol{D}}_{(K,i)} \boldsymbol{z}_i \right\|_2$$
(A.54)

$$\leq \frac{1}{\sqrt{Km}} \sup_{\boldsymbol{z}\in\mathcal{T}_s} \sum_{i=1}^{N} |\boldsymbol{z}_i| \left\| \widetilde{\boldsymbol{D}}_{(K,i)} \right\|_2 \leq \frac{L(\boldsymbol{D},K)}{\sqrt{Km}} \sup_{\boldsymbol{z}\in\mathcal{T}_s} \|\boldsymbol{z}\|_1$$
(A.55)

$$\leq L\left(\boldsymbol{D},K\right)\sqrt{\frac{s}{Km}}\left\|\boldsymbol{z}\right\|,\tag{A.56}$$

$$=\sqrt{\frac{s}{Km}L\left(\boldsymbol{D},K\right)}.$$
(A.57)

where (A.54) and (A.55) follow from the definitions of  $\widetilde{D}_{(K,i)}$  and L(D, K) in (2.31) and (2.32), respectively. Also, (A.56) is because  $\boldsymbol{z}$  is at most s-sparse. Hence, the proof is complete.

## A.6 Proof of Lemma A.5

From Lemma A.4, for all  $\boldsymbol{Z} \in \widetilde{\mathcal{Z}}_{(K)}$  and any  $\boldsymbol{z} \in \mathbb{R}^{mN}$ ,

$$\left\|\frac{1}{\sqrt{s}}\boldsymbol{Z}\boldsymbol{z}\right\| \leq \sqrt{\frac{1}{Km}}L\left(\boldsymbol{D},K\right)\|\boldsymbol{z}\| \leq \sqrt{\frac{1}{Km}}L\left(\boldsymbol{D},K\right)\|\boldsymbol{z}\|_{1}.$$
 (A.58)

Then, for some positive constant C', we have

$$\begin{split} \frac{1}{\sqrt{s}}\zeta\left(\widetilde{\mathcal{Z}}_{(K)}\right) &\leq \frac{1}{\sqrt{s}} \int_{0}^{\infty} \ln^{1/2} \left[ \mathsf{CN}\left\{\widetilde{\mathcal{Z}}_{(K)}, u\right\} \right] du \quad (A.59) \\ &= C' \frac{1}{\sqrt{s}} \int_{0}^{\frac{L(\mathbf{D},K)}{\sqrt{sKm}}} \ln^{1/2} \left[ \mathsf{CN}\left\{\widetilde{\mathcal{Z}}_{(K)}, u\right\} \right] du \quad (A.60) \\ &\quad + C' \frac{1}{\sqrt{s}} \int_{\frac{L(\mathbf{D},K)}{\sqrt{sKm}}}^{\frac{L(\mathbf{D},K)}{\sqrt{sKm}}} \ln^{1/2} \left[ \mathsf{CN}\left\{\widetilde{\mathcal{Z}}_{(K),u}\right\} \right] du \quad (A.60) \\ &\leq C' \int_{0}^{\frac{L(\mathbf{D},K)}{\sqrt{sKm}}} \sqrt{s\ln N + s\ln\left(1 + \frac{2L\left(\mathbf{D},K\right)}{u\sqrt{Km}}\right)} du \\ &\quad + C' \int_{\frac{L(\mathbf{D},K)}{\sqrt{sKm}}}^{\frac{L(\mathbf{D},K)}{\sqrt{sKm}}} \frac{L\left(\mathbf{D},K\right)}{u\sqrt{Km}} \ln N du, \quad (A.61) \end{split}$$

where (A.59) and (A.61) follow from Lemma A.1 and Lemma A.2 with  $\kappa = \sqrt{\frac{1}{Km}} L(\boldsymbol{D}, K)$ , respectively. Further, we have

$$\frac{1}{\sqrt{s}}\zeta\left(\widetilde{\mathcal{Z}}_{(K)}\right) \leq C'\int_{0}^{\frac{L(\boldsymbol{D},K)}{\sqrt{sKm}}}\sqrt{s\ln N} + \sqrt{s\ln\left(1 + \frac{2L(\boldsymbol{D},K)}{u\sqrt{Km}}\right)}du + C'\frac{L(\boldsymbol{D},K)}{\sqrt{Km}}\ln N\ln\sqrt{s}$$
(A.62)

$$\leq C' \frac{L(\boldsymbol{D}, K)}{\sqrt{Km}} \left( \sqrt{\ln N} + 3/2\sqrt{\ln\left(1 + 2\sqrt{s}\right)} + \ln N \ln \sqrt{s} \right)$$
(A.63)

$$\leq C \frac{L\left(\boldsymbol{D},K\right)}{\sqrt{Km}} \ln N \ln s \tag{A.64}$$

where C = 3C'. Also, (A.62) uses the fact that  $\sqrt{a+b} \le \sqrt{a} + \sqrt{b}$ , for any a, b > 0, and (A.63) uses Lemma A.3. Thus, the proof is complete.

# A.7 Proof of Lemma A.6

From Lemma A.4 and Lemma A.5, we get

$$F_{1} = \zeta \left( \widetilde{\mathcal{Z}}_{(K)} \right) \left[ \zeta \left( \widetilde{\mathcal{Z}}_{(K)} \right) + d_{F} \left( \widetilde{\mathcal{Z}}_{(K)} \right) \right] + d_{F} \left( \widetilde{\mathcal{Z}}_{(K)} \right) d_{2} \left( \widetilde{\mathcal{Z}}_{(K)} \right)$$

$$\leq CL \left( \boldsymbol{D}, K \right) \sqrt{\frac{s}{Km}} \ln N \ln s \left( CL \left( \boldsymbol{D}, K \right) \sqrt{\frac{s}{Km}} \ln N \ln s + 1 \right)$$

$$+ L \left( \boldsymbol{D}, K \right) \sqrt{\frac{s}{Km}} \frac{\ln N \ln s}{\ln^{2} 2},$$
(A.66)

where we use the bound  $\frac{\ln N \ln s}{\ln^2 2} > 1$  when  $N \ge s > 1$  to get (A.66). Next, we use assumption (2.33) in Theorem 2.3, i.e.,

$$L(\mathbf{D}, K) \sqrt{\frac{s}{Km}} \ln N \ln s \le \sqrt{\widetilde{c}\widetilde{\delta}},$$
 (A.67)

to get

$$F_1 \le C\sqrt{\widetilde{c}\widetilde{\delta}} \left(C\sqrt{\widetilde{c}\widetilde{\delta}} + 1\right) + \frac{\sqrt{\widetilde{c}\widetilde{\delta}}}{\ln^2 2} \tag{A.68}$$

$$\leq \sqrt{\widetilde{c}\widetilde{\delta}} \left( C^2 \sqrt{\widetilde{c}} + C + \frac{1}{\ln^2 2} \right). \tag{A.69}$$

The last step above follows because of the bound  $\tilde{\delta} \leq 1$ . Finally, we choose  $\tilde{c}$  in (2.33) of Theorem 2.3 such that

$$\sqrt{\widetilde{c}}\left(C^2\sqrt{\widetilde{c}} + C + \frac{1}{\ln^2 2}\right) \le \frac{1}{2c_1},\tag{A.70}$$

where  $c_1$  and C are the same constants as in Theorem A.1 and Lemma A.5, respectively. We note that  $\sqrt{\tilde{c}} \left( C^2 \sqrt{\tilde{c}} + C + \frac{1}{\ln^2 2} \right)$  is strictly increasing with  $\tilde{c}$ , for  $\tilde{c} \ge 0$ , and the left hand side equals zero when  $\tilde{c} = 0$ . Thus, there exists a  $\tilde{c} > 0$  that satisfies (A.70), for any  $c_1$  and C. Thus, from (A.69) and (A.70) we get

$$F_1 \le \frac{\tilde{\delta}}{2c_1},\tag{A.71}$$

and thus, we complete of the first part of the proof. Similarly, we simplify  $F_2$  using Lemma A.4, Lemma A.5 and (A.67) as follows:

$$F_{2} = d_{2}^{2} \left( \widetilde{\mathcal{Z}}_{(K)} \right) \left[ \zeta \left( \widetilde{\mathcal{Z}}_{(K)} \right) + d_{F} \left( \widetilde{\mathcal{Z}}_{(K)} \right) \right]^{2}$$
(A.72)

$$\leq \frac{s}{Km}L^{2}\left(\boldsymbol{D},K\right)\left(CL\left(\boldsymbol{D},K\right)\sqrt{\frac{s}{Km}}\ln N\ln s+1\right)^{2}$$
(A.73)

$$\leq \frac{s}{Km} L^2 \left( \boldsymbol{D}, K \right) \left( C \sqrt{\tilde{c}} \tilde{\delta} + 1 \right)^2 \tag{A.74}$$

$$\leq C_1 \frac{s}{Km} L^2 \left( \boldsymbol{D}, K \right) \tag{A.75}$$

where we use the fact that  $\tilde{\delta} < 1$  and define  $C_1 \triangleq (C\sqrt{\tilde{c}} + 1)^2$ . Finally, we have,

$$F_3 = d_2^2 \left( \widetilde{\mathcal{Z}}_{(K)} \right) \le \frac{s}{Km} L^2 \left( \boldsymbol{D}, K \right), \qquad (A.76)$$

which completes the proof.

## A.8 Proof of Proposition 2.2

To prove the result, we first upper and lower bound the term  $L(\mathbf{D}, K)$ . We have,

$$L(\boldsymbol{D}, K) \le \max_{i} \left\| \widetilde{\boldsymbol{D}}_{(K,i)} \right\|_{F}$$
(A.77)

$$= \max_{i} \sqrt{\sum_{k=0}^{K-1} \left\| \boldsymbol{D}_{i}^{k} \right\|^{2}} \leq \sqrt{K}, \qquad (A.78)$$

where we obtain the last step from the fact that the largest singular value of  $D^k$  is at most unity, and Rayleigh-Ritz theorem [180, Theorem 4.2.2] which gives

$$1 = \sup_{\boldsymbol{z} \in \mathbb{R}^{N}, \boldsymbol{z} \neq 0} \frac{\left\|\boldsymbol{z}^{\mathsf{T}} \boldsymbol{D}^{k} \boldsymbol{z}\right\|}{\left\|\boldsymbol{z}\right\|^{2}} \ge \max_{i} \left(\boldsymbol{D}^{k} \mathbf{D}^{k}\right)_{i,i} = \max_{i} \left\|\boldsymbol{D}_{i}^{k}\right\|^{2}.$$
(A.79)

Similarly, we also have,

$$L(\boldsymbol{D},K)^{2} = \max_{i=1,2,\dots,N} \left[ \sup_{\boldsymbol{z} \in \mathbb{R}^{K}, \boldsymbol{z} \neq 0} \frac{\left\| \boldsymbol{z}^{\mathsf{T}} \widetilde{\boldsymbol{D}}_{(K,i)}^{\mathsf{T}} \widetilde{\boldsymbol{D}}_{(K,i)} \boldsymbol{z} \right\|}{\left\| \boldsymbol{z} \right\|^{2}} \right]$$
(A.80)

$$\geq \max_{\substack{i=1,2,\dots,N\\k=0,1,\dots,K-1}} \left( \widetilde{\boldsymbol{D}}_{(K,i)}^{\mathsf{T}} \widetilde{\boldsymbol{D}}_{(K,i)} \right)_{k,k}$$
(A.81)

$$= \max_{i} \left[ \max_{k=0,1,\dots,K-1} \left\| \boldsymbol{D}_{i}^{k} \right\|^{2} \right] = 1, \qquad (A.82)$$

where (A.82) uses the fact that  $\|\boldsymbol{D}_{i}^{0}\|^{2} = \|\boldsymbol{I}_{i}\|^{2} = 1$  and  $\|\boldsymbol{D}_{i}^{k}\|^{2} \leq 1$ , for  $k = 1, 2, \dots, K-1$  from (A.79). Combining (A.78) and (A.82), we obtain the desired result.

# A.9 Proof of Proposition 2.3

To prove the monotonicity of the function, we need to show that

$$L^{2}\left(\boldsymbol{D}, K+1\right) \leq L^{2}\left(\boldsymbol{D}, K\right)\left(1+\frac{1}{K}\right).$$
(A.83)

Therefore, we consider the following relation connecting  $L^2(\mathbf{D}, K+1)$  and  $L^2(\mathbf{D}, K)$ :

$$L^{2}(\boldsymbol{D}, K+1) = \max_{i} \left\| \widetilde{\boldsymbol{D}}_{(K+1,i)} \right\|_{2}$$
(A.84)

$$= \max_{\substack{i \\ \|\boldsymbol{z}\|=1}} \sup_{\boldsymbol{z} \in \mathbb{R}^{N} \\ \|\boldsymbol{z}\|=1}} \boldsymbol{z}^{\mathsf{T}} \widetilde{\boldsymbol{D}}_{(K+1,i)} \widetilde{\boldsymbol{D}}_{(K+1,i)}^{\mathsf{T}} \boldsymbol{z}$$
(A.85)

$$= \max_{\substack{i \\ \|\boldsymbol{z}\|=1}} \sup_{j=0}^{K} \sum_{j=0}^{K} \left| \boldsymbol{z}^{\mathsf{T}} \boldsymbol{D}^{j} \boldsymbol{I}_{i} \right|^{2}$$
(A.86)

$$\leq \max_{i} \sup_{\substack{\boldsymbol{z} \in \mathbb{R}^{N} \\ \|\boldsymbol{z}\|=1}} \sum_{j=0}^{K-1} |\boldsymbol{z}^{\mathsf{T}} \boldsymbol{D}^{j} \boldsymbol{I}_{i}|^{2} + \max_{i} \sup_{\substack{\boldsymbol{z} \in \mathbb{R}^{N} \\ \|\boldsymbol{z}\|=1}} |\boldsymbol{z}^{\mathsf{T}} \boldsymbol{D}^{K} \boldsymbol{I}_{i}|^{2}$$
(A.87)

$$\leq L^{2}\left(\boldsymbol{D},K\right) + \max_{i} \left\|\boldsymbol{D}^{K}\boldsymbol{I}_{i}\right\|^{2}, \qquad (A.88)$$

where we use the definition of  $L^{2}(\mathbf{D}, K)$  and Cauchy-Schwarz inequality to get the last step.

Now, to complete the proof, it suffices to show that

$$\max_{i} \left\| \boldsymbol{D}^{K} \boldsymbol{I}_{i} \right\|^{2} \leq L^{2} \left( \boldsymbol{D}, K \right) / K.$$
(A.89)

Since D is a psd matrix with largest singular value as unity,  $D^{j} - D^{k}$  is a psd matrix, for

any pair of integers  $j \leq k$ . Then, we have

$$K^{2} \left\| \boldsymbol{D}^{K} \boldsymbol{I}_{i} \right\|^{2} = K^{2} \boldsymbol{I}_{i}^{\mathsf{T}} \boldsymbol{D}^{2K} \boldsymbol{I}_{i} \leq \sum_{j=0}^{K-1} \sum_{k=0}^{K-1} \boldsymbol{I}_{i}^{\mathsf{T}} \boldsymbol{D}^{j+k} \boldsymbol{I}_{i}$$
(A.90)

$$=\sum_{j=0}^{K-1}\sum_{k=0}^{K-1}\boldsymbol{I}_{i}^{\mathsf{T}}\boldsymbol{D}^{j\mathsf{T}}\boldsymbol{D}^{k}\boldsymbol{I}_{i}$$
(A.91)

$$= K \left( \mathbf{1} / \sqrt{K} \right)^{\mathsf{T}} \widetilde{\boldsymbol{D}}_{(K+1,i)}^{\mathsf{T}} \widetilde{\boldsymbol{D}}_{(K+1,i)} \left( \mathbf{1} / \sqrt{K} \right)$$
(A.92)

$$\leq KL^2\left(\boldsymbol{D}, K\right). \tag{A.93}$$

Hence, (A.89) holds, which in turn shows (A.83). Thus, the proof is complete.

# Appendix B

# Appendix to Chapter 3

## B.1 Proof of Theorem 3.1

*Proof.* The proof is by showing that the conditions of the theorem are equivalent to the Kalman-type rank test. The proof relies on the fact that the Kalman rank test for the unconstrained system is equivalent to the PBH test is same as condition 1 of Theorem 3.1 [90].

We first prove that conditions of Theorem 3.1 imply the Kalman-type rank test. Suppose that the Kalman-type rank test fails. Then, we consider the following matrix of size  $N \times N\tilde{K}s$ :

$$\tilde{\boldsymbol{H}}^{*} = [\boldsymbol{D}^{\tilde{K}N-1}\boldsymbol{H}_{\mathcal{S}_{1}} \quad \boldsymbol{D}^{\tilde{K}N-2}\boldsymbol{H}_{\mathcal{S}_{1}} \quad \dots \quad \boldsymbol{D}^{(\tilde{K}-1)N}\boldsymbol{H}_{\mathcal{S}_{1}}$$

$$\dots \quad \boldsymbol{D}^{(\tilde{K}-1)N-1}\boldsymbol{H}_{\mathcal{S}_{2}} \quad \dots \quad \boldsymbol{D}^{(\tilde{K}-2)N}\boldsymbol{H}_{\mathcal{S}_{2}} \quad \dots$$

$$\dots \quad \boldsymbol{D}^{N-1}\boldsymbol{H}_{\mathcal{S}_{\tilde{K}}} \quad \dots \quad \boldsymbol{H}_{\mathcal{S}_{\tilde{K}}}], \quad (B.1)$$

where we define  $\tilde{K} \triangleq \lceil L/s \rceil$  index sets as follows:

$$|\mathcal{S}_i| = s, \quad \bigcup_{i=1}^{\tilde{K}} \mathcal{S}_i = \{1, 2, \dots, L\}.$$
(B.2)

Since the Kalman-type rank test fails,  $\tilde{\boldsymbol{H}}^*$  does not have full row rank for any finite K. Further, we can rearrange the columns of  $\tilde{\boldsymbol{H}}^*$  to get the following matrix which has the same rank as that of  $\tilde{\boldsymbol{H}}^*$ :  $\begin{bmatrix} \boldsymbol{D}^{N-1}\boldsymbol{H}^* & \boldsymbol{D}^{N-2}\boldsymbol{H}^* & \dots & \boldsymbol{H}^* \end{bmatrix}$ , where  $\boldsymbol{H}^* \in \mathbb{R}^{N \times \tilde{K}s}$  is defined as follows:

$$\boldsymbol{H}^* \triangleq \begin{bmatrix} \boldsymbol{D}^{(\tilde{K}-1)N} \boldsymbol{H}_{\mathcal{S}_1} & \boldsymbol{D}^{(\tilde{K}-2)N} \boldsymbol{H}_{\mathcal{S}_2} \dots & \boldsymbol{H}_{\mathcal{S}_{\tilde{K}}} \end{bmatrix}.$$
(B.3)

Thus, using the Kalman rank test for unconstrained inputs, the system with transfer matrix  $\boldsymbol{D}$  and input matrix  $\boldsymbol{H}^*$  is not controllable. Then, the PBH test for unconstrained inputs implies that the matrix  $\begin{bmatrix} \boldsymbol{D} - \lambda \boldsymbol{I} & \boldsymbol{H}^* \end{bmatrix} \in \mathbb{R}^{N \times N + \tilde{K}s}$  has rank less than N, for some  $\lambda \in \mathbb{C}$ . Therefore, there exists a nonzero vector  $\boldsymbol{z} \in \mathbb{R}^N$  such that  $\boldsymbol{z}^{\mathsf{T}}\boldsymbol{D} = \lambda \boldsymbol{z}^{\mathsf{T}}$  and  $\boldsymbol{z}^{\mathsf{T}}\boldsymbol{H}^* = \boldsymbol{0}$ . However, we have

$$\mathbf{0} = \boldsymbol{z}^{\mathsf{T}} \boldsymbol{H}^{*} = \boldsymbol{z}^{\mathsf{T}} \Big[ \boldsymbol{\lambda}^{(\tilde{K}-1)N} \boldsymbol{H}_{\mathcal{S}_{1}} \quad \boldsymbol{\lambda}^{(\tilde{K}-2)N} \boldsymbol{H}_{\mathcal{S}_{2}} \dots \quad \boldsymbol{H}_{\mathcal{S}_{\tilde{K}}} \Big] \cdot$$
(B.4)

So either  $\lambda = 0$  and  $\boldsymbol{z}^{\mathsf{T}} \boldsymbol{H}_{\mathcal{S}_{\tilde{K}}} = \boldsymbol{0}$ , or, if  $\lambda$  is nonzero, then  $\boldsymbol{z}^{\mathsf{T}} \boldsymbol{H} = \boldsymbol{0}$  because  $\boldsymbol{z}$  is orthogonal to all columns of  $\boldsymbol{H}$  due to (B.2). Repeating the same arguments for all possible index sets  $\{S_i\}$  satisfying (B.2), we get that for every index set  $S_i$  with s entries, there exists a vector  $\boldsymbol{z} \in \mathbb{R}^N$  such that  $\boldsymbol{z}^{\mathsf{T}} \boldsymbol{D} = \lambda \boldsymbol{z}^{\mathsf{T}}$ , and one of the following conditions hold:  $\lambda = 0$ and  $\boldsymbol{z}^{\mathsf{T}} \boldsymbol{H}_{S_i} = \boldsymbol{0}$ , or  $\boldsymbol{z}^{\mathsf{T}} \boldsymbol{H} = \boldsymbol{0}$ . Therefore, we get that one of the following cases hold:

- 1. There exists a left eigenvector  $\boldsymbol{z}$  of  $\boldsymbol{D}$ , such that  $\boldsymbol{z}^{\mathsf{T}}\boldsymbol{H} = \boldsymbol{0}$ . In this case, condition 1 of Theorem 3.1 does not hold.
- 2. For every index set S with s entries, there exists a nonzero vector  $\boldsymbol{z} \in \mathbb{R}^N$  such that  $\boldsymbol{z}^{\mathsf{T}} \boldsymbol{D} = \boldsymbol{0}$ , and  $\boldsymbol{z}^{\mathsf{T}} \boldsymbol{H}_S = \boldsymbol{0}$ . This implies that  $\boldsymbol{z}^{\mathsf{T}} \begin{bmatrix} \boldsymbol{D} & \boldsymbol{H}_S \end{bmatrix} = \boldsymbol{0}$ . Therefore, rank of  $\begin{bmatrix} \boldsymbol{D} & \boldsymbol{H}_S \end{bmatrix}$  is less than N, for every index set S. Thus, condition 2 of Theorem 3.1

does not hold.

Thus, when the Kalman-type rank test is unsuccessful, the conditions of the theorem are also violated.

Next, we prove that the Kalman-type rank test implies the conditions of the theorem. Suppose that the two conditions do not hold simultaneously. This could happen under the following two exhaustive cases:

- 1. Suppose that condition 1 does not hold. Then, the PBH test is violated, and the system is not controllable. When a system is not controllable, it cannot be sparse-controllable.
- 2. Suppose condition 2 does not hold. Then, for every index set S with s entries, there exists a nonzero vector z such that  $z^{\mathsf{T}}H_S = 0$  and  $z^{\mathsf{T}}D = 0$ . This implies that for any set of K index sets  $\{S_i : |S_i| = s\}_{i=1}^K$  there exists a nonzero vector  $z \in \mathbb{R}^N$  such that

$$\boldsymbol{z}^{\mathsf{T}} \begin{bmatrix} \boldsymbol{D}^{K-1} \boldsymbol{H}_{\mathcal{S}_1} & \boldsymbol{D}^{K-2} \boldsymbol{H}_{\mathcal{S}_2} & \dots & \boldsymbol{H}_{\mathcal{S}_K} \end{bmatrix} = \boldsymbol{0}.$$
 (B.5)

Hence, the Kalman-type rank test fails.

Thus, the proof is complete.

### B.2 Proof of Corollary 3.2

*Proof.* We first note that if a system is s-sparse-controllable, it is controllable. Hence, we need to prove that if a system with an invertible state transition matrix D is controllable, it is s-sparse-controllable, for any positive integer  $s \leq L$ .

If the system is controllable, the PBH test for unconstrained input succeeds, which implies that the condition 1 of Theorem 3.1 holds. Further, if D is invertible, the matrix  $\begin{bmatrix} D & H_{\mathcal{S}} \end{bmatrix}$ has rank N for any H and index set  $\mathcal{S}$ . Therefore, condition 2 of Theorem 3.1 always holds, for any  $s \leq L$ . Hence, from Theorem 3.1, the system is s-sparse-controllable. Thus, the desired result is proved.

### B.3 Proof of Theorem 3.3

Using the Kalman-type rank test, the minimum number of input vectors required to ensure controllability is the smallest integer K that satisfies rank condition of the test. So, for any finite K, we define  $\mathcal{H}_{(K)} \subseteq \mathbb{R}^{N \times Ks}$  as the set of submatrices of  $\tilde{H}_{(K)}$  of the following form (as defined for the Kalman-type rank test):

$$\begin{bmatrix} \boldsymbol{D}^{K-1}\boldsymbol{H}_{\mathcal{S}_1} & \boldsymbol{D}^{K-2}\boldsymbol{H}_{\mathcal{S}_2} & \dots & \boldsymbol{H}_{\mathcal{S}_K} \end{bmatrix}.$$
(B.6)

Here, the index set  $S_i$  has s entries, for i = 1, 2, ..., K. Also, we define the following:

$$R_{(K)}^* = \max_{\boldsymbol{H}_{(K)} \in \mathcal{H}_{(K)}} \operatorname{\mathsf{Rank}}\left\{\boldsymbol{H}_{(K)}\right\}.$$
(B.7)

$$\mathcal{H}_{(K)}^* = \left\{ \boldsymbol{H}_{(K)} \in \mathcal{H}_{(K)} : \mathsf{Rank}\left\{ \boldsymbol{H}_{(K)} \right\} = R_{(K)}^* \right\}$$
(B.8)

With these definitions,  $K^*$  is the smallest integer such that  $R^*_{(K^*)} = N$ .

Before starting the proof, we outline the main steps involved. At a high level, there are five steps to the proof:

1. We begin by showing that for any matrix  $H_{(K)} \in \mathcal{H}_{(K)}$ , we can find a matrix

 $\boldsymbol{H}_{(K)}^* \in \mathcal{H}_{(K)}^*$  such that

$$\mathcal{CS}\left\{\boldsymbol{H}_{(K)}\right\} \subseteq \mathcal{CS}\left\{\boldsymbol{H}_{(K)}^{*}\right\}.$$
(B.9)

2. Second, using the above claim, we show that if K is any integer such that

$$R_{(K)}^* = R_{(K+1)}^*, \tag{B.10}$$

then  $R^*_{(K+Q)} = R^*_{(K)}$ , for any positive integer Q.

- 3. Third, we prove that  $K^*$  is the smallest integer K such that (B.10) holds, which in turn leads to the upper bound:  $K^* \leq N + 1 - R^*_{H,s}$ , where  $R^*_{H,s}$  is as defined in the statement of the theorem.
- 4. Fourth, we show that in order to satisfy the rank criterion in (B.10),  $\boldsymbol{H}_{(K^*)}^*$  needs to contain at most  $qR_{\boldsymbol{H}}$  number of columns with a particular structure. Then, we provide a choice of index sets  $\{S_i\}_{i=1}^{K=q\lceil R_{\boldsymbol{H}}/s\rceil}$  which can lead to that particular structure. Since the smallest integer K that can achieve rank criterion in (B.10) is  $K^*$ , we assert that  $K^* \leq q\lceil R_{\boldsymbol{H}}/s\rceil$ . Thus, together with the above step, we establish the upper bound in the theorem.
- 5. Finally, we lower bound  $K^*$  to complete the proof.

## B.3.1 Characterizing $\mathcal{H}^*_{(K)}$

If  $\boldsymbol{H}_{(K)} \in \mathcal{H}_{(K)}^*$ , the result is trivial:  $\boldsymbol{H}_{(K)}^* = \boldsymbol{H}_{(K)}$ . Suppose that  $\boldsymbol{H}_{(K)} \notin \mathcal{H}_{(K)}^*$ , then Rank  $\{\boldsymbol{H}_{(K)}\} < R_{(K)}^*$ . Therefore, to find  $\boldsymbol{H}_{(K)}^*$ , we have to replace some linearly dependent columns of  $\boldsymbol{H}_{(K)}$  with columns which are linearly independent of the rest of the columns of  $\boldsymbol{H}_{(K)}$ , as follows:

- 1. Find a set  $\{u_i\}_{i=1}^{\mathsf{Rank}\{H_{(K)}\}}$  of columns of  $H_{(K)}$  that are linearly independent and span  $\mathcal{CS}\{H_{(K)}\}.$
- 2. Since  $\boldsymbol{H}_{(K)}$  is a submatrix of  $\tilde{\boldsymbol{H}}_{(K)}$ , we can extend the set  $\{\boldsymbol{u}_i\}_{i=1}^{\mathsf{Rank}\left\{\boldsymbol{H}_{(K)}\right\}}$  to form a basis  $\{\boldsymbol{u}_i\}_{i=1}^{\mathsf{Rank}\left\{\tilde{\boldsymbol{H}}_{(K)}\right\}}$  of  $\mathcal{CS}\left\{\tilde{\boldsymbol{H}}_{(K)}\right\}$  by adding columns from  $\tilde{\boldsymbol{H}}_{(K)}$ . We note that  $\boldsymbol{u}_i = \boldsymbol{D}^p \boldsymbol{H}_j$  for some integers p and j because of the structure of  $\tilde{\boldsymbol{H}}_{(K)}$ .
- 3. Replace the linearly dependent columns of  $\boldsymbol{H}_{(K)}$  with the columns from the set  $\{\boldsymbol{u}_i\}_{i=\mathsf{Rank}\{\boldsymbol{H}_{(K)}\}^{+1}}^{\mathsf{Rank}\{\tilde{\boldsymbol{H}}_{(K)}\}}$  to get a new matrix  $\bar{\boldsymbol{H}}_{(K)} \in \mathbb{R}^{N \times Ks}$ . We only replace a column of form  $\boldsymbol{D}^p \boldsymbol{H}_j$  in  $\boldsymbol{H}_{(K)}$  with another column of the form  $\boldsymbol{D}^p \boldsymbol{H}_{j'}$ , for all p and j and some integer j'. This ensures that  $\bar{\boldsymbol{H}}_{(K)} \in \mathcal{H}_{(K)}$ . In this fashion, we replace as many columns of  $\boldsymbol{H}_{(K)}$  as necessary to ensure that  $\bar{\boldsymbol{H}}_{(K)}$  has the maximum rank,  $R^*_{(K)}$ . However, since we are only replacing linearly dependent columns, we have

$$\mathcal{CS}\left\{\boldsymbol{H}_{(K)}\right\} \subseteq \mathcal{CS}\left\{\bar{\boldsymbol{H}}_{(K)}\right\}.$$
(B.11)

Since Rank  $\{\bar{\boldsymbol{H}}_{(K)}\} = R^*_{(K)}$  and  $\bar{\boldsymbol{H}}_{(K)} \in \mathcal{H}_{(K)}$ , we get that  $\bar{\boldsymbol{H}}_{(K)} \in \mathcal{H}^*_{(K)}$ , satisfying (B.11). Hence, the first step of the proof is complete.

## **B.3.2** Characterizing $R^*_{(K)}$

We use the proof by induction technique to show that  $R^*_{(K+Q)} = R^*_{(K)}$ , for any integer Q > 0. Hence, it suffices to show the following:

$$R^*_{(K+2)} = R^*_{(K+1)}.$$
(B.12)

From (B.7), we know that  $R^*_{(K+2)} \ge R^*_{(K+1)}$ . Also,

$$R_{(K)}^{*} = \max_{\boldsymbol{H}_{(K)} \in \mathcal{H}_{(K)}} \dim \left\{ \mathcal{CS} \left\{ \boldsymbol{H}_{(K)} \right\} \right\},$$
(B.13)

where dim $\{\cdot\}$  denotes the dimension of a subspace. Thus, we establish (B.12) by showing that for any matrix  $\boldsymbol{H}_{(K+2)} \in \mathcal{H}_{(K+2)}$ , there exists a matrix  $\boldsymbol{H}_{(K+1)}^* \in \mathcal{H}_{(K+1)}^*$  such that

$$\mathcal{CS}\left\{\boldsymbol{H}_{(K+2)}\right\} \subseteq \mathcal{CS}\left\{\boldsymbol{H}_{(K+1)}^{*}\right\}.$$
(B.14)

We prove this relation by separately looking at the column spaces spanned by the first s columns and the last (K + 1)s columns of  $\mathbf{H}_{(K+2)}$ . We know that the submatrix formed by the last (K + 1)s columns of any matrix in  $\mathcal{H}_{(K+2)}$  belongs to  $\mathcal{H}_{(K+1)}$ . Thus, using the claim in the first step, we can find a matrix  $\mathbf{H}^*_{(K+1)}$  such that the column space spanned by the last (K + 1)s columns of  $\mathbf{H}_{(K)}$  is contained in  $\mathcal{CS} \{\mathbf{H}^*_{(K+1)}\}$ . Therefore, it suffices to show that the column space spanned by the first s columns of  $\mathbf{H}_{(K+2)}$  is contained in the column space of the same matrix  $\mathbf{H}^*_{(K+1)}$ .

To prove the above statement, we note that the column space of the first *s* columns of  $\boldsymbol{H}_{(K+2)}$  is contained in  $\mathcal{CS} \{ \boldsymbol{D}^{K+1} \boldsymbol{H} \}$ . Also,  $\mathcal{CS} \{ \boldsymbol{H}_{(K+1)}^* \}$  contains the intersection  $\bigcap_{\boldsymbol{H}_{(K+1)}^* \in \mathcal{H}_{(K+1)}^*} \mathcal{CS} \{ \boldsymbol{H}_{(K+1)}^* \}$ . Hence, it suffices to show that

$$\mathcal{CS}\left\{\boldsymbol{D}^{K+1}\boldsymbol{H}\right\} \subseteq \bigcap_{\boldsymbol{H}_{(K+1)}^{*} \in \mathcal{H}_{(K+1)}^{*}} \mathcal{CS}\left\{\boldsymbol{H}_{(K+1)}^{*}\right\},$$
(B.15)

which we prove using the relation (B.10).

To show that (B.15) holds, we consider an index set  $S \subseteq \{1, 2, ..., L\}$  with s entries and a matrix  $\boldsymbol{H}_{(K)}^* \in \mathcal{H}_{(K)}^*$ . Now, the matrix  $\begin{bmatrix} \boldsymbol{D}^K \boldsymbol{H}_S & \boldsymbol{H}_{(K)}^* \end{bmatrix} \in \mathbb{R}^{N \times (K+1)s}$  belongs to  $\mathcal{H}_{(K+1)}$ . Thus, from (B.7) and (B.10) we have

$$\operatorname{\mathsf{Rank}}\left\{ \begin{bmatrix} \boldsymbol{D}^{K}\boldsymbol{H}_{\mathcal{S}} & \boldsymbol{H}_{(K)}^{*} \end{bmatrix} \right\} \leq R_{(K+1)}^{*} = R_{(K)}^{*}. \tag{B.16}$$

However, we also have

$$\operatorname{\mathsf{Rank}}\left\{ \begin{bmatrix} \boldsymbol{D}^{K}\boldsymbol{H}_{\mathcal{S}} & \boldsymbol{H}_{(K)}^{*} \end{bmatrix} \right\} \geq \operatorname{\mathsf{Rank}}\left\{ \boldsymbol{H}_{(K)}^{*} \right\} = R_{(K)}^{*}. \tag{B.17}$$

Thus, for all index sets S with s entries and any matrix  $H^*_{(K)} \in \mathcal{H}^*_{(K)}$ ,

$$\operatorname{\mathsf{Rank}}\left\{ \begin{bmatrix} \boldsymbol{D}^{K}\boldsymbol{H}_{\mathcal{S}} & \boldsymbol{H}_{(K)}^{*} \end{bmatrix} \right\} = \operatorname{\mathsf{Rank}}\left\{ \boldsymbol{H}_{(K)}^{*} \right\}$$
(B.18)

This relation immediately implies the following:

$$\operatorname{\mathsf{Rank}}\left\{ \begin{bmatrix} \boldsymbol{D}^{K}\boldsymbol{H} & \boldsymbol{H}_{(K)}^{*} \end{bmatrix} \right\} = \operatorname{\mathsf{Rank}}\left\{ \boldsymbol{H}_{(K)}^{*} \right\}, \qquad (B.19)$$

for any matrix  $\boldsymbol{H}_{(K)}^* \in \mathcal{H}_{(K)}^*$ . Thus, we get that the columns of  $\boldsymbol{D}^K \boldsymbol{H}$  belong to  $\mathcal{CS} \{ \boldsymbol{H}_{(K)}^* \}$ , for any matrix  $\boldsymbol{H}_{(K)}^* \in \mathcal{H}_{(K)}^*$ . Hence,

$$\mathcal{CS}\left\{\boldsymbol{D}^{K}\boldsymbol{H}\right\} \subseteq \underset{\boldsymbol{H}_{(K)}^{*}\in\mathcal{H}_{(K)}^{*}}{\cap} \mathcal{CS}\left\{\boldsymbol{H}_{(K)}^{*}\right\}.$$
(B.20)

Therefore, we get

$$\mathcal{CS}\left\{\boldsymbol{D}^{K+1}\boldsymbol{H}\right\} \subseteq \underset{\boldsymbol{H}_{(K)}^{*}\in\mathcal{H}_{(K)}^{*}}{\cap} \mathcal{CS}\left\{\boldsymbol{D}\boldsymbol{H}_{(K)}^{*}\right\}.$$
(B.21)

Hence, to prove (B.15), we need to show that

$$\bigcap_{\boldsymbol{H}_{(K)}^{*}\in\mathcal{H}_{(K)}^{*}}\mathcal{CS}\left\{\boldsymbol{DH}_{(K)}^{*}\right\}\subseteq\bigcap_{\boldsymbol{H}_{(K+1)}^{*}\in\mathcal{H}_{(K+1)}^{*}}\mathcal{CS}\left\{\boldsymbol{H}_{(K+1)}^{*}\right\}.$$
(B.22)

We prove the above relation by showing that there exists a matrix  $\boldsymbol{H}^*_{(K+1)} \in \mathcal{H}^*_{(K+1)}$ 

such that

$$\mathcal{CS}\left\{\boldsymbol{DH}_{(K)}^{*}\right\} \subseteq \mathcal{CS}\left\{\boldsymbol{H}_{(K+1)}^{*}\right\},\tag{B.23}$$

for every matrix  $\boldsymbol{H}_{(K)}^* \in \mathcal{H}_{(K)}^*$ . So we consider a new matrix  $\bar{\boldsymbol{H}}_{(K+1)} \in \mathbb{R}^{N \times (K+1)s}$  as follows:

$$\bar{\boldsymbol{H}}_{(K+1)} \triangleq \begin{bmatrix} \boldsymbol{D}\boldsymbol{H}_{(K)}^* & \boldsymbol{H}_{\mathcal{S}} \end{bmatrix}, \qquad (B.24)$$

for some index set  $S \subseteq \{1, 2, ..., L\}$  and |S| = s. Since  $\overline{H}_{(K+1)} \in \mathcal{H}_{(K+1)}$ , using the arguments in the first step, we can find a matrix  $H^*_{(K+1)} \in \mathcal{H}^*_{(K+1)}$  such that

$$\mathcal{CS}\left\{\bar{\boldsymbol{H}}_{(K+1)}\right\} \subseteq \mathcal{CS}\left\{\boldsymbol{H}_{(K+1)}^{*}\right\}.$$
(B.25)

However, (B.24) implies that

$$\mathcal{CS}\left\{\boldsymbol{DH}_{(K)}^{*}\right\} \subseteq \mathcal{CS}\left\{\bar{\boldsymbol{H}}_{(K+1)}\right\}.$$
(B.26)

Therefore, (B.23) holds, and hence (B.22) is proved.

Recall that (B.22) implies (B.15), which in turn establishes the relation (B.12). By mathematical induction, we conclude that  $\operatorname{Rank} \{ H^*_{(K+Q)} \} = \operatorname{Rank} \{ H^*_{(K)} \}$ , for any positive integer Q, and the proof of the second step in the outline is complete.

#### B.3.3 First part of the upper bound

Suppose that  $K_*$  is the smallest integer such that  $R^*_{(K_*)} = R^*_{(K_*+1)}$ . From (B.7), it is clear that

$$R_{(K)}^* \le R_{(K+1)}^* \le N,$$
 (B.27)

for any positive integer K. Since  $R_{(K^*)}^* = N$ , we have  $R_{(K^*)}^* = R_{(K^*+1)}^* = N$ . Therefore,  $K_* \leq K^*$ , and  $R_{(K_*)}^* = N$  from the claim in the second step.

Further, we know that  $K^*$  is the smallest integer such that  $R^*_{(K^*)} = N$ . Therefore,  $K_* = K^*$ . Hence, we conclude that  $R^*_{(K)}$  strictly increases with K, for  $1 \le K \le K^*$ . Therefore, we have

$$N = R^*_{(K^*)} \ge R^*_{(K^*-1)} + 1 \ge R^*_{(K^*-2)} + 2$$
(B.28)

$$\geq R_{(1)}^* + K^* - 1 \tag{B.29}$$

$$= R_{H,s}^* + K^* - 1. \tag{B.30}$$

Hence, the third step in the outline is complete.

#### **B.3.4** Upper bounding $K^*$

To prove that  $K^* \leq q \lceil R_H/s \rceil$ , we first look at the linearly independent columns in  $H^*_{(K^*)}$ . We note that for any K, each column of  $H^*_{(K)}$  is of the form  $D^p H_j$ , for some integer p, and  $j \in \{1, 2, ..., L\}$ . However, since q is the degree of the minimal polynomial of D, then for any integer Q,  $D^p$  can be expressed as a linear combination of  $\{D^i\}_{i=Q}^{Q+q-1}$ , for all  $p \geq Q$ . Therefore, for any j, if  $\{D^i H_j \in \mathbb{R}^N\}_{i=Q}^{Q+q-1}$  are any q columns of  $H^*_{(K)}$ , further adding columns of the form  $D^p H_j$ , for  $p \geq Q$  does not improve the rank of the matrix. Therefore, for a given j, at most q columns of the form  $D^p H_j$  need to be present in  $H^*_{(K)}$ to ensure the rank criterion in (B.10).

Further, let  $\boldsymbol{H}_{\mathcal{S}'}$  represents the set of  $R_{\boldsymbol{H}}$  linearly independent columns of  $\boldsymbol{H}$ , i.e.,  $\mathcal{S}' \subseteq \{1, 2, \ldots, L\}$  and  $|\mathcal{S}'| = R_{\boldsymbol{H}}$ . Then, for any integer p, if  $\{\boldsymbol{D}^p \boldsymbol{H}_j \in \mathbb{R}^N\}_{j \in \mathcal{S}'}$  are any  $R_{\boldsymbol{H}}$  columns of  $\boldsymbol{H}^*_{(K)}$ , further adding columns of the form  $\boldsymbol{D}^p \boldsymbol{H}_j$ , for  $j \notin \mathcal{S}'$  does not improve

the rank of the matrix. Therefore, for any given p, at most  $R_H$  columns of the form  $D^p H_j$ need to be present in  $H^*_{(K)}$  to ensure the rank criterion.

In short, we have proved that, in order to ensure the rank criterion in (B.10),  $\boldsymbol{H}_{(K)}^*$  needs to have at most q columns of the form  $\boldsymbol{D}^p \boldsymbol{H}_j$ , for any given j, and at most  $R_{\boldsymbol{H}}$  columns of the form  $\boldsymbol{D}^p \boldsymbol{H}_j$ , for any given p. Hence,  $\boldsymbol{H}_{(K)}^*$  needs to have at most  $qR_{\boldsymbol{H}}$  columns to satisfy the rank criterion in (B.10).

We complete the proof for the upper bound by providing a choice of index sets for each input vector, that satisfies the above conditions. We form index sets  $\{S'_i\}_{i=1}^{K=\lceil R_H/s\rceil}$  that partition the set of  $R_H$  linearly independent columns into groups of size at most s. The index sets are selected such that  $\bigcup_{j=1}^K S'_i = S'$ ,  $|S_i| = s$ , and  $S_K$  is such that  $\begin{bmatrix} D & H_{S_K} \end{bmatrix}$  has rank N. The existence of such an index set  $S_K$  is ensured by condition 2 of Theorem 3.1. Also, we note that due to the condition  $|S_i| = s$ , the index sets need not be disjoint. Next, we choose  $S_i = S'_j$ , for  $i = (j-1)q + 1, (j-1)q + 2, \ldots, jq$ . Hence, we get the following  $N \times qKs$  submatrix of  $\tilde{H}_{(K)} \in \mathbb{R}^{N \times qKL}$ :

$$\boldsymbol{H}_{(K)}^{*} = [\boldsymbol{D}^{Kq-1}\boldsymbol{H}_{\mathcal{S}_{1}} \quad \boldsymbol{D}^{Kq-2}\boldsymbol{H}_{\mathcal{S}_{1}} \quad \dots \quad \boldsymbol{D}^{(K-1)q}\boldsymbol{H}_{\mathcal{S}_{1}}$$
$$\dots \quad \boldsymbol{D}^{(K-1)q-1}\boldsymbol{H}_{\mathcal{S}_{2}} \quad \dots \quad \boldsymbol{D}^{(K-2)q}\boldsymbol{H}_{\mathcal{S}_{2}} \quad \dots$$
$$\dots \quad \boldsymbol{D}^{q-1}\boldsymbol{H}_{\mathcal{S}_{K}} \quad \dots \quad \boldsymbol{H}_{\mathcal{S}_{K}}]. \quad (B.31)$$

It is easy to see that this choice of index sets ensures that for any given p,  $R_H$  columns of the form  $D^p H_j$  are present in  $H^*_{(K)}$ . Also, for any given  $j \in S'$ , q columns of  $\{D^i H_j \in \mathbb{R}^N\}_{i=Q}^{Q+q-1}$  are present in  $H^*_{(K)}$ . Hence,  $K^* \leq q \lceil R_H/s \rceil$ , which establishes the upper bound in (3.15).

### **B.3.5** Lower bounding $K^*$

The lower bound is achieved when all columns of  $\boldsymbol{H}_{(K)}^*$  are linearly independent. Thus, to ensure that rank  $\boldsymbol{H}_{(K)}^*$  is  $N, Ks \geq N$ . However, if  $s \geq R_H$ , the maximum number of independent columns become  $KR_H$ , and thus we get that  $KR_H \geq N$ . Hence, the lower bound in (3.15) is proved.

As noted in the proof outline, this suffices to establish Theorem 3.3.

# Appendix C

# Appendix to Chapter 5

### C.1 Proof of Proposition 5.1

We first prove a lemma to show that the noise term  $e_k$  is bounded, which then enables us to establish the required result.

**Lemma C.1.** In our online algorithm given by (5.2),  $\lim_{k\to\infty} \sum_{t=1}^{k} \frac{1}{t} \boldsymbol{e}_t$  exists and is finite. *Proof.* We define  $\boldsymbol{l}_k = \sum_{t=1}^{k} \frac{1}{t} \boldsymbol{e}_t$ , and  $\mathcal{F}_k$  as the  $\sigma$ -algebra generated by  $\boldsymbol{y}^k$ . Then,  $\mathbb{E} \{ \boldsymbol{l}_k | \mathcal{F}_{k-1} \} = \mathbb{E} \{ \boldsymbol{l}_{k-1} | \mathcal{F}_{k-1} \} + \frac{1}{k} \mathbb{E} \{ \boldsymbol{e}_k | \mathcal{F}_{k-1} \} = \boldsymbol{l}_{k-1}$ . Thus,  $\boldsymbol{l}_{k-1}$  is a martingale. Further, using the orthogonality property of martingales [181],

$$\mathbb{E}\left\{\|\boldsymbol{l}_{k}\|^{2}\right\} = \sum_{t=1}^{k} \mathbb{E}\left\{\|\boldsymbol{l}_{t} - \boldsymbol{l}_{t-1}\|^{2}\right\} = \sum_{t=1}^{k} \frac{1}{t^{2}} \mathbb{E}\left\{\|\boldsymbol{e}_{t}\|^{2}\right\}.$$
(C.1)

We note that  $\|\boldsymbol{y}\|_{\infty} < \infty$  a.s., thus (5.8) shows that  $\|\boldsymbol{e}_t\| < \infty$  a.s., if  $\|\boldsymbol{\gamma}_{k-1}\|_{\infty} < \infty$ . When  $\|\boldsymbol{\gamma}_{k-1}\|_{\infty} \to \infty$ , from (5.8), it can be shown that

$$\lim_{\|\boldsymbol{\gamma}\|_{\infty}\to\infty} \boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} \left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}}+\boldsymbol{R}\right)^{-1} = \lim_{\|\boldsymbol{\gamma}\|_{\infty}\to\infty} \|\boldsymbol{\gamma}\|_{\infty}^{-\frac{1}{2}} \boldsymbol{\Gamma}^{\frac{1}{2}} \left[\boldsymbol{R}^{-\frac{1}{2}} \boldsymbol{A} \left(\|\boldsymbol{\gamma}\|_{\infty}^{-1} \boldsymbol{\Gamma}\right) \boldsymbol{A}^{\mathsf{T}} \boldsymbol{R}^{-\frac{1}{2}}\right]^{\dagger} \boldsymbol{R}^{-\frac{1}{2}}. \quad (C.2)$$

Hence, all entries of  $\lim_{\gamma\to\infty} \Gamma A^{\mathsf{T}} (A\Gamma A^{\mathsf{T}} + R)^{-1}$  are finite, and  $||e_t|| < \infty$  with probability one. Thus,  $\mathbb{E} \{ ||e_t||^2 \}$  is bounded, and hence by Jensen's inequality and (C.1), the martingale is bounded in  $\mathcal{L}^1$ . Applying Doob's forward convergence theorem [181] to each coordinate of the martingale  $l_k[i], i = 1, 2, ..., N$ , the limit  $\lim_{k\to\infty} l_k = \lim_{k\to\infty} \sum_{t=1}^k \frac{1}{t} e_t$  exists, and is finite.

We now formally prove Proposition 5.1.

*Proof.* Using (5.2), we have,

$$\boldsymbol{\gamma}_{k} = \frac{k-1}{k} \boldsymbol{\gamma}_{k-1} + \frac{1}{k} \operatorname{Diag} \left\{ \boldsymbol{P}(\boldsymbol{\gamma}_{k-1}) + \widehat{\boldsymbol{x}}(\boldsymbol{y}_{k}, \boldsymbol{\gamma}_{k-1}) \widehat{\boldsymbol{x}}(\boldsymbol{y}_{k}, \boldsymbol{\gamma}_{k-1})^{\mathsf{T}} \right\}.$$
(C.3)

All entries of  $\text{Diag}\left\{\boldsymbol{P}(\boldsymbol{\gamma}_{k-1}) + \hat{\boldsymbol{x}}(\boldsymbol{y}_{k}, \boldsymbol{\gamma}_{k-1})\hat{\boldsymbol{x}}(\boldsymbol{y}_{k}, \boldsymbol{\gamma}_{k-1})^{\mathsf{T}}\right\}$  are nonnegative. This ensures that  $\boldsymbol{\gamma}_{k}[i] \geq 0$  for  $i = 1, 2, \ldots, N$  and  $\forall k$ , if  $\boldsymbol{\gamma}_{0}$  is a nonnegative vector. Thus, the sequence  $\boldsymbol{\gamma}_{k}$  is bounded from below.

Next, we use [182, Theorem 7] to show that the sequence is bounded from above, and hence it remains in a compact set. For that, we check if the conditions below hold in our case:

- (i) The function  $\boldsymbol{f}$  is Lipschitz
- (ii)  $\lim_{k\to\infty} \sum_{t=1}^k \frac{1}{t} \boldsymbol{e}_t$  exists
- (iii) The function  $f_{\infty}(\gamma) = \lim_{c \to \infty} f(c\gamma)/c$  is continuous, and the ordinary differential equation (ODE)

$$\frac{d}{dt}\boldsymbol{\gamma}(t) = \boldsymbol{f}_{\infty}(\boldsymbol{\gamma}(t)) \tag{C.4}$$

has the origin as its unique globally asymptotic stable equilibrium.

Since  $P(\gamma)$  and  $\Gamma A^{\mathsf{T}} (A \Gamma A^{\mathsf{T}} + R)^{-1} A \Gamma$  are positive semidefinite, all of their diagonal entries are nonnegative. Hence, using (5.3),

$$\boldsymbol{f}(\boldsymbol{\gamma}) \geq -\boldsymbol{\gamma} + \mathsf{Diag}\left\{\boldsymbol{P}(\boldsymbol{\gamma})\boldsymbol{A}^{\mathsf{T}}\boldsymbol{R}^{-1}\mathbb{E}\left\{\boldsymbol{y}\boldsymbol{y}^{\mathsf{T}}\right\}\boldsymbol{R}^{-1}\boldsymbol{A}\boldsymbol{P}(\boldsymbol{\gamma})\right\} \geq -\boldsymbol{\gamma}, \tag{C.5}$$

where  $\boldsymbol{a} \geq \boldsymbol{b}$  denotes that every entry of  $\boldsymbol{a}$  is greater than or equal to the corresponding entry of  $\boldsymbol{b}$ . Further, since the matrix  $\Gamma \boldsymbol{A}^{\mathsf{T}} \left( \boldsymbol{A} \Gamma \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right)^{-1} \boldsymbol{A} \Gamma$  is positive semidefinite, every diagonal entry of  $\boldsymbol{P}(\boldsymbol{\gamma}) = \Gamma - \Gamma \boldsymbol{A}^{\mathsf{T}} \left( \boldsymbol{A} \Gamma \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right)^{-1} \boldsymbol{A} \Gamma$  is less than the corresponding diagonal entry of  $\boldsymbol{\Gamma}$ . Thus, we get

$$\begin{aligned} \boldsymbol{f}(\boldsymbol{\gamma}) &\leq \mathsf{Diag}\left\{\boldsymbol{P}(\boldsymbol{\gamma})\boldsymbol{A}^{\mathsf{T}}\boldsymbol{R}^{-1}\mathbb{E}\left\{\boldsymbol{y}\boldsymbol{y}^{\mathsf{T}}\right\}\boldsymbol{R}^{-1}\boldsymbol{A}\boldsymbol{P}(\boldsymbol{\gamma})\right\} \\ &\leq \lambda\mathsf{Diag}\left\{\boldsymbol{P}(\boldsymbol{\gamma})\boldsymbol{A}^{\mathsf{T}}\boldsymbol{R}^{-2}\boldsymbol{A}\boldsymbol{P}(\boldsymbol{\gamma})\right\} \end{aligned} \tag{C.6}$$

where  $\lambda$  is the largest eigenvalue of the positive semidefinite matrix  $\mathbb{E} \{ yy^{\mathsf{T}} \}$ , and  $a \leq b$  denotes an entry-wise inequality. Thus,

$$-\boldsymbol{\gamma}[i] \leq \boldsymbol{f}(\boldsymbol{\gamma})[i] \leq \lambda \mathsf{Diag}\left\{\boldsymbol{P}(\boldsymbol{\gamma})\boldsymbol{A}^{\mathsf{T}}\boldsymbol{R}^{-2}\boldsymbol{A}\boldsymbol{P}(\boldsymbol{\gamma})\right\}[i], \tag{C.7}$$

for i = 1, 2, ..., N. To further bound the last term of the inequality, we use (5.6) to get

$$\boldsymbol{P}(\boldsymbol{\gamma})\boldsymbol{A}^{\mathsf{T}}\boldsymbol{R}^{-2}\boldsymbol{A}\boldsymbol{P}(\boldsymbol{\gamma}) = \boldsymbol{\Gamma}^{\frac{1}{2}}\boldsymbol{B}\left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}}+\boldsymbol{R}\right)^{-1}\boldsymbol{B}^{\mathsf{T}}\boldsymbol{\Gamma}^{\frac{1}{2}}.$$
 (C.8)

where  $\boldsymbol{B} \triangleq \boldsymbol{\Gamma}^{\frac{1}{2}} \boldsymbol{A}^{\mathsf{T}} \left( \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right)^{-\frac{1}{2}}$ . This implies

Diag 
$$\{ \boldsymbol{P}(\boldsymbol{\gamma}) \boldsymbol{A}^{\mathsf{T}} \boldsymbol{R}^{-2} \boldsymbol{A} \boldsymbol{P}(\boldsymbol{\gamma}) \} [i] = \boldsymbol{\gamma}[i] \boldsymbol{B}[i]^{\mathsf{T}} (\boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R})^{-1} \boldsymbol{B}[i]$$
 (C.9)

$$\leq \boldsymbol{\gamma}[i]\boldsymbol{B}[i]^{\mathsf{T}}\boldsymbol{R}^{-1}\boldsymbol{B}[i],$$
 (C.10)

where  $\boldsymbol{B}[i] \in \mathbb{R}^N$  is the *i*<sup>th</sup> column of  $\boldsymbol{B}^{\mathsf{T}}$ . Then, we have

$$\boldsymbol{B}\boldsymbol{B}^{\mathsf{T}} = \boldsymbol{\Gamma}^{\frac{1}{2}}\boldsymbol{A}^{\mathsf{T}} \left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right)^{-1} \boldsymbol{A}\boldsymbol{\Gamma}^{\frac{1}{2}}$$
$$= \boldsymbol{I} - \left(\boldsymbol{I} + \boldsymbol{\Gamma}^{\frac{1}{2}}\boldsymbol{A}^{\mathsf{T}}\boldsymbol{R}^{-1}\boldsymbol{A}\boldsymbol{\Gamma}^{\frac{1}{2}}\right)^{-1}.$$
(C.11)

This shows that  $I - BB^{\mathsf{T}}$  is a positive semidefinite matrix, and its diagonal entries are nonnegative. Thus,  $B[i]^{\mathsf{T}}B[i] \leq 1$ , for i = 1, 2, ..., N. Hence, we get

Diag 
$$\{ \boldsymbol{P}(\boldsymbol{\gamma}) \boldsymbol{A}^{\mathsf{T}} \boldsymbol{R}^{-2} \boldsymbol{A} \boldsymbol{P}(\boldsymbol{\gamma}) \} [i] \leq \bar{\lambda} \boldsymbol{\gamma}[i],$$
 (C.12)

where  $\bar{\lambda}$  is the largest eigenvalue of  $\mathbf{R}^{-1}$ . Substituting this relation in (C.7), we get

$$-\boldsymbol{\gamma}[i] \le \boldsymbol{f}(\boldsymbol{\gamma})[i] \le \bar{\lambda}\lambda\boldsymbol{\gamma}[i]. \tag{C.13}$$

Thus, (i) is satisfied. The assumption (ii) is true by Lemma C.1. To check (iii), we start with (5.7) to get

$$\boldsymbol{f}_{\infty}(\boldsymbol{\gamma}) = \lim_{c \to \infty} \frac{1}{c} \operatorname{Diag} \left\{ c^{2} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} (c \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R})^{-1} (\mathbb{E} \left\{ \boldsymbol{y}_{k} \boldsymbol{y}_{k}^{\mathsf{T}} \right\} - c \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} - \boldsymbol{R}) \left( c \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right)^{-1} \boldsymbol{A} \boldsymbol{\Gamma} \right\} \quad (C.14)$$

$$= -\lim_{c \to \infty} \operatorname{Diag} \left\{ \boldsymbol{\Gamma} \left( \boldsymbol{R}^{-\frac{1}{2}} \boldsymbol{A} \boldsymbol{\Gamma}^{\frac{1}{2}} \right)^{\mathsf{T}} \left[ \boldsymbol{R}^{-\frac{1}{2}} \boldsymbol{A} \boldsymbol{\Gamma}^{\frac{1}{2}} \left( \boldsymbol{R}^{-\frac{1}{2}} \boldsymbol{A} \boldsymbol{\Gamma}^{\frac{1}{2}} \right)^{\mathsf{T}} + \boldsymbol{I}/c \right]^{-1} \boldsymbol{R}^{-\frac{1}{2}} \boldsymbol{A} \boldsymbol{\Gamma}^{\frac{1}{2}} \right\} \quad (C.15)$$

$$= -\mathsf{Diag}\left\{\Gamma\left(\mathbf{R}^{-\frac{1}{2}}\mathbf{A}\Gamma^{\frac{1}{2}}\right)^{\dagger}\left(\mathbf{R}^{-\frac{1}{2}}\mathbf{A}\Gamma^{\frac{1}{2}}\right)\right\}.$$
 (C.16)

Note that  $\operatorname{Rank}\{(\mathbf{R}^{-\frac{1}{2}}\mathbf{A}\mathbf{\Gamma}^{\frac{1}{2}})\} = \min\{\operatorname{Rank}\{\mathbf{\Gamma}\}, m\}$ . For the case when  $\operatorname{Rank}\{\mathbf{\Gamma}\} < m$ , we have  $\operatorname{Rank}\{(\mathbf{R}^{-\frac{1}{2}}\mathbf{A}\mathbf{\Gamma}^{\frac{1}{2}})\} = \operatorname{Rank}\{\mathbf{\Gamma}\}$ , and thus,  $\mathbf{f}_{\infty}(\boldsymbol{\gamma}) = -\boldsymbol{\gamma}$ . Since **0** is the only globally asymptotically stable equilibrium of the ODE  $\frac{d}{dt}\boldsymbol{\gamma}(t) = -\boldsymbol{\gamma}(t)$ , (iii) holds. When

 $\operatorname{Rank}\{\boldsymbol{R}^{-\frac{1}{2}}\boldsymbol{A}\boldsymbol{\Gamma}^{\frac{1}{2}}\}=m,$ 

$$\left(\boldsymbol{R}^{-\frac{1}{2}}\boldsymbol{A}\boldsymbol{\Gamma}^{\frac{1}{2}}\right)^{\dagger} = \boldsymbol{\Gamma}^{\frac{1}{2}}\boldsymbol{A}^{\mathsf{T}}\boldsymbol{R}^{-\frac{1}{2}}\left(\boldsymbol{R}^{-\frac{1}{2}}\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}}\boldsymbol{R}^{-\frac{1}{2}}\right)^{-1}, \qquad (C.17)$$

which implies the following:

$$(\boldsymbol{R}^{-\frac{1}{2}}\boldsymbol{A}\boldsymbol{\Gamma}^{\frac{1}{2}})^{\dagger}(\boldsymbol{R}^{-\frac{1}{2}}\boldsymbol{A}\boldsymbol{\Gamma}^{\frac{1}{2}}) = \boldsymbol{\Gamma}^{\frac{1}{2}}\boldsymbol{A}^{\mathsf{T}}\left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}}\right)^{-1}\boldsymbol{A}\boldsymbol{\Gamma}^{\frac{1}{2}}.$$
 (C.18)

Since the diagonal entries of  $\mathbf{A}^{\mathsf{T}} (\mathbf{A} \Gamma \mathbf{A}^{\mathsf{T}})^{-1} \mathbf{A}$  are positive, the only possible equilibrium for the ODE is **0**. However, when  $\gamma = \mathbf{0}$ , Rank $\{\mathbf{R}^{\frac{1}{2}}\mathbf{A}\Gamma^{\frac{1}{2}}\} \neq m$  which is a contradiction. Hence, there is no equilibrium point with Rank $\{\mathbf{R}^{\frac{1}{2}}\mathbf{A}\Gamma^{\frac{1}{2}}\} = m$ . Thus, (iii) holds, and the proof is complete.

## C.2 Proof of Theorem 5.1

Before we prove the main theorem, we need two lemmas.

Lemma C.2. The solution set of  $f(\gamma) = 0$  is  $\{0\} \cup \{\gamma \in \mathbb{R}^N : A\Gamma A^{\mathsf{T}} = A\Gamma_{\mathsf{opt}}A^{\mathsf{T}}\},$ when  $\mathbb{E}\{yy^{\mathsf{T}}\} = A\Gamma_{\mathsf{opt}}A^{\mathsf{T}} + R.$ 

*Proof.* From (5.7), we get

$$\boldsymbol{f}(\boldsymbol{\gamma}) = \operatorname{Diag}\left\{\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}}\left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right)^{-1}\boldsymbol{A}\left(\boldsymbol{\Gamma}_{\mathsf{opt}} - \boldsymbol{\Gamma}\right)\boldsymbol{A}^{\mathsf{T}}\left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right)^{-1}\boldsymbol{A}\boldsymbol{\Gamma}\right\}.$$
(C.19)

Clearly,  $\gamma = 0$  is a zero of  $f(\gamma)$ . Let us consider the solutions whose support is the vector  $s \in \{0, 1\}^N$  and  $s \neq 0$ , and let the number of nonzero entries in s be denoted by s. The union of the solutions over all possible supports gives the solution set. Let  $\gamma_s \in \mathbb{R}^{s \times 1}$  be the vector of nonzero entries of  $\gamma$  and  $A_s \in \mathbb{R}^{m \times s}$  be the matrix formed by restricting A

to the *s* columns corresponding to the support *s*. Let  $B_s = (A\Gamma A^{\mathsf{T}} + R)^{-\frac{1}{2}} A_s \in \mathbb{R}^{m \times s}$ , and  $B = (A\Gamma A^{\mathsf{T}} + R)^{-\frac{1}{2}} A \in \mathbb{R}^{m \times N}$ . Then, the reduced set of equations corresponding to  $f(\gamma) = 0$  is given by

$$\operatorname{Diag}\left\{\boldsymbol{B}_{\boldsymbol{s}}^{\mathsf{T}}\boldsymbol{B}_{\boldsymbol{s}}\boldsymbol{\Gamma}_{\boldsymbol{s}}\boldsymbol{B}_{\boldsymbol{s}}^{\mathsf{T}}\boldsymbol{B}_{\boldsymbol{s}}\right\} = \operatorname{Diag}\left\{\boldsymbol{B}_{\boldsymbol{s}}^{\mathsf{T}}\boldsymbol{B}\boldsymbol{\Gamma}_{\mathsf{opt}}\boldsymbol{B}^{\mathsf{T}}\boldsymbol{B}_{\boldsymbol{s}}\right\},\tag{C.20}$$

where  $\Gamma_s = \text{Diag} \{\gamma_s\}$  is an invertible matrix. We note that the above system of equations is linear in the vector  $\gamma_s$ , for any given fixed matrices  $B_s$  and B. However,  $\text{Diag} \{B_s^{\mathsf{T}} B_s \Gamma_s B_s^{\mathsf{T}} B_s\} = (B_s^{\mathsf{T}} B_s) \circ (B_s^{\mathsf{T}} B_s) \gamma_s$ , where  $\circ$  represents the Hadamard product of matrices. Thus, the solution set of the system of equations is an affine space  $\mathcal{U}_s$  of dimension given by

$$\dim(\mathcal{U}_s) = s - \mathsf{Rank}\left\{\left(\boldsymbol{B}_s^\mathsf{T}\boldsymbol{B}_s\right) \circ \left(\boldsymbol{B}_s^\mathsf{T}\boldsymbol{B}_s\right)\right\}$$
(C.21)

$$= s - \mathsf{Rank} \left\{ \left( \boldsymbol{B}_{\boldsymbol{s}} \odot \boldsymbol{B}_{\boldsymbol{s}} \right)^{\mathsf{T}} \left( \boldsymbol{B}_{\boldsymbol{s}} \odot \boldsymbol{B}_{\boldsymbol{s}} \right) \right\}$$
(C.22)

$$= s - \mathsf{Rank} \left\{ \boldsymbol{B}_{\boldsymbol{s}} \odot \boldsymbol{B}_{\boldsymbol{s}} \right\}.$$
(C.23)

We now consider another affine space  $\mathcal{W}_s$  of dimension  $s - \mathsf{Rank} \{ B_s \odot B_s \}$  given by the set of  $\gamma_s$  satisfying

$$\operatorname{vec}\left\{\boldsymbol{B}_{\boldsymbol{s}}\boldsymbol{\Gamma}_{\boldsymbol{s}}\boldsymbol{B}_{\boldsymbol{s}}^{\mathsf{T}}\right\} = \left(\boldsymbol{B}_{\boldsymbol{s}}\odot\boldsymbol{B}_{\boldsymbol{s}}\right)\boldsymbol{\gamma}_{\boldsymbol{s}} = \operatorname{vec}\left\{\boldsymbol{B}\boldsymbol{\Gamma}_{\mathsf{opt}}\boldsymbol{B}^{\mathsf{T}}\right\}.$$
(C.24)

It is easy to see that  $\mathcal{W}_s \subseteq \mathcal{U}_s$  and  $\dim(\mathcal{U}_s) = \dim(\mathcal{W}_s)$ , which implies  $\mathcal{W}_s = \mathcal{U}_s$ . Rearranging, we get, for  $\gamma_s \in \mathcal{U}_s$ ,

$$(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R})^{-\frac{1}{2}} \boldsymbol{A}_{\boldsymbol{s}}\boldsymbol{\Gamma}_{\boldsymbol{s}}\boldsymbol{A}_{\boldsymbol{s}}^{\mathsf{T}} (\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R})^{-\frac{1}{2}}$$
$$= (\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R})^{-\frac{1}{2}} \boldsymbol{A}\boldsymbol{\Gamma}_{\mathsf{opt}}\boldsymbol{A}^{\mathsf{T}} (\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R})^{-\frac{1}{2}}. \quad (C.25)$$

Thus,

$$\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} = \boldsymbol{A}_{\boldsymbol{s}}\boldsymbol{\Gamma}_{\boldsymbol{s}}\boldsymbol{A}_{\boldsymbol{s}}^{\mathsf{T}} = \boldsymbol{A}\boldsymbol{\Gamma}_{\mathsf{opt}}\boldsymbol{A}^{\mathsf{T}}, \qquad (C.26)$$

and  $\mathcal{U}_{s} \subseteq \{\gamma : A\Gamma A^{\mathsf{T}} = A\Gamma_{\mathsf{opt}}A^{\mathsf{T}}\}$ , for all support sets  $s \neq 0$ . From (C.19), it is easy to see that  $\{\gamma \in \mathbb{R}^{N} : A(\Gamma - \Gamma_{\mathsf{opt}})A^{\mathsf{T}} = 0\}$  satisfies  $f(\gamma) = 0$ . Therefore,  $\bigcup_{s \in \{0,1\}^{N} \setminus 0} \mathcal{U}_{s} =$  $\{\gamma : A\Gamma A^{\mathsf{T}} = A\Gamma_{\mathsf{opt}}A^{\mathsf{T}}\}$ . Thus, we get that the solution set of  $f(\gamma) = 0$  is  $\{0\} \cup \{\gamma \in \mathbb{R}^{N} : A(\Gamma - \Gamma_{\mathsf{opt}})A^{\mathsf{T}} = 0\}$ .

We define some notation to state the next lemma. The notation  $X \succ 0$  denotes that X is a positive definite matrix and  $X \succeq 0$  denotes that X is a positive semidefinite matrix.

**Lemma C.3.** The set  $\mathbb{O} = \{ \boldsymbol{\gamma} \in \mathbb{R}^N : \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \succ \boldsymbol{0} \}$  is an open set and its closure is  $\{ \boldsymbol{\gamma} \in \mathbb{R}^N : \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \succeq \boldsymbol{0} \}.$ 

Proof. Let  $\gamma \in \mathbb{O}$ . Then,  $\boldsymbol{u}^{\mathsf{T}}(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R})\boldsymbol{u} > 0 \ \forall \boldsymbol{u} \in \mathbb{R}^m \setminus \{\boldsymbol{0}\}$ , and the minimum eigenvalue of  $\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}$  is strictly greater than some  $\beta > 0$ . We need to show that there exists an  $\epsilon > 0$  such that  $\boldsymbol{A}\boldsymbol{\widetilde{\Gamma}}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}$  is positive definite for all  $\boldsymbol{\widetilde{\gamma}}$  in the  $\epsilon$ -neighborhood of  $\gamma$ , i.e.,  $\|\boldsymbol{\gamma} - \boldsymbol{\widetilde{\gamma}}\| < \epsilon$ .

For a given  $\boldsymbol{u} \in \mathbb{R}^m \setminus \{\boldsymbol{0}\}$ , if  $\boldsymbol{u}^{\mathsf{T}} (\boldsymbol{A} \widetilde{\boldsymbol{\Gamma}} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}) \boldsymbol{u} \geq \boldsymbol{u}^{\mathsf{T}} (\boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}) \boldsymbol{u}$ , then  $\boldsymbol{u}^{\mathsf{T}} (\boldsymbol{A} \widetilde{\boldsymbol{\Gamma}} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}) \boldsymbol{u}$ 

 $\mathbf{R}$ ) $\mathbf{u} > 0$ . Otherwise,

$$\boldsymbol{u}^{\mathsf{T}}\left(\boldsymbol{A}\widetilde{\boldsymbol{\Gamma}}\boldsymbol{A}^{\mathsf{T}}+\boldsymbol{R}\right)\boldsymbol{u}=\boldsymbol{u}^{\mathsf{T}}\left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}}+\boldsymbol{R}\right)\boldsymbol{u}-\left|\boldsymbol{u}^{\mathsf{T}}\boldsymbol{A}\left(\boldsymbol{\Gamma}-\widetilde{\boldsymbol{\Gamma}}\right)\boldsymbol{A}^{\mathsf{T}}\boldsymbol{u}\right|$$
(C.27)

$$\geq \left(\beta - \|\boldsymbol{\Gamma} - \widetilde{\boldsymbol{\Gamma}}\|_2 \|\boldsymbol{A}\|_2^2\right) \|\boldsymbol{u}\|^2 \tag{C.28}$$

$$\geq \left(\beta - \epsilon \|\boldsymbol{A}\|_{2}^{2}\right) \|\boldsymbol{u}\|^{2}, \tag{C.29}$$

where  $\|\cdot\|_2$  denotes the induced  $l_2$  norm. We can always find an  $\epsilon > 0$  such that  $(\beta - \epsilon \|\mathbf{A}\|_2^2) > 0$ . Therefore,  $\mathbf{u}^{\mathsf{T}}(\mathbf{A}\widetilde{\Gamma}\mathbf{A}^{\mathsf{T}} + \mathbf{R})\mathbf{u} > 0 \forall \mathbf{u} \in \mathbb{R}^m \setminus \{\mathbf{0}\}$ , and thus  $\mathbb{O}$  is an open set.

To prove the second part of the lemma, suppose the sequence  $\gamma_k \in \mathbb{O}$  converges to  $\gamma$ . Then, for any vector  $\boldsymbol{u} \in \mathbb{R}^m \setminus \{\boldsymbol{0}\}, \, \boldsymbol{u}^{\mathsf{T}} \left(\boldsymbol{A}\boldsymbol{\Gamma}_k \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right) \boldsymbol{u}$  converges to  $\boldsymbol{u}^{\mathsf{T}} \left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right) \boldsymbol{u}$ by the continuity of the function. Therefore,

$$\boldsymbol{u}^{\mathsf{T}} \left( \boldsymbol{A} \boldsymbol{\Gamma}_{k} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right) \boldsymbol{u} > 0 \implies \boldsymbol{u}^{\mathsf{T}} \left( \boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R} \right) \boldsymbol{u} \ge 0.$$
 (C.30)

Thus  $A\Gamma A^{\mathsf{T}} + R \succeq 0$ . Conversely, if there is exists a  $\gamma \in \mathbb{R}^m$  such that  $A\Gamma A^{\mathsf{T}} + R \succeq 0$ , the sequence  $\gamma_k = \gamma + (1/k)\mathbf{1}$  converges to  $\gamma$ . We also note that  $A\Gamma_k A^{\mathsf{T}} + R = A\Gamma A^{\mathsf{T}} + \mathbf{R} + (1/k)AA^{\mathsf{T}} \succ \mathbf{0}$  since A has full row rank. Thus, there exists a sequence  $\{\gamma_k\} \in \mathbb{O}$  that converges to  $\gamma$ . Hence, the proof is complete.

#### Proof of Theorem 5.1

We prove the convergence using [183, Theorem 2] which states that: Suppose  $f(\cdot)$  is a continuous vector field defined on an open set  $\mathbb{O} \subset \mathbb{R}^N$  such that  $\mathbb{G} = \{\gamma \in \mathbb{O} : f(\gamma) = 0\}$  is a compact subset of  $\mathbb{O}$ . Then the distance of the sequence  $\gamma_k$  given by (5.2) to the set  $\mathbb{G}$  converges to 0 *a.s.* provided:

- (i) There exists a  $\mathcal{C}^1$  function  $V : \mathbb{O} \to \mathbb{R}_+$  such that
  - (a)  $V(\boldsymbol{\gamma}) \to \infty$  if  $\boldsymbol{\gamma} \to$  the boundary of  $\mathbb{O}$  or  $\|\boldsymbol{\gamma}\| \to \infty$
  - (b)  $\langle \nabla_{\boldsymbol{\gamma}} V(\boldsymbol{\gamma}), \boldsymbol{f}(\boldsymbol{\gamma}) \rangle < 0, \, \forall \boldsymbol{\gamma} \notin \mathbb{G}.$
- (ii)  $\boldsymbol{\gamma}_k$  belongs to a compact set of  $\mathbb{O}$ .
- (iii)  $\lim_{k\to\infty} \sum_{t=1}^k \frac{1}{t} \boldsymbol{e}_t$  exists and is finite.

To check whether assumptions (i)-(iii) hold in our case, we define the set  $\mathbb{O} = \{\gamma : \text{Rank} \{A\Gamma A^{\mathsf{T}} + R\} = m\}$  which is an open set by Lemma C.3. Note that f is a continuous function of  $\gamma$ . Also, the inverse image of the compact set  $\{0\}$  by  $f(\gamma)$  is compact, and hence,  $\mathbb{G}$  is a compact subset of  $\mathbb{O}$ .

We define the  $C^1$  function in (i) as follows:

$$V(\boldsymbol{\gamma}) = \mathsf{Tr}\left\{\left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right)^{-1}\left(\boldsymbol{A}\boldsymbol{\Gamma}_{\mathsf{opt}}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right)\right\} - \log\left|\left(\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right)^{-1}\left(\boldsymbol{A}\boldsymbol{\Gamma}_{\mathsf{opt}}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right)\right|. \quad (C.31)$$

Note that  $V(\boldsymbol{\gamma}) - m$  gives the KL divergence between the distributions  $\mathcal{N}(\mathbf{0}, \boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R})$ and  $\mathcal{N}(\mathbf{0}, \boldsymbol{A}\boldsymbol{\Gamma}_{\mathsf{opt}}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R})$ . Therefore,  $V(\boldsymbol{\gamma}) \geq m > 0$ . By Lemma C.3, if  $\boldsymbol{\gamma}$  is on the boundary of  $\mathbb{O}$ , at least one eigenvalue of  $\boldsymbol{A}\boldsymbol{\Gamma}\boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}$  is zero. Hence, (ia) is satisfied. The gradient of  $V(\boldsymbol{\gamma})$  is given by

$$\nabla_{\gamma} V(\gamma) = \operatorname{Diag} \left\{ \boldsymbol{A}^{\mathsf{T}} \nabla_{\{\boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\}} V\left(\boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right) \boldsymbol{A} \right\}$$
$$= \operatorname{Diag} \left\{ \boldsymbol{A}^{\mathsf{T}} \left(\boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right)^{-1} \boldsymbol{A} \left(\boldsymbol{\Gamma} - \boldsymbol{\Gamma}_{\mathsf{opt}}\right) \boldsymbol{A}^{\mathsf{T}} \left(\boldsymbol{A} \boldsymbol{\Gamma} \boldsymbol{A}^{\mathsf{T}} + \boldsymbol{R}\right)^{-1} \boldsymbol{A} \right\}. \quad (C.32)$$

Substituting this relation in (5.7) gives  $\boldsymbol{f}(\boldsymbol{\gamma}) = -\Gamma^2 \nabla_{\boldsymbol{\gamma}} V(\boldsymbol{\gamma})$ . Therefore, for  $\boldsymbol{\gamma} \in \mathbb{O} \setminus \mathbb{G}$ ,

we have  $\langle \nabla_{\gamma} V(\gamma), \boldsymbol{f}(\gamma) \rangle < 0$ . Thus, (ib) is satisfied.

Assumptions (ii) and (iii) holds because of Proposition 5.1 and Lemma C.1, respectively. Hence,  $\gamma_k$  converges to the set  $\mathbb{G}$ . Further, Proposition 5.1 shows that  $\gamma_k \geq 0$ , and hence, we get that  $\gamma_k$  converges to the set  $\{\mathbf{0}\} \cup \{\gamma \in \mathbb{R}^N_+ : \mathbf{A} (\Gamma - \Gamma_{opt}) \mathbf{A}^{\mathsf{T}} = \mathbf{0}\}$ . Finally, if Rank  $\{\mathbf{A} \odot \mathbf{A}\} = N$ , then  $\{\gamma \in \mathbb{R}^N_+ : \mathbf{A} (\Gamma - \Gamma_{opt}) \mathbf{A}^{\mathsf{T}} = \mathbf{0}\} = \{\gamma_{opt}\}$ . Thus, the proof is complete.

### C.3 Proof of Proposition 5.2

*Proof.* From (5.14), we get,

$$\boldsymbol{P}_{k} = \boldsymbol{P}_{k-1} - \boldsymbol{P}_{k-1}\boldsymbol{A}_{k}^{\mathsf{T}} \left(\boldsymbol{A}_{k}\boldsymbol{P}_{k-1}\boldsymbol{A}_{k}^{\mathsf{T}} + \boldsymbol{R}_{k}\right)^{-1} \boldsymbol{A}_{k}\boldsymbol{P}_{k-1}$$
$$= \left(\boldsymbol{P}_{k-1}^{-1} + \boldsymbol{A}_{k}^{\mathsf{T}}\boldsymbol{R}_{k}^{-1}\boldsymbol{A}_{k}\right)^{-1}$$
(C.33)

$$= \left( \boldsymbol{P}_0^{-1} + \sum_{t=1}^k \boldsymbol{A}_t^{\mathsf{T}} \boldsymbol{R}_t^{-1} \boldsymbol{A}_t \right)^{-1}.$$
(C.34)

Let  $\boldsymbol{Q} \triangleq \lim_{k \to \infty} \frac{1}{k} \sum_{t=1}^{k} \boldsymbol{A}_{t}^{\mathsf{T}} \boldsymbol{R}^{-1} \boldsymbol{A}_{t}$ . From assumptions A1 and A2, we have

$$\boldsymbol{Q} \triangleq \lim_{k \to \infty} \frac{1}{k} \sum_{t=1}^{k} \boldsymbol{A}_{t}^{\mathsf{T}} \boldsymbol{R}^{-1} \boldsymbol{A}_{t}$$
(C.35)

Thus, we get

$$\boldsymbol{Q} = \boldsymbol{S} + \mathbb{E}\left\{\boldsymbol{A}_{t}^{\mathsf{T}}\right\} \boldsymbol{R}^{-1} \mathbb{E}\left\{\boldsymbol{A}_{t}\right\}, \qquad (C.36)$$

where  $\boldsymbol{S} \triangleq \text{Diag} \{ \text{Tr} \{ \boldsymbol{R}^{-1} \text{cov}(\boldsymbol{A}_t[i]) \}, i = 1, 2, ..., N \}$ . Further, since  $\boldsymbol{A}_t$  is random,  $\boldsymbol{S}$  is a positive definite matrix and hence,  $\boldsymbol{Q}$  is a positive definite matrix. Let  $\boldsymbol{Q} = \boldsymbol{U} \boldsymbol{\Lambda} \boldsymbol{U}^{\mathsf{T}}$  be the eigen decomposition such that  $\boldsymbol{\Lambda}$  is a diagonal matrix containing the positive eigen values of Q. Then,

$$\lim_{k \to \infty} \boldsymbol{P}_k = \lim_{k \to \infty} \boldsymbol{U} (\boldsymbol{U} \boldsymbol{P}_0^{-1} \boldsymbol{U}^{\mathsf{T}} + k \boldsymbol{\Lambda})^{-1} \boldsymbol{U}^{\mathsf{T}}.$$
 (C.37)

Let  $\lambda_{\min} > 0$  and  $\gamma_{\min} > 0$  be the smallest diagonal entries of  $\Lambda$  and  $P_0^{-1}$ , respectively. Then, the largest eigenvalue of  $(\boldsymbol{U}\boldsymbol{P}_0^{-1}\boldsymbol{U}^{\mathsf{T}} + k\Lambda)^{-1}$ , denoted by  $\tilde{\lambda}_k$ , can be bounded using Weyl's inequality as follows:

$$0 \le \tilde{\lambda}_k \le \frac{1}{\gamma_{\min} + k\lambda_{\min}}.$$
(C.38)

Hence, we get

$$\lim_{k \to \infty} \boldsymbol{P}_k = \boldsymbol{U} \lim_{k \to \infty} (\boldsymbol{U} \boldsymbol{P}_0^{-1} \boldsymbol{U}^{\mathsf{T}} + k \boldsymbol{\Lambda})^{-1} \boldsymbol{U}^{\mathsf{T}} = \boldsymbol{0}.$$
(C.39)

From (5.12), we get  $\lim_{k\to\infty} J_k = 0$ , and from (5.13), we get  $\hat{x}_k = \hat{x}_{k-1}$  as  $k \to \infty$ . Hence, the algorithm converges.

### C.4 Proof of Theorem 5.3

*Proof.* Let the true solution be  $\boldsymbol{x}$ , and  $\hat{\boldsymbol{x}}_{\infty} \triangleq \lim_{k \to \infty} \hat{\boldsymbol{x}}_k$  as guaranteed by Proposition 5.2. From (5.13),

$$\widehat{\boldsymbol{x}}_{k} = (\boldsymbol{I} - \boldsymbol{J}_{k} \boldsymbol{A}_{k}) \,\widehat{\boldsymbol{x}}_{k-1} + \boldsymbol{J}_{k} \boldsymbol{y}_{k} \tag{C.40}$$

$$= \boldsymbol{P}_{k} \boldsymbol{P}_{k-1}^{-1} \widehat{\boldsymbol{x}}_{k-1} + \boldsymbol{J}_{k} \boldsymbol{y}_{k}$$
(C.41)

$$= \boldsymbol{P}_{k} \boldsymbol{P}_{0}^{-1} \widehat{\boldsymbol{x}}_{0} + \boldsymbol{P}_{k} \sum_{t=1}^{k} \boldsymbol{P}_{t}^{-1} \boldsymbol{J}_{t} \boldsymbol{y}_{t}.$$
(C.42)

Using (C.39), we get

$$\boldsymbol{x}_{\infty} = \lim_{k \to \infty} \boldsymbol{P}_k \sum_{t=1}^{k} \boldsymbol{P}_t^{-1} \boldsymbol{J}_t \boldsymbol{y}_t.$$
(C.43)

Since  $\boldsymbol{y}_t = \boldsymbol{A}_t \boldsymbol{x} + \boldsymbol{w}_t$ , and from (5.14)  $\boldsymbol{J}_t \boldsymbol{A}_t = \boldsymbol{I} - \boldsymbol{P}_t \boldsymbol{P}_{t-1}^{-1}$ ,

$$\widehat{\boldsymbol{x}}_{\infty} = \lim_{k \to \infty} \boldsymbol{P}_{k} \sum_{t=1}^{k} \left[ \boldsymbol{P}_{t}^{-1} (\boldsymbol{I} - \boldsymbol{P}_{t} \boldsymbol{P}_{t-1}^{-1}) \boldsymbol{x} + \boldsymbol{P}_{t}^{-1} \boldsymbol{J}_{t} \boldsymbol{w}_{t} \right]$$
$$= \lim_{k \to \infty} \left[ \left( \boldsymbol{I} - \boldsymbol{P}_{k} \boldsymbol{P}_{0}^{-1} \right) \boldsymbol{x} + \boldsymbol{P}_{k} \sum_{t=1}^{k} \boldsymbol{P}_{t}^{-1} \boldsymbol{J}_{t} \boldsymbol{w}_{t} \right]$$
(C.44)

$$= \boldsymbol{x} + \lim_{k \to \infty} \boldsymbol{P}_k \sum_{t=1}^{\kappa} \boldsymbol{P}_t^{-1} \boldsymbol{J}_t \boldsymbol{w}_t.$$
(C.45)

We now consider the term  $P_t^{-1}J_t$  to simplify the second term in the above expression, and using (C.33) and (5.12) we get

$$\boldsymbol{P}_{t}^{-1}\boldsymbol{J}_{t} = \left(\boldsymbol{P}_{t-1}^{-1} + \boldsymbol{A}_{t}^{\mathsf{T}}\boldsymbol{R}^{-1}\boldsymbol{A}_{t}\right)\boldsymbol{P}_{t-1}\boldsymbol{A}_{t}^{\mathsf{T}}\left(\boldsymbol{A}_{t}\boldsymbol{P}_{t-1}\boldsymbol{A}_{t}^{\mathsf{T}} + \boldsymbol{R}\right)^{-1}$$
(C.46)

$$= \boldsymbol{A}_{t}^{\mathsf{T}} (\boldsymbol{I} + \boldsymbol{R}^{-1} \boldsymbol{A}_{t} \boldsymbol{P}_{t-1} \boldsymbol{A}_{t}^{\mathsf{T}}) (\boldsymbol{A}_{t} \boldsymbol{P}_{t-1} \boldsymbol{A}_{t}^{\mathsf{T}} + \boldsymbol{R})^{-1} = \boldsymbol{A}_{t}^{\mathsf{T}} \boldsymbol{R}^{-1}. \quad (C.47)$$

Thus,

$$\widehat{\boldsymbol{x}}_{\infty} = \boldsymbol{x} + \lim_{k \to \infty} \left( k \boldsymbol{P}_k \right) \left( \frac{1}{k} \sum_{t=1}^k \boldsymbol{A}_t^{\mathsf{T}} \boldsymbol{R}^{-1} \boldsymbol{w}_t \right).$$
(C.48)

We note that

$$\lim_{k \to \infty} \frac{1}{k} \sum_{t=1}^{k} \boldsymbol{A}_{t}^{\mathsf{T}} \boldsymbol{R}^{-1} \boldsymbol{w}_{t} = \mathbb{E} \left\{ \boldsymbol{A}_{t}^{\mathsf{T}} \boldsymbol{R}^{-1} \boldsymbol{w}_{t} \right\}$$
(C.49)

$$= \mathbb{E} \left\{ \boldsymbol{A}_{t} \right\}^{\mathsf{T}} \boldsymbol{R}^{-1} \mathbb{E} \left\{ \boldsymbol{w}_{t} \right\} = \boldsymbol{0}.$$
 (C.50)

Here, we use the fact that  $A_t$  and  $w_t$  are independent and the mean of  $w_t$  is zero. Further, from (C.37), we get

$$\lim_{k \to \infty} k \boldsymbol{P}_k = \lim_{k \to \infty} k \boldsymbol{U} (\boldsymbol{U} \boldsymbol{P}_0^{-1} \boldsymbol{U}^{\mathsf{T}} + k \boldsymbol{\Lambda})^{-1} \boldsymbol{U}^{\mathsf{T}}$$
(C.51)

$$= \lim_{k \to \infty} (k^{-1} \boldsymbol{P}_0^{-1} + \boldsymbol{U} \boldsymbol{\Lambda} \boldsymbol{U}^{\mathsf{T}})^{-1}$$
(C.52)

$$= \boldsymbol{U}\boldsymbol{\Lambda}^{-1}\boldsymbol{U}^{\mathsf{T}}.$$
 (C.53)

Substituting (C.50) and (C.53) in (C.48), we get  $\widehat{x}_{\infty} = x$ , and the proof is complete.  $\Box$ 

# Appendix D

# Appendix to Chapter 6

### D.1 Proof of Proposition 6.1

*Proof.* For the AM procedure, since we optimize one column of  $\boldsymbol{A}$  at a time, it is easy to see that

$$g\left(\mathbf{A}^{(r,u-1)}\right) \ge g\left(\mathbf{A}^{(r,u)}\right).$$
 (D.1)

The above relation holds even if we skip the update of a column when  $\left\| \boldsymbol{v}_{i}^{(r,u)} \right\| = 0$ , in which case the value of the cost function remains unchanged. Similarly, from (6.15), the sequence  $\left\{ g\left(\boldsymbol{A}^{(r,u)}\right) \right\}_{u\in\mathbb{N}}$  generated by the ALS algorithm is also nonincreasing. Thus, we conclude that in both cases, the sequence  $\left\{ g\left(\boldsymbol{A}^{(r,u)}\right) \right\}_{u\in\mathbb{N}}$  is a nonincreasing sequence bounded by  $g\left(\boldsymbol{A}^{(r,0)}\right)$  from above. From (6.10), we get

$$g(\boldsymbol{A}) = \frac{1}{2} \operatorname{Tr} \left\{ \left( \boldsymbol{Y} \boldsymbol{M}^{\mathsf{T}} - \boldsymbol{A} \right)^{\mathsf{T}} \left( \boldsymbol{Y} \boldsymbol{M}^{\mathsf{T}} - \boldsymbol{A} \right) + \boldsymbol{A} \boldsymbol{\Sigma} \boldsymbol{A}^{\mathsf{T}} \right\} - \frac{1}{2} \operatorname{Tr} \left\{ \boldsymbol{Y}^{\mathsf{T}} \boldsymbol{M} \boldsymbol{M}^{\mathsf{T}} \boldsymbol{Y} + \boldsymbol{\Sigma} \right\} - N/2$$
(D.2)

$$\geq -\frac{1}{2} \operatorname{Tr} \left\{ \boldsymbol{Y}^{\mathsf{T}} \boldsymbol{M} \boldsymbol{M}^{\mathsf{T}} \boldsymbol{Y} + \boldsymbol{\Sigma} \right\} - N/2.$$
 (D.3)

Therefore, the nonincreasing sequence  $\left\{g\left(\mathbf{A}^{(r,u)}\right)\right\}_{u\in\mathbb{N}}$  is bounded from below, and hence it converges.

## D.2 Proof of Theorem 6.1

We first present a Lemma which characterizes a Nash equilibrium point of the objective function in (6.9).

**Lemma D.1.** A matrix  $\mathbf{A}$  with unit norm columns is a Nash equilibrium point of the objective function in (6.9) if and only if  $\mathbf{A}$  satisfies the relation:

$$\boldsymbol{A}\boldsymbol{L} = \boldsymbol{Y}\boldsymbol{M}^{\mathsf{T}} - \boldsymbol{A}\left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right) \tag{D.4}$$

for some diagonal psd matrix L.

*Proof.* If (D.4) is satisfied, from (6.13), it is a solution to the reduced optimization problem

$$\boldsymbol{A}_{i} = \underset{\boldsymbol{a}:\|\boldsymbol{a}\|=1}{\operatorname{arg\,min}} g\left( \left[ \boldsymbol{A}_{1}, \dots, \boldsymbol{A}_{i-1}, \boldsymbol{a}, \boldsymbol{A}_{i+1}, \dots, \boldsymbol{A}_{N} \right] \right). \tag{D.5}$$

Thus,  $\boldsymbol{A}$  is a Nash equilibrium point.

Next, suppose that  $\boldsymbol{A}$  is a Nash equilibrium point, and let  $\boldsymbol{Z} = \boldsymbol{Y}\boldsymbol{M}^{\mathsf{T}} - \boldsymbol{A} (\boldsymbol{\Sigma} - \mathcal{D} \{\boldsymbol{\Sigma}\}).$ We decompose every column of  $\boldsymbol{Z}$  along the corresponding column of  $\boldsymbol{A}$  to get  $\boldsymbol{Z} = \boldsymbol{A}\boldsymbol{L} + \boldsymbol{A}^{\perp}$ , where  $\boldsymbol{L} \in \mathbb{R}^{N \times N}$  is a diagonal matrix and  $\boldsymbol{A}^{\perp} \in \mathbb{R}^{m \times N}$  is such that  $\boldsymbol{A}_{i}^{\mathsf{T}} \boldsymbol{A}_{i}^{\perp} = 0$ ,  $\forall i$ . Let i be an index such that  $\boldsymbol{A}_{i}^{\perp} \neq \mathbf{0}$ . We define the following matrix:

$$\boldsymbol{B} = \begin{bmatrix} \boldsymbol{A}_1 & \dots & \boldsymbol{A}_{i-1} & \boldsymbol{Z}_i / \| \boldsymbol{Z}_i \| & \boldsymbol{A}_{i+1} & \dots & \boldsymbol{A}_N \end{bmatrix} \in \mathbb{O}.$$
(D.6)

From Definition 6.1, and using the expression for  $g(\cdot)$  in (6.10), we have

$$0 \leq g(\boldsymbol{B}) - g(\boldsymbol{A})$$
(D.7)  
$$= - \left(\boldsymbol{Y}^{\mathsf{T}}\boldsymbol{M}\right)_{i}^{\mathsf{T}} \left(\boldsymbol{Z}_{i} / \|\boldsymbol{Z}_{i}\| - \boldsymbol{A}_{i}\right)$$
$$+ \left(\boldsymbol{B}^{\mathsf{T}}\boldsymbol{B} - \boldsymbol{A}^{\mathsf{T}}\boldsymbol{A}\right)_{i}^{\mathsf{T}} \left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right)_{i}$$
(D.8)

$$= -\boldsymbol{Z}_{i}^{\mathsf{T}}\left(\boldsymbol{Z}_{i}/\|\boldsymbol{Z}_{i}\|-\boldsymbol{A}_{i}\right) = \boldsymbol{L}_{ii} - \|\boldsymbol{Z}_{i}\|.$$
(D.9)

However, we have  $\|\boldsymbol{Z}_i\|^2 = |\boldsymbol{L}_{ii}|^2 + \|\boldsymbol{A}_i^{\perp}\|^2$ , which gives  $\|\boldsymbol{Z}_i\| \ge |\boldsymbol{L}_{ii}|$ . Therefore, from (D.7), we get  $\boldsymbol{L}_{ii} = \|\boldsymbol{Z}_i\| \ge 0$ , and  $\|\boldsymbol{A}_i^{\perp}\| = 0, \forall i$ . Hence,  $\boldsymbol{Z} = \boldsymbol{A}\boldsymbol{L}$  for some diagonal psd matrix  $\boldsymbol{L}$ , and the proof is complete.

### Proof of Theorem 6.1

*Proof.* It is easy to see that any matrix that satisfies (D.4) is a fixed point of G. Now, from Lemma D.1, to complete the proof, it suffices to show that any fixed point of the AM procedure satisfies the relation (D.4) for some diagonal matrix  $\boldsymbol{L}$  with nonnegative entries. From (6.13), we get

$$G(\boldsymbol{A})_i \|\boldsymbol{v}_i\| = \boldsymbol{v}_i, \tag{D.10}$$

where

$$\boldsymbol{v}_{i} = \sum_{k=1}^{K} \boldsymbol{\mu}_{k}[i] \boldsymbol{y}_{k} - \sum_{j=1}^{i-1} \boldsymbol{\Sigma}[i, j] \boldsymbol{G}(\boldsymbol{A})_{j} - \sum_{j=i+1}^{N} \boldsymbol{\Sigma}[i, j] \boldsymbol{A}_{j}$$
$$= \left(\boldsymbol{Y} \boldsymbol{M}^{\mathsf{T}}\right)_{i} - \boldsymbol{G}(\boldsymbol{A}) \left(\hat{\boldsymbol{\Sigma}}^{\mathsf{T}}\right)_{i} - \boldsymbol{A}\hat{\boldsymbol{\Sigma}}_{i}, \qquad (D.11)$$

where  $\hat{\Sigma}$  is a lower triangular matrix with zero diagonal entries and  $\hat{\Sigma} + \hat{\Sigma}^{\mathsf{T}} = \Sigma - \mathcal{D} \{ \Sigma \}$ . When  $\boldsymbol{A}$  is a fixed point of  $\boldsymbol{G}$ , we get

$$\boldsymbol{v}_{i} = \left(\boldsymbol{Y}\boldsymbol{M}^{\mathsf{T}}\right)_{i} - \boldsymbol{A}\left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right)_{i}.$$
 (D.12)

Now, from (D.10) and (D.12), it can be seen that  $\boldsymbol{A}$  satisfies (D.4) with  $\boldsymbol{L}_{ii} = \|\boldsymbol{v}_i\| \ge 0$ , which concludes the proof.

# D.3 Proof of Theorem 6.2

The proof of the theorem rests on the following lemmas.

**Lemma D.2.** Let  $\{A^{(r,u)}\}_{u\in\mathbb{N}}$  be a sequence generated by the ALS procedure. Then, there exists  $C_1 > 0$  such that

$$\tilde{g}\left(\boldsymbol{A}^{(r,u-1)}\right) - \tilde{g}\left(\boldsymbol{A}^{(r,u)}\right) \ge C_1 \left\|\boldsymbol{A}^{(r,u-1)} - \boldsymbol{A}^{(r,u)}\right\|^2.$$
(D.13)

*Proof.* We note from (6.19) that

$$\boldsymbol{A}_{i}^{(r,u)} = \frac{\boldsymbol{A}_{i}^{(r,u-1)} + \beta^{p} \alpha \boldsymbol{Z}_{i}^{(r,u-1)}}{\left\| \boldsymbol{A}_{i}^{(r,u-1)} + \beta^{p} \alpha \boldsymbol{Z}_{i}^{(r,u-1)} \right\|}.$$
 (D.14)

Also, from (6.18), we know that

$$\boldsymbol{A}_{i}^{(r,u-1)\mathsf{T}}\boldsymbol{Z}_{i}^{(r,u-1)} = 0.$$
 (D.15)

Therefore, we get

$$\frac{1}{2} \left\| \boldsymbol{A}^{(r,u-1)} - \boldsymbol{A}^{(r,u)} \right\|^2 = \sum_{i=1}^{N} \frac{1}{2} \left\| \boldsymbol{A}_i^{(r,u-1)} - \boldsymbol{A}_i^{(r,u)} \right\|^2$$
(D.16)

$$=\sum_{i=1}^{N} \left(1 - \boldsymbol{A}_{i}^{(r,u-1)\mathsf{T}} \boldsymbol{A}_{i}^{(r,u)}\right) \tag{D.17}$$

$$=\sum_{i=1}^{N} \left(1 - \frac{1}{\sqrt{1 + \left\|\beta^{p} \alpha \boldsymbol{Z}_{i}^{(r,u-1)}\right\|^{2}}}\right)$$
(D.18)

$$\leq \sum_{i=1}^{N} \left\| \beta^{p} \alpha \boldsymbol{Z}_{i}^{(r,u-1)} \right\|^{2}, \qquad (D.19)$$

$$\leq \frac{1}{c} \left[ g \left( \boldsymbol{A}^{(r,u-1)} \right) - g \left( \boldsymbol{A}^{(r,u)} \right) \right]$$
(D.20)

where (D.17) is because  $A_i^{(r,u-1)}$  and  $A_i^{(r,u)}$  are unit norm vectors, and (D.18) is a direct consequence of (D.14) and (D.15). (D.19) is due to the fact that  $x^2 + 1/\sqrt{1+x^2} - 1 \ge 0$  for all  $x \in \mathbb{R}$ , and (D.20) follows immediately from (6.15). Thus, the proof is complete.  $\Box$ 

**Lemma D.3** (Subgradient of  $\delta_{\text{norm}}$ ). For any matrix  $A \in \mathbb{O} \subset \mathbb{R}^{m \times N}$ .

$$\partial \delta_{norm} \left( \boldsymbol{A} \right) = \left\{ \boldsymbol{A} \tilde{\boldsymbol{L}}, \, \tilde{\boldsymbol{L}} \in \mathbb{R}^{N \times N} : \substack{\boldsymbol{L}_{ii} \ge 0, \forall i \\ \tilde{\boldsymbol{L}}_{ij} = 0, i \neq j} \right\}.$$
(D.21)

Proof. Let  $\mathbf{Z} \in \partial \delta_{\text{norm}}(\mathbf{A})$ . From the definition of the subgradient, we get  $\delta_{\text{norm}}(\mathbf{A}) + \text{Tr} \{ \mathbf{Z}^{\mathsf{T}}(\mathbf{B} - \mathbf{A}) \} \leq \delta_{\text{norm}}(\mathbf{B}), \forall \mathbf{B} \in \mathbb{R}^{m \times N}$ . This relation is trivially satisfied for all  $\mathbf{Z}$  and for any  $\mathbf{B} \notin \mathbb{O}$ . However, when  $\mathbf{B} \in \mathbb{O}, \mathbf{Z}$  should satisfy

$$\operatorname{Tr}\left\{\boldsymbol{Z}^{\mathsf{T}}\boldsymbol{B}\right\} \leq \operatorname{Tr}\left\{\boldsymbol{Z}^{\mathsf{T}}\boldsymbol{A}\right\},$$
 (D.22)

since  $\delta_{\text{norm}}(\boldsymbol{A}) = \delta_{\text{norm}}(\boldsymbol{B}).$ 

To prove the result, we consider three different cases that cover all possible values for Z.

1. We express the columns of the matrix  $\mathbf{Z}$  as  $\mathbf{Z}_i = \tilde{\mathbf{L}}_{ii}\mathbf{A}_i + \mathbf{A}_i^{\perp}$ , where  $\tilde{\mathbf{L}}_{ii} \in \mathbb{R}$  and  $\mathbf{A}_i^{\perp} \in \mathbb{R}^m$  is such that  $\mathbf{A}_i^{\mathsf{T}}\mathbf{A}_i^{\perp} = 0$ ,  $\forall i$ . Suppose  $\mathbf{A}_i^{\perp} \neq \mathbf{0}$  for at least one value of i. Also, let  $\mathbf{B} \in \mathbb{R}^{m \times N} \in \mathbb{O}$  be defined as

$$\boldsymbol{B}_{i} = \begin{cases} \boldsymbol{e}, & \text{for } \|\boldsymbol{Z}_{i}\| = 0, \\ \boldsymbol{Z}_{i}/\|\boldsymbol{Z}_{i}\|, & \text{for } \|\boldsymbol{Z}_{i}\| \neq 0, \end{cases}$$
(D.23)

where e is any unit norm vector. Then,

$$\operatorname{Tr}\left\{\boldsymbol{Z}^{\mathsf{T}}\boldsymbol{A}\right\} = \sum_{i=1}^{N} \tilde{\boldsymbol{L}}_{ii} < \sum_{i=1}^{N} \|\boldsymbol{Z}_{i}\| = \operatorname{Tr}\left\{\boldsymbol{Z}^{\mathsf{T}}\boldsymbol{B}\right\}.$$
(D.24)

Therefore, there exists a matrix  $B \in \mathbb{O}$  such that (D.22) is not satisfied. Thus, we get

$$\partial \delta_{\text{norm}} \left( \boldsymbol{A} \right) \subseteq \left\{ \boldsymbol{A} \tilde{\boldsymbol{L}}, \tilde{\boldsymbol{L}} \in \mathbb{R}^{N \times N} : \tilde{\boldsymbol{L}}_{ij} = 0, \text{ if } i \neq j \right\}.$$
 (D.25)

2. Let  $\mathbf{Z} = \mathbf{A}\tilde{\mathbf{L}}$  for some diagonal matrix such that at least one of the diagonal entries of  $\tilde{\mathbf{L}}$  is negative. Let  $\mathbf{B} \in \mathbb{R}^{m \times N} \in \mathbb{O}$  be defined such that  $\mathbf{B}_i = \operatorname{sign} \left\{ \tilde{\mathbf{L}}_{ii} \right\} \mathbf{A}_i$ , where the function  $\operatorname{sign}\{\cdot\}$  takes values 1 and -1 for nonnegative and negative arguments, respectively. Then,

$$\operatorname{Tr}\left\{\boldsymbol{Z}^{\mathsf{T}}\boldsymbol{A}\right\} = \sum_{i=1}^{N} \tilde{\boldsymbol{L}}_{ii} < \sum_{i=1}^{N} \left|\tilde{\boldsymbol{L}}_{ii}\right| \leq \operatorname{Tr}\left\{\boldsymbol{Z}^{\mathsf{T}}\boldsymbol{B}\right\}, \qquad (D.26)$$

Therefore, (D.22) does not hold for  $\boldsymbol{B} \in \mathbb{O}$ , and from (D.25) we get

$$\partial \delta_{\text{norm}} \left( \boldsymbol{A} \right) \subseteq \left\{ \boldsymbol{A} \tilde{\boldsymbol{L}}, \tilde{\boldsymbol{L}} \in \mathbb{R}^{N \times N} : \underset{\tilde{\boldsymbol{L}}_{ij}=0, \text{ if } i \neq j}{\overset{L}{\boldsymbol{L}}} \right\}.$$
(D.27)

3. Let  $\mathbf{Z} = \mathbf{A}\tilde{\mathbf{L}}$ , for some diagonal positive semidefinite (psd) matrix  $\tilde{\mathbf{L}}$ . Here, for any

matrix  $\boldsymbol{B} \in \mathbb{O}$ ,

$$\operatorname{Tr}\left\{\boldsymbol{Z}^{\mathsf{T}}\boldsymbol{B}\right\} = \operatorname{Tr}\left\{\tilde{\boldsymbol{L}}\boldsymbol{A}^{\mathsf{T}}\boldsymbol{B}\right\} = \sum_{i=1}^{N} \tilde{\boldsymbol{L}}_{ii}\boldsymbol{A}_{i}^{\mathsf{T}}\boldsymbol{B}_{i}$$
(D.28)

$$\leq \sum_{i=1}^{N} \tilde{\boldsymbol{L}}_{ii} = \sum_{i=1}^{N} \tilde{\boldsymbol{L}}_{ii} \boldsymbol{A}_{i}^{\mathsf{T}} \boldsymbol{A}_{i}$$
(D.29)

$$= \operatorname{Tr}\left\{\tilde{\boldsymbol{L}}\boldsymbol{A}^{\mathsf{T}}\boldsymbol{A}\right\} = \operatorname{Tr}\left\{\boldsymbol{Z}^{\mathsf{T}}\boldsymbol{A}\right\}.$$
(D.30)

Therefore, from (D.27) we get

$$\partial \delta_{\text{norm}} \left( \boldsymbol{A} \right) = \left\{ \boldsymbol{A} \tilde{\boldsymbol{L}}, \tilde{\boldsymbol{L}} \in \mathbb{R}^{N \times N} : \frac{\boldsymbol{L}_{ii} \ge 0, \forall i}{\tilde{\boldsymbol{L}}_{ij} = 0, \text{ otherwise.}} \right\}.$$
(D.31)

Hence, the proof is complete.

#### D.3.1 Proof of Theorem 6.2

*Proof.* In [156, Theorem 2], the authors provide a Kurdyka-Łojasiewicz property based proof of convergence of a proximal algorithm. By careful examination their proof, it can be shown that a bounded sequence of iterates converges to a stationary point of  $\tilde{g}$  if the following four conditions hold:

(i) The objective function  $\tilde{g}(\mathbf{A})$  satisfies

$$\inf_{\boldsymbol{A} \in \mathbb{R}^{m \times N}} \tilde{g}\left(\boldsymbol{A}\right) > -\infty. \tag{D.32}$$

(ii) There exist constants  $\theta \in [0,1), \, C, \epsilon > 0$  such that

$$\left|\tilde{g}\left(\boldsymbol{A}\right) - \tilde{g}\left(\boldsymbol{A}^{*}\right)\right|^{\theta} \leq C \left\|\boldsymbol{Z}\right\|$$
(D.33)

for any stationary point  $A^*$  of  $\tilde{g}$ , any A such that  $||A - A^*|| \le \epsilon$ , and any Z such that

 $Z \in \partial g(A)$ . The constant  $\theta$  is called the *Lojasiewicz exponent* of the Lojasiewicz gradient inequality.

(iii) There exists  $C_1 > 0$  such that

$$\tilde{g}\left(\boldsymbol{A}^{(r,u-1)}\right) - \tilde{g}\left(\boldsymbol{A}^{(r,u)}\right) \ge C_1 \left\|\boldsymbol{A}^{(r,u-1)} - \boldsymbol{A}^{(r,u)}\right\|^2 \tag{D.34}$$

(iv) There exist  $u_0 > 1$ ,  $C_2 > 0$  and  $\mathbf{Z} \in \partial g\left(\mathbf{A}^{(r,u)}\right)$  such that for all  $u > u_0$ 

$$\|\boldsymbol{Z}\| \le C_2 \left\| \boldsymbol{A}^{(r,u-1)} - \boldsymbol{A}^{(r,u)} \right\|.$$
 (D.35)

Here, the first two conditions are on the cost function, and the last two are on the iterates. In [156, Theorem 2], these conditions are verified to hold for the proximal algorithm. The rest of the proof below is the verification of the four conditions for the ALS procedure.

As discussed in Appendix C (see (D.3)), the cost function g is bounded from below. Therefore,  $\tilde{g}$  is also bounded from below, and hence assumption (i) is satisfied.

Next, we note that  $\delta_{\text{norm}}(\cdot)$  is an indicator function of a semi-algebraic set, and g is a real analytic function. Therefore,  $\tilde{g}$  is a sum of real analytic and semi-algebraic functions. Thus, from [184, Section 2.2], it can be shown that  $\tilde{g}$  satisfies the desired condition (ii). Assumption (iii) follows from Lemma D.2.

Finally, to verify assumption (iv), we first compute the subgradient of  $\tilde{g}$  using Lemma D.3. Hence, the desired condition is true if and only if, for all  $u > u_0$ , it holds that

$$\min_{\tilde{\boldsymbol{Z}} \in \partial \tilde{g}\left(\boldsymbol{A}^{(r,u)}\right)} \left\| \tilde{\boldsymbol{Z}} \right\| \le C_2 \left\| \boldsymbol{A}^{(r,u-1)} - \boldsymbol{A}^{(r,u)} \right\|.$$
(D.36)

Now, from Lemma D.3, we have,

$$\min_{\tilde{\boldsymbol{Z}}\in\partial\tilde{g}(\boldsymbol{A})}\left\|\tilde{\boldsymbol{Z}}\right\|^{2} = \min_{\tilde{\boldsymbol{L}}_{ii}\geq0}\left\|\nabla g(\boldsymbol{A}) + \boldsymbol{A}\tilde{\boldsymbol{L}}\right\|^{2}.$$
 (D.37)

Since the optimization problem is separable in the diagonal entries of  $\tilde{L}$ , we get the optimum value  $\tilde{L}^*$  as

$$\tilde{\boldsymbol{L}}_{ii}^{*} = \begin{cases} -\boldsymbol{A}_{i}^{\mathsf{T}} \nabla g \left(\boldsymbol{A}\right)_{i}, \text{ if } \boldsymbol{A}_{i}^{\mathsf{T}} \nabla g \left(\boldsymbol{A}\right)_{i} \leq 0 \\ 0, \quad \text{otherwise} \end{cases}$$
(D.38)

for  $i = 1, 2, \ldots, N$ . This gives

$$\underset{\tilde{\boldsymbol{Z}}\in\partial\tilde{g}(\boldsymbol{A})}{\operatorname{arg\,min}} \left\| \tilde{\boldsymbol{Z}} \right\| \leq \sqrt{\sum_{i=1}^{N} \max\left\{ \left\| \left( \boldsymbol{I} - \boldsymbol{A}_{i} \boldsymbol{A}_{i}^{\mathsf{T}} \right) \nabla g\left(\boldsymbol{A}\right)_{i} \right\|, \left\| \nabla g\left(\boldsymbol{A}\right)_{i} \right\| \right\}}$$
(D.39)

$$= \left\| \nabla g\left( \boldsymbol{A} \right) \right\|. \tag{D.40}$$

(D.41)

Here, (D.40) follows from the fact that  $\boldsymbol{I} - \boldsymbol{A}_i \boldsymbol{A}_i^{\mathsf{T}}$  is the projection matrix for the subspace orthogonal to the unit norm column  $\boldsymbol{A}_i$ . Therefore,  $\left\| \left( \boldsymbol{I} - \boldsymbol{A}_i \boldsymbol{A}_i^{\mathsf{T}} \right) \nabla g \left( \boldsymbol{A} \right)_i \right\| \leq \| \nabla g \left( \boldsymbol{A} \right)_i \|$ . Thus, we have

$$\min_{\tilde{\boldsymbol{Z}}\in\partial\tilde{g}\left(\boldsymbol{A}^{(r,u)}\right)}\left\|\tilde{\boldsymbol{Z}}\right\| = \left\|\left(\boldsymbol{A}^{(r,u-1)} - \boldsymbol{A}^{(r,u)}\right)\left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right)\right\|$$
(D.42)

$$\leq C_2 \left\| \left( \boldsymbol{A}^{(r,u-1)} - \boldsymbol{A}^{(r,u)} \right) \right\|, \qquad (D.43)$$

where  $C_2$  is the spectral norm of  $\Sigma - \mathcal{D} \{\Sigma\}$ . Also, (D.42) is due to the definition of g in (6.10). Hence, assumption (iv) is satisfied for all u. Therefore, all four conditions are met, and consequently, the convergence is guaranteed.

# D.4 Proof of Proposition 6.2

Proof. From (6.15) and Proposition 5.2,

$$\mathbf{0} = \lim_{u \to \infty} \mathbf{Z}^{(r,u)} = P_{\mathbf{A}^{(r)}} \left( \mathbf{Y} \mathbf{M}^{\mathsf{T}} - \mathbf{A}^{(r)} \mathbf{\Sigma} \right).$$
(D.44)

Thus, (6.18) gives

$$\boldsymbol{Y}\boldsymbol{M}^{\mathsf{T}} - \boldsymbol{A}^{(r)}\left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right) = \boldsymbol{A}^{(r)}\boldsymbol{L},\tag{D.45}$$

for some diagonal matrix  $\boldsymbol{L}$ . Then, we have

$$\nabla g\left(\boldsymbol{A}^{(r)}\right) = -\boldsymbol{A}^{(r)}\boldsymbol{L}.$$
 (D.46)

Let  $\boldsymbol{\Delta} = \boldsymbol{A} - \boldsymbol{A}^{(r)}$ , where  $\boldsymbol{A}$  is any matrix in  $\mathbb{O}$ . Then, for i = 1, 2, ..., N we have

$$1 = \|\boldsymbol{A}_{i}\|^{2} = \left\|\boldsymbol{\Delta}_{i} + \boldsymbol{A}_{i}^{(r)}\right\|^{2} = \|\boldsymbol{\Delta}_{i}\|^{2} + 1 + 2\boldsymbol{\Delta}_{i}^{\mathsf{T}}\boldsymbol{A}_{i}^{(r)}.$$
 (D.47)

Thus, we get  $\frac{1}{2} \|\boldsymbol{\Delta}_i\|^2 = -\boldsymbol{\Delta}_i^{\mathsf{T}} \boldsymbol{A}_i^{(r)}$ , and similarly, expanding  $\|\boldsymbol{A}_i - \boldsymbol{\Delta}_i\|^2$ , we get  $\frac{1}{2} \|\boldsymbol{\Delta}_i\|^2 = \boldsymbol{\Delta}_i^{\mathsf{T}} \boldsymbol{A}_i$ . Therefore,

$$\mathcal{D}\left\{\boldsymbol{\Delta}^{\mathsf{T}}\boldsymbol{A}\right\} = -\mathcal{D}\left\{\boldsymbol{\Delta}^{\mathsf{T}}\boldsymbol{A}^{(r)}\right\} = \frac{1}{2}\mathcal{D}\left\{\boldsymbol{\Delta}^{\mathsf{T}}\boldsymbol{\Delta}\right\}.$$
 (D.48)

Now, using Taylor series expansion around  $A^{(r)}$ , we have

$$g(\boldsymbol{A}) - g\left(\boldsymbol{A}^{(r)}\right) = \operatorname{Tr}\left\{\boldsymbol{\Delta}^{\mathsf{T}} \nabla g\left(\boldsymbol{A}^{(r)}\right) + \frac{1}{2}\boldsymbol{\Delta}^{\mathsf{T}} \boldsymbol{\Delta} \left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right)\right\}$$
(D.49)

$$= \operatorname{Tr}\left\{-\boldsymbol{\Delta}^{\mathsf{T}}\boldsymbol{A}^{(r)}\boldsymbol{L} + \frac{1}{2}\boldsymbol{\Delta}^{\mathsf{T}}\boldsymbol{\Delta}\left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right)\right\}$$
(D.50)

$$=\frac{1}{2}\operatorname{Tr}\left\{\boldsymbol{\Delta}^{\mathsf{T}}\boldsymbol{\Delta}\boldsymbol{L}+\boldsymbol{\Delta}^{\mathsf{T}}\boldsymbol{\Delta}\left(\boldsymbol{\Sigma}-\mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right)\right\}$$
(D.51)

$$= \frac{1}{2} \operatorname{Tr} \left\{ \boldsymbol{\Delta} \left( \boldsymbol{L} + \boldsymbol{\Sigma} - \mathcal{D} \left\{ \boldsymbol{\Sigma} \right\} \right) \boldsymbol{\Delta}^{\mathsf{T}} \right\}, \qquad (D.52)$$

where we use (D.46) and (D.48) to get (D.50) and (D.51) respectively. Note that the Taylor series expansion is not an approximation here, as our cost function is quadratic. The right hand side of (D.52) is non-negative if and only if  $\boldsymbol{L} + \boldsymbol{\Sigma} - \mathcal{D} \{\boldsymbol{\Sigma}\}$  is positive semi-definite, and strictly positive if and only if  $\boldsymbol{L} + \boldsymbol{\Sigma} - \mathcal{D} \{\boldsymbol{\Sigma}\}$  is positive definite. Hence, the proof is complete.

### D.5 Proof of Theorem 6.3

*Proof.* The first part of the result directly follows from Proposition 5.2 and [147, Theorem 4.4.1].

To prove the second part, suppose that  $\mathbf{A}^{(r)}$  is a strict local minimum. Then, for any neighborhood  $\mathcal{U}$  of  $\mathbf{A}^{(r)}$ , there exists  $\epsilon > 0$  such that, in the closed ball  $\mathcal{H}_{\epsilon} \subseteq \mathcal{U}$  around  $\mathbf{A}^{(r)}, g(\mathbf{A}) > g(\mathbf{A}^{(r)})$  for all  $\mathbf{A} \neq \mathbf{A}^{(r)} \in \mathcal{H}_{\epsilon}$ . Here, the closed ball is defined as follows:

$$\mathcal{H}_{\epsilon} = \left\{ \boldsymbol{A} \in \mathbb{O} : \left\| \boldsymbol{A} - \boldsymbol{A}^{(r)} \right\| \le \epsilon \right\}.$$
 (D.53)

Moreover, from Lemma D.2, we get

$$\left\| G(\boldsymbol{A}) - \boldsymbol{A}^{(r)} \right\| \le \left\| G(\boldsymbol{A}) - \boldsymbol{A} \right\| + \left\| \boldsymbol{A} - \boldsymbol{A}^{(r)} \right\|$$
(D.54)

$$\leq C_1 \left[ g \left( G(\boldsymbol{A}) \right) - g \left( \boldsymbol{A} \right) \right] + \left\| \boldsymbol{A} - \boldsymbol{A}^{(r)} \right\|$$
(D.55)

$$\leq C_1 \left[ g\left( \boldsymbol{A} \right) - g\left( \boldsymbol{A}^{(r)} \right) \right] + \left\| \boldsymbol{A} - \boldsymbol{A}^{(r)} \right\|, \qquad (D.56)$$

where the last step is because of Proposition 5.2 which gives  $g(\mathbf{A}) \ge g(G(\mathbf{A})) \ge g(\mathbf{A}^{(r)})$ . From Proposition 6.2, we know that  $\mathbf{A}^{(r)}$  satisfies the relation:

$$\boldsymbol{A}^{(r)}\boldsymbol{L} = \boldsymbol{Y}\boldsymbol{M}^{\mathsf{T}} - \boldsymbol{A}^{(r)}\left(\boldsymbol{\Sigma} - \mathcal{D}\left\{\boldsymbol{\Sigma}\right\}\right), \qquad (\mathrm{D.57})$$

for some diagonal matrix L. Following the same steps as (D.50)-(D.52), we get

$$0 < g(\boldsymbol{A}) - g(\boldsymbol{A}^{(r)}) = \frac{1}{2} \operatorname{Tr} \left\{ \left( \boldsymbol{A} - \boldsymbol{A}^{(r)} \right) \left( \boldsymbol{L} + \boldsymbol{\Sigma} - \mathcal{D} \left\{ \boldsymbol{\Sigma} \right\} \right) \left( \boldsymbol{A} - \boldsymbol{A}^{(r)} \right)^{\mathsf{T}} \right\}$$
(D.58)

$$\leq \frac{\lambda_{\max}}{2} \left\| \boldsymbol{A} - \boldsymbol{A}^{(r)} \right\|^2, \tag{D.59}$$

where  $\lambda_{\max} > 0$  is the largest singular value of the matrix  $(\boldsymbol{L} + \boldsymbol{\Sigma} - \mathcal{D} \{\boldsymbol{\Sigma}\})$ . Thus, from (D.56),

$$\left\| G(\boldsymbol{A}) - \boldsymbol{A}^{(r)} \right\| \leq \frac{C_1 \lambda_{\max}}{2} \left\| \boldsymbol{A} - \boldsymbol{A}^{(r)} \right\|^2 + \left\| \boldsymbol{A} - \boldsymbol{A}^{(r)} \right\|.$$
(D.60)

Let  $\epsilon' > 0$  be such that

$$\max_{\boldsymbol{A}\in\mathcal{H}_{\epsilon'}} \left\| G(\boldsymbol{A}) - \boldsymbol{A}^{(r)} \right\| = \epsilon \le \left( \frac{C_1 \lambda_{\max}}{2} \epsilon' + 1 \right) \epsilon'.$$
(D.61)

Therefore, for all  $A \in \mathcal{H}_{\epsilon'}$ ,  $G(A) \in \mathcal{H}_{\epsilon}$ . Now, using the proof technique used in [147,

Theorem 4.4.2, we define the set

$$\mathcal{V} = \{ \boldsymbol{A} \in \mathcal{H}_{\epsilon} : g(\boldsymbol{A}) < \alpha \} \subseteq \mathcal{H}_{\epsilon}, \tag{D.62}$$

where  $\alpha$  is defined as below:

$$\alpha = \begin{cases} \min_{\boldsymbol{B} \in \mathcal{H}_{\epsilon} \setminus \mathcal{H}_{\epsilon'}} G(\boldsymbol{B}) & \epsilon' \leq \epsilon \\ \infty & \epsilon' > \epsilon. \end{cases}$$
(D.63)

Note that, when  $\epsilon' \leq \epsilon$ ,  $g(\mathbf{A}) \geq \alpha$ , for all  $\mathbf{A} \in \mathcal{H}_{\epsilon} \setminus \mathcal{H}_{\epsilon'}$ . Thus,  $\mathcal{V} \subseteq \mathcal{H}_{\epsilon'}$ . Also, when  $\epsilon' > \epsilon$ ,  $\mathcal{H}_{\epsilon'} \supset \mathcal{H}_{\epsilon'} \supseteq \mathcal{V}$ . Therefore, in both cases,  $\mathcal{V} \subseteq \mathcal{H}_{\epsilon'}$ . Hence, for every  $\mathbf{A} \in \mathcal{V}$ ,  $G(\mathbf{A}) \in \mathcal{H}_{\epsilon}$ . Further, by Theorem 6.1, the sequence  $g(G^{(u)}(\mathbf{A}))$  generated by ALS is nonincreasing, and thus

$$g(G(\boldsymbol{A})) \le g(\boldsymbol{A}) < \alpha. \tag{D.64}$$

Therefore,  $G(\mathbf{A}) \in \mathcal{V}$  for all  $\mathbf{A} \in \mathcal{V}$ , hence  $G^{(u)}(\mathbf{A}) \in \mathcal{V} \subseteq \mathcal{U}$  for all  $u \in \mathbb{N}$ . Thus, stability of the point is guaranteed. Moreover, since by assumption  $\mathbf{A}^{(r)}$  is the only critical point (strict local minimum) of g in  $\mathcal{V}$ , it follows that  $\lim_{u\to\infty} G^{(u)}(\mathbf{A}) = \mathbf{A}^{(r)}$ , for all  $\mathbf{A} \in \mathcal{V}$ , which shows the asymptotic stability of  $\mathbf{A}^{(r)}$ . This completes the proof.

### D.6 Proof of Proposition 6.3

*Proof.* Under noiseless condition, the dictionary learning problem reduces to a matrix factorization problem:  $\mathbf{Y} = \mathbf{A}\mathbf{X}$ . Suppose that  $\mathbf{X}$  is already known to the algorithm. Then, to uniquely estimate  $\mathbf{A}$ , the condition (6.25) is necessary. Similarly, when  $\mathbf{A}$  is known to the algorithm, to uniquely estimate the sparse  $\mathbf{X}$ , the condition (6.26) is necessary. Thus, the first part of the result is obtained.

Next, consider the special case of  $\max_{k=1,2,...,K} \|\boldsymbol{x}_k\|_0 = 1$ . Then, every nonzero measurement vector is a scaled version of some column of the measurement matrix. The condition (6.25) guarantees that there is no all-zero row in  $\boldsymbol{X}$  and thus, there exists a measurement vector  $\boldsymbol{y}_k$  corresponding to every column  $\boldsymbol{A}_i$  of the dictionary such that  $\boldsymbol{y}_k = \boldsymbol{X}_{ik}\boldsymbol{A}_i$  where  $\boldsymbol{X}_{ik}$  is the only nonzero entry of the  $k^{\text{th}}$  column of  $\boldsymbol{X}$ . Further, by assumption, the columns of  $\boldsymbol{A}$  are unit norm, and hence, given  $\boldsymbol{y}_k$ , the tuple  $(\boldsymbol{X}_{ik}, \boldsymbol{A}_i)$  is unique upto the sign of  $\boldsymbol{X}_{ik}$ . Thus, the solution is unique under (6.25) and (6.26).

## D.7 Proof of Theorem 6.4

*Proof.* The cost function T in (6.3) consists of two terms: the logarithm of the determinant of the product of matrices of the form  $\sigma^2 I + A^* \Gamma_k^* A^{*\mathsf{T}}$ , and sum of projections of the inverses of the same matrices. Since the second term is positive, the minimum is achieved when the first term goes to minus infinity while maintaining some finite upper bound on the second term. We note that, from (6.26)

$$\begin{split} \lim_{\sigma^2 \to 0} \operatorname{\mathsf{Rank}} \left\{ \sigma^2 \boldsymbol{I} + \boldsymbol{A}^* \boldsymbol{\Gamma}_k^* \boldsymbol{A}^{*\mathsf{T}} \right\} &\leq \operatorname{\mathsf{Rank}} \left\{ \boldsymbol{\Gamma}_k^* \right\} \\ &= \|\operatorname{\mathsf{Diag}} \left\{ \boldsymbol{\Gamma}_k^* \right\} \|_0 < \frac{1}{2} \operatorname{Spark} \left\{ \boldsymbol{A}^* \right\} \leq \frac{m+1}{2} \leq m. \quad (\mathrm{D.65}) \end{split}$$

Thus, the first term goes to minus infinity. Using arguments similar to those in [23, Theorem 1], we can show that

$$\lim_{\sigma^2 \to 0} \boldsymbol{y}_k^{\mathsf{T}} \left( \sigma^2 \boldsymbol{I} + \boldsymbol{A}^* \boldsymbol{\Gamma}_k^* \boldsymbol{A}^{*\mathsf{T}} \right)^{-1} \boldsymbol{y}_k \le \frac{1}{c} \left\| \boldsymbol{x}_k^* \right\|^2.$$
(D.66)

Thus, the second term in the cost function is upper bounded by  $\frac{1}{c} \| \boldsymbol{X}^* \|_F^2 < \infty$ . Hence,  $\left( \boldsymbol{A}^*, \{ \boldsymbol{\Gamma}_k^* \}_{k=1}^K \right)$  achieves global minimum. Further, it is easy to see that the cost function takes the same value over the set  $\left( \boldsymbol{A}^* \boldsymbol{P}, \{ \boldsymbol{P} \boldsymbol{\Gamma}_k^* \boldsymbol{P} \}_{k=1}^K \right)$ , and thus the result is proved.  $\Box$ 

# D.8 Proof of Theorem 6.5

*Proof.* It is easy to see that the goal of DL-SBL is to solve the optimization problem:

$$\min_{\boldsymbol{A}\in\mathbb{O}}\left[\sum_{k=1}^{K}\min_{\boldsymbol{\gamma}_{k}\in\mathbb{R}^{N}_{+}}\log\left|\sigma^{2}\boldsymbol{I}+\boldsymbol{A}\boldsymbol{\Gamma}_{k}\boldsymbol{A}^{\mathsf{T}}\right|+\boldsymbol{y}_{k}^{\mathsf{T}}\left(\sigma^{2}\boldsymbol{I}+\boldsymbol{A}\boldsymbol{\Gamma}_{k}\boldsymbol{A}^{\mathsf{T}}\right)^{-1}\right].$$
(D.67)

For any given A, the local minima of the objective function of the sub-optimization problem within the square brackets is at most m-sparse [23, Theorem 2]. Hence, the local minima of the DL-SBL cost function are all at most m-sparse.

## Bibliography

- K. Zhou, J. C. Doyle, K. Glover *et al.*, *Robust and optimal control*. Prentice hall New Jersey, 1996, vol. 40.
- [2] B. Anderson and J. Moore, *Optimal filtering*. Courier Dover, 2005.
- [3] R. Prasad, C. Murthy, and B. Rao, "Joint approximately sparse channel estimation and data detection in OFDM systems using sparse Bayesian learning," *IEEE Trans. Signal Process.*, vol. 62, no. 14, pp. 3591–3603, Jul. 2014.
- [4] P. J. Brockwell, R. A. Davis, and M. V. Calder, Introduction to time series and forecasting. Springer, 2002, vol. 2.
- [5] C. A. Pope III, R. T. Burnett, M. J. Thun, E. E. Calle, D. Krewski, K. Ito, and G. D. Thurston, "Lung cancer, cardiopulmonary mortality, and long-term exposure to fine particulate air pollution," *JAMA*, vol. 287, no. 9, pp. 1132–1141, Mar. 2002.
- [6] M. Shao, X. Tang, Y. Zhang, and W. Li, "City clusters in China: air and surface water pollution," *Front. Ecol. Environ.*, vol. 4, no. 7, pp. 353–361, Sep. 2006.
- [7] G. Neumann, T. Noda, and Y. Kawaoka, "Emergence and pandemic potential of swine-origin H1N1 influenza virus," *Nature*, vol. 459, no. 7249, pp. 931–939, Jun. 2009.

- [8] M. Hvistendahl, D. Normile, and J. Cohen, "Despite large research effort, H7N9 continues to baffle," *Science*, vol. 340, no. 6131, pp. 414–415, Apr. 2013.
- [9] L. Donoho, "Compressed sensing," *IEEE Trans. Inf. Theory*, vol. 52, no. 4, pp. 1289–1306, Apr. 2006.
- [10] E. J. Candes, J. Romberg, and T. Tao, "Robust uncertainty principles: exact signal reconstruction from highly incomplete frequency information," *IEEE Trans. Inf. Theory*, vol. 52, no. 2, pp. 489–509, Feb. 2006.
- [11] R. G. Baraniuk, "Compressive sensing [lecture notes]," *IEEE Signal Process. Mag.*, vol. 24, no. 4, pp. 118–121, Jul. 2007.
- [12] S. S. Chen, D. L. Donoho, and M. A. Saunders, "Atomic decomposition by basis pursuit," SIAM review, vol. 43, no. 1, pp. 129–159, 2001.
- [13] R. Tibshirani, "Regression shrinkage and selection via the lasso," . J. Roy. Stat. Soc.
   B, vol. 58, no. 1, pp. 267–288, Jan. 1996.
- [14] E. Candes and T. Tao, "The Dantzig selector: Statistical estimation when p is much larger than n," Ann. Stat., vol. 35, no. 6, pp. 2313–2351, 2007.
- T. Blumensath and M. E. Davies, "Iterative thresholding for sparse approximations,"
   J. Fourier Anal., vol. 14, no. 5-6, pp. 629–654, Dec. 2008.
- [16] S. G. Mallat and Z. Zhang, "Matching pursuits with time-frequency dictionaries," *IEEE Trans. Signal Process.*, vol. 41, no. 12, pp. 3397–3415, Dec 1993.
- [17] S. Qian and D. Chen, "Signal representation using adaptive normalized Gaussian functions," *Signal Process.*, vol. 36, no. 1, pp. 1–11, Mar. 1994.

- [18] G. M. Davis, S. G. Mallat, and Z. Zhang, "Adaptive time-frequency decompositions," Opt. Eng., vol. 33, no. 7, pp. 2183–2192, Jul. 1994.
- [19] Y. C. Pati, R. Rezaiifar, and P. S. Krishnaprasad, "Orthogonal matching pursuit: Recursive function approximation with applications to wavelet decomposition," in *Proc. ASILOMAR*, Nov. 1993, pp. 40–44.
- [20] S. Chen, S. A. Billings, and W. Luo, "Orthogonal least squares methods and their application to non-linear system identification," *Intl. J. Control*, vol. 50, no. 5, pp. 1873–1896, Nov. 1989.
- [21] D. Needell and J. A. Tropp, "CoSaMP: Iterative signal recovery from incomplete and inaccurate samples," *Appl. Comput. Harmon. Anal.*, vol. 26, no. 3, pp. 301–321, May 2009.
- [22] M. E. Tipping, "Sparse Bayesian learning and the relevance vector machine," J. Mach. Learn. Res., vol. 1, pp. 211–214, Sep. 2001.
- [23] D. Wipf and B. Rao, "Sparse Bayesian learning for basis selection," *IEEE Trans. Signal Process.*, vol. 52, no. 8, pp. 2153–2164, Aug. 2004.
- [24] J. P. Vila and P. Schniter, "Expectation-maximization Gaussian-mixture approximate message passing," *IEEE Trans. Signal Process.*, vol. 61, no. 19, pp. 4658–4672, Oct 2013.
- [25] S. Som and P. Schniter, "Compressive imaging using approximate message passing and a Markov-tree prior," *IEEE Trans. Signal Process.*, vol. 60, no. 7, pp. 3439–3448, July 2012.

- [26] P. Schniter and S. Rangan, "Compressive phase retrieval via generalized approximate message passing," *IEEE Trans. Signal Process.*, vol. 63, no. 4, pp. 1043–1055, Feb 2015.
- [27] E. J. Candes and T. Tao, "Decoding by linear programming," IEEE Trans. Inf. Theory, vol. 51, no. 12, pp. 4203–4215, Dec. 2005.
- [28] —, "Near-optimal signal recovery from random projections: Universal encoding strategies?" *IEEE Trans. Inf. Theory*, vol. 52, no. 12, pp. 5406–5425, Dec 2006.
- [29] T. T. Cai and A. Zhang, "Sharp RIP bound for sparse signal and lowrank matrix recovery," *CoRR*, vol. abs/1302.1236, 2013. [Online]. Available: http://arxiv.org/abs/1302.1236
- [30] —, "Sparse representation of a polytope and recovery of sparse signals and low-rank matrices," *IEEE Trans. Inf. Theory*, vol. 60, no. 1, pp. 122–132, Jan. 2014.
- [31] R. Zhang and S. Li, "A proof of conjecture on restricted isometry property constants  $\delta_{tk} \left(0 < t < \frac{4}{3}\right)$ ," *IEEE Trans. Inf. Theory*, vol. 64, no. 3, pp. 1699–1705, Mar. 2018.
- [32] S. Foucart and H. Rauhut, A Mathematical Introduction to Compressive Sensing. Birkhäuser, 2013.
- [33] J. Wen, Z. Zhou, J. Wang, X. Tang, and Q. Mo, "A sharp condition for exact support recovery of sparse signals with orthogonal matching pursuit," in *Proc. IEEE Int. Symp. Inf. Theory*, Jul. 2016.
- [34] Y. C. Eldar, P. Kuppinger, and H. Bolcskei, "Block-sparse signals: Uncertainty

relations and efficient recovery," *IEEE Trans. Signal Process.*, vol. 58, no. 6, pp. 3042–3054, Jun. 2010.

- [35] M. Stojnic, F. Parvaresh, and B. Hassibi, "On the reconstruction of block-sparse signals with an optimal number of measurements," *IEEE Trans. Signal Process.*, vol. 57, no. 8, pp. 3075–3085, Aug. 2009.
- [36] K. Li, C. R. Rojas, T. Yang, H. Hjalmarsson, K. H. Johansson, and S. Cong, "Piecewise sparse signal recovery via piecewise orthogonal matching pursuit," in *Proc. ICASSP*, Mar. 2016, pp. 4608–4612.
- [37] S. F. Cotter, B. D. Rao, K. Engan, and K. Kreutz-Delgado, "Sparse solutions to linear inverse problems with multiple measurement vectors," *IEEE Trans. Signal Process.*, vol. 53, no. 7, pp. 2477–2488, Jul. 2005.
- [38] J. A. Tropp, A. C. Gilbert, and M. J. Strauss, "Algorithms for simultaneous sparse approximation. part I: Greedy pursuit," *IEEE Trans. Signal Process.*, vol. 86, no. 3, pp. 572–588, Mar. 2006.
- [39] J. D. Blanchard, M. Cermak, D. Hanle, and Y. Jing, "Greedy algorithms for joint sparse recovery," *IEEE Trans. Signal Process.*, vol. 62, no. 7, pp. 1694–1704, Apr. 2014.
- [40] J. Ziniel and P. Schniter, "Efficient high-dimensional inference in the multiple measurement vector problem," *IEEE Trans. Signal Process.*, vol. 61, no. 2, pp. 340–354, Jan. 2013.
- [41] D. Wipf and B. Rao, "An empirical Bayesian strategy for solving the simultaneous

sparse approximation problem," *IEEE Trans. Signal Process.*, vol. 55, no. 7, pp. 3704–3716, Jul. 2007.

- [42] R. Zdunek and A. Cichocki, "Improved M-FOCUSS algorithm with overlapping blocks for locally smooth sparse signals," *IEEE Trans. Signal Process.*, vol. 56, no. 10, pp. 4752–4761, Oct. 2008.
- [43] Z. Zhang and B. D. Rao, "Sparse signal recovery in the presence of correlated multiple measurement vectors," in *Proc. ICASSP*, Mar. 2010.
- [44] A. L. Lloyd and R. M. May, "How viruses spread among computers and people?" Science, vol. 292, no. 5520, pp. 1316–1317, May 2001.
- [45] P. Wang, M. C. González, C. A. Hidalgo, and A.-L. Barabási, "Understanding the spreading patterns of mobile phone viruses," *Science*, vol. 324, no. 5930, pp. 1071– 1076, May 2009.
- [46] M. Cha, A. Mislove, and K. P. Gummadi, "A measurement-driven analysis of information propagation in the flickr social network," in *Proc. WWW*, Apr. 2009.
- [47] Z.-L. Hu, X. Han, Y.-C. Lai, and W.-X. Wang, "Optimal localization of diffusion sources in complex networks," *Roy. Soc. Open Science*, vol. 4, no. 4, p. 170091, Apr. 2017.
- [48] M. Nagahara and D. E. Quevedo, "Sparse representations for packetized predictive networked control," *IFAC Proc. Vol.*, vol. 44, no. 1, pp. 84–89, Jan. 2011.
- [49] Z. Li, Y. Xu, H. Huang, and S. Misra, "Sparse control and compressed sensing

in networked switched systems," *IET Control Theory & Appl.*, vol. 10, no. 9, pp. 1078–1087, Jun. 2016.

- [50] G. Joseph and C. R. Murthy, "On the observability of a linear system with a sparse initial state," *IEEE Signal Process. Lett.*, vol. 25, no. 7, pp. 994 – 998, Jul. 2018.
- [51] —, "Measurement bounds for observability of linear dynamical systems under sparsity constraints," *IEEE Trans. Signal Process.*, vol. 67, no. 8, pp. 1992–2006, Feb. 2019.
- [52] —, "Controllability of linear dynamical systems under input sparsity constraints," Submitted, IEEE Trans. Autom. Control, 2019.
- [53] G. Joseph, C. R. Murthy, R. Prasad, and B. D. Rao, "Online recovery of temporally correlated sparse signals using multiple measurement vectors," in *Proc. Globecom*, Dec. 2015.
- [54] G. Joseph and C. R. Murthy, "A noniterative online bayesian algorithm for the recovery of temporally correlated sparse vectors," *IEEE Trans. Signal Process.*, vol. 65, no. 20, pp. 5510–5525, Oct. 2017.
- [55] ——, "A Bayesian algorithm for joint dictionary learning and sparse signal recovery," Submitted, IEEE Trans. Signal Process., 2018.
- [56] G. Joseph, A. B. Zoubi, C. R. Murthy, and V. J. Mathews, "Anomaly imaging for structural health monitoring exploiting clustered sparsity," in *Proc. ICASSP*, May 2019.

- [57] R. Kalman, "On the general theory of control systems," IRE Tran. Autom. Control, vol. 4, no. 3, pp. 110–110, Dec 1959.
- [58] C. T. Chen, *Linear System Theory and Design*. Oxford University Press, 1999.
- [59] C. Böß, "Using model reduction techniques within the incremental 4D-Var method," Ph.D. dissertation, Universität Bremen, 2008.
- [60] M. Verlaan, "Efficient Kalman filtering algorithms for hydrodynamic models," Ph.D. dissertation, Delft University of Technology, 1998.
- [61] A. Lawless, N. Nichols, C. Boess, and A. Bunse-Gerstner, "Using model reduction methods within incremental four-dimensional variational data assimilation," *Monthly Weather Review*, vol. 136, no. 4, pp. 1511–1522, Apr. 2008.
- [62] M. B. Wakin, B. M. Sanandaji, and T. L. Vincent, "On the observability of linear systems from random, compressive measurements," in *Proc. CDC*, Dec. 2010.
- [63] B. M. Sanandaji, M. B. Wakin, and T. L. Vincent, "Technical report: Observability with random observations," *CoRR*, vol. abs/1211.4077, 2013. [Online]. Available: http://arxiv.org/abs/1211.4077
- [64] —, "Observability with random observations," *IEEE Trans. Autom. Control*, vol. 59, no. 11, pp. 3002–3007, Nov. 2014.
- [65] K. P. Singh, A. Malik, and S. Sinha, "Water quality assessment and apportionment of pollution sources of Gomti river (India) using multivariate statistical techniques-a case study," *Analytica Chimica Acta*, vol. 538, no. 1, pp. 355 – 374, May 2005.

- [66] S. Bhattacharya and T. Basar, "Sparsity based feedback design: A new paradigm in opportunistic sensing," in *Proc. ACC*, Jun. 2011.
- [67] W. Dai and S. Yüksel, "Observability of a linear system under sparsity constraints," *IEEE Trans. Autom. Control*, vol. 58, no. 9, pp. 2372–2376, Sep. 2013.
- [68] S. Sefati, N. J. Cowan, and R. Vidal, "Linear systems with sparse inputs: Observability and input recovery," in *Proc. ACC*, Jul. 2015.
- [69] H. Rauhut, "Compressive sensing and structured random matrices," 2011.
- [70] M. Rudelson and R. Vershynin, "On sparse reconstruction from Fourier and Gaussian measurements," *Comm. Pure Appl. Math.*, vol. 61, no. 8, pp. 1025–1045, Aug. 2008.
- [71] J. Haupt, W. U. Bajwa, G. Raz, and R. Nowak, "Toeplitz compressed sensing matrices with applications to sparse channel estimation," *IEEE Trans. Inf. Theory*, vol. 56, no. 11, pp. 5862–5875, Nov 2010.
- [72] F. Krahmer, S. Mendelson, and H. Rauhut, "Suprema of chaos processes and the restricted isometry property," *Comm. Pure Appl. Math.*, vol. 67, no. 11, pp. 1877– 1904, Nov. 2014.
- [73] H. Rauhut, J. K. Romberg, and J. A. Tropp, "Restricted isometries for partial random circulant matrices," *Appl. Comput. Harmon. Anal.*, vol. 32, no. 2, pp. 242– 254, Mar. 2012.
- [74] A. Eftekhari, H. Yap, C. Rozell, and M. Wakin, "The restricted isometry property

for random block diagonal matrices," *Appl. Comput. Harmon. Anal.*, vol. 38, no. 1, pp. 1–31, Jan. 2015.

- [75] S. Khanna and C. R. Murthy, "On the restricted isometry of the columnwise Khatri-Rao product," *IEEE Trans. Signal Process.*, vol. 66, no. 5, pp. 1170 – 1183, Dec. 2017.
- [76] P. Rigollet, "18.S997: High-Dimensional Statistics, Cambridge, MA, USA:MIT OpenCourseWare," Jul. 2015. [Online]. Available: http://www-math.mit.edu/ rigollet/PDFs/RigNotes15.pdf
- [77] Q. Mo and S. Li, "New bounds on the restricted isometry constant  $\delta_{2k}$ ," Appl. Comput. Harmon. Anal., vol. 31, no. 3, pp. 460 – 468, 2011.
- [78] T. T. Cai, L. Wang, and G. Xu, "New bounds for restricted isometry constants," *IEEE Trans. Inf. Theory*, vol. 56, no. 9, pp. 4388–4394, Sep. 2010.
- [79] T. Tao and V. Vu, "Random matrices: The distribution of the smallest singular values," *Geom. Funct. Anal.*, vol. 20, no. 1, pp. 260–297, Jun. 2010.
- [80] D. Chakrabarti, Y. Wang, C. Wang, J. Leskovec, and C. Faloutsos, "Epidemic thresholds in real networks," ACM Trans. Inf. Syst. Security, vol. 10, no. 4, p. 1, Jan 200-8.
- [81] R. Durrett, "Some features of the spread of epidemics and information on a random graph," Proc. of the National Acad. of Sciences, Feb. 2010.
- [82] —, Random graph dynamics. Cambridge university press Cambridge, 2007.

- [83] M. E. Newman, S. Forrest, and J. Balthrop, "Email networks and the spread of computer viruses," *Physical Review E*, vol. 66, no. 3, p. 035101, Sep. 2002.
- [84] L. A. Pastur and M. Shcherbina, Eigenvalue distribution of large random matrices. American Mathematical Soc., 2011.
- [85] F. Pasqualetti, S. Zampieri, and F. Bullo, "Controllability metrics, limitations and algorithms for complex networks," *IEEE Trans. Control Netw. Syst.*, vol. 1, no. 1, pp. 40–52, Mar. 2014.
- [86] S. C. Tatikonda, "Control under communication constraints," Ph.D. dissertation, Massachusetts Institute of Technology, 2000.
- [87] Y.-Y. Liu and A.-L. Barabási, "Control principles of complex systems," Rev. Mod. Phys., vol. 88, no. 3, p. 035006, Sep. 2016.
- [88] P. V. Chanekar, N. Chopra, and S. Azarm, "Optimal actuator placement for linear systems with limited number of actuators," in *Proc. ACC*, May 2017, pp. 334–339.
- [89] E. Nozari, F. Pasqualetti, and J. Cortés, "Time-invariant versus time-varying actuator scheduling in complex networks," in *Proc. ACC*, May 2017, pp. 4995–5000.
- [90] M. Hautus, "Stabilization controllability and observability of linear autonomous systems," in *Indagationes mathematicae (proceedings)*, vol. 73, Jan. 1970, pp. 448–455.
- [91] A. Jadbabaie, A. Olshevsky, and M. Siami, "Deterministic and randomized actuator scheduling with guaranteed performance bounds," arXiv preprint arXiv:1805.00606, May 2018.

- [92] A. Olshevsky, "Minimal controllability problems." *IEEE Trans. Control Netw. Syst.*, vol. 1, no. 3, pp. 249–258, Sep. 2014.
- [93] V. Tzoumas, M. A. Rahimian, G. J. Pappas, and A. Jadbabaie, "Minimal actuator placement with bounds on control effort," *IEEE Trans. Control Netw. Syst.*, vol. 3, no. 1, pp. 67–78, Mar. 2016.
- [94] A. Jadbabaie, A. Olshevsky, G. J. Pappas, and V. Tzoumas, "Minimal reachability is hard to approximate," *IEEE Trans. Autom. Control*, vol. 64, no. 2, pp. 783–789, Dec. 2015.
- [95] Z. Liu, Y. Long, A. Clark, P. Lee, L. Bushnell, D. Kirschen, and R. Poovendran, "Minimal input selection for robust control," *CoRR*, vol. abs/1712.01232, 2017.
  [Online]. Available: http://arxiv.org/abs/1712.01232
- [96] A. S. Charles, H. L. Yap, and C. J. Rozell, "Short-term memory capacity in networks via the restricted isometry property," *Neural Comput.*, vol. 26, no. 6, pp. 1198–1235, Jun. 2014.
- [97] M. Kafashan, A. Nandi, and S. Ching, "Relating observability and compressed sensing of time-varying signals in recurrent linear networks," *Neural Networks*, vol. 83, pp. 11–20, Nov. 2016.
- [98] V. Y. Pan and Z. Q. Chen, "The complexity of the matrix eigenproblem," in Proc. STOC, May 1999, pp. 507–516.
- [99] C.-T. Chen, Linear system theory and design. Oxford University Press, Inc., 1998.

- [100] R. E. Kalman, "Mathematical description of linear dynamical systems," SIAM J. Control, vol. 1, no. 2, pp. 152–192, 1963.
- [101] L. C. Westphal, Handbook of Control Systems Engineering. Springer Science & Business Media, 2012.
- [102] F. E. Hohn, *Elementary matrix algebra*. Courier Corporation, 2013.
- [103] D. Malioutov, M. Cetin, and A. S. Willsky, "A sparse signal reconstruction perspective for source localization with sensor arrays," *IEEE Trans. Signal Process.*, vol. 53, no. 8, pp. 3010–3022, Aug. 2005.
- [104] J. H. G. Ender, "On compressive sensing applied to radar," Signal Processing, vol. 90, no. 5, pp. 1402–1414, May 2010.
- [105] I. F. Gorodnitsky, J. S. George, and B. D. Rao, "Neuromagnetic source imaging with FOCUSS: a recursive weighted minimum norm algorithm," *Electroencephalogr. Clin. Neurophysiol.*, vol. 95, no. 4, pp. 231–251, Oct. 1995.
- [106] D. Wipf, J. Owen, H. Attias, K. Sekihara, and S. Nagarajan, "Robust Bayesian estimation of the location, orientation, and time course of multiple correlated neural sources using MEG," *NeuroImage*, vol. 49, no. 1, pp. 641–655, Jan. 2010.
- [107] U. Gamper, P. Boesiger, and S. Kozerke, "Compressed sensing in dynamic MRI," Magn. Reson. Med., vol. 59, no. 2, pp. 365–373, Feb. 2008.
- [108] Z. Zhang, T.-P. Jung, S. Makeig, Z. Pi, and B. D. Rao, "Spatiotemporal sparse

Bayesian learning with applications to compressed sensing of multichannel physiological signals," *IEEE Trans. Neural Syst. Rehabil. Eng.*, vol. 22, no. 6, pp. 1186–1197, Nov. 2014.

- [109] N. Vaswani, "LS-CS-residual (LS-CS): Compressive sensing on least squares residual," *IEEE Trans. Signal Process.*, vol. 58, no. 8, pp. 4108–4120, Aug 2010.
- [110] X. Zhu, L. Dai, W. Dai, Z. Wang, and M. Moonen, "Tracking a dynamic sparse channel via differential orthogonal matching pursuit," in *Proc. MILCOM*, Oct. 2015.
- [111] A. S. Charles, A. Balavoine, and C. J. Rozell, "Dynamic filtering of time-varying sparse signals via l<sub>1</sub> minimization," *IEEE Trans. Signal Process.*, vol. 64, no. 21, pp. 5644–5656, Nov 2016.
- [112] N. Vaswani, "Kalman filtered compressed sensing," in *ICIP*, Oct. 2008.
- [113] E. Karseras, K. K. Leung, and W. Dai, "Tracking dynamic sparse signals using hierarchical Bayesian Kalman filters," in *Proc. ICASSP*, May 2013.
- [114] R. Chalasani and J. C. Principe, "Dynamic sparse coding with smoothing proximal gradient method," in *Proc. ICASSP*, May 2014.
- [115] J. B. Moore, P. K. S. Tam, C. John, and M. Richardson, "Fixed-lag smoothing of nonlinear systems with discrete measurements," *Information Sciences*, vol. 6, pp. 151–160, 1973.
- [116] V. Krishnamurthy and J. B. Moore, "On-line estimation of hidden markov model parameters based on the Kullback-Leibler information measure," *IEEE Trans. Signal Process.*, vol. 41, no. 8, pp. 2557–2573, Aug. 1993.

- [117] R. Hunger, "Floating point operations in matrix-vector calculus," Munich University of Technology, TUM-LNS-TR-05-05, Tech. Rep. TUM-LNS-TR-05-05, Sep. 2007.
- [118] Z. Zhang and B. D. Rao, "Sparse signal recovery with temporally correlated source vectors using sparse Bayesian learning," *IEEE Trans. Signal Process.*, vol. 5, no. 5, pp. 912–926, Sep. 2011.
- [119] S. Chen, D. Donoho, and M. Saunders, "Atomic decomposition by basis pursuit," SIAM Rev., vol. 43, no. 1, pp. 129–159, Jan. 2001.
- [120] I.-R. Recommendation, "Guidelines for evaluation of radio transmission technologies (RTTs) for IMT-2000," ITU, Tech. Rep. M.1225, Feb. 1997.
- [121] C. Studer, C. Benkeser, S. Belfanti, and Q. Huang, "Design and implementation of a parallel turbo-decoder ASIC for 3GPP-LTE," *IEEE J. Solid-State Circuits*, vol. 46, no. 1, pp. 8–17, Jan. 2011.
- [122] J. Zyren and W. McCoy, "Overview of the 3GPP long term evolution physical layer," Freescale Semiconductor, Inc., Austin, TX, USA, Tech. Rep. 3GPPEVO-LUTIONWP, Jul. 2007.
- [123] Y. Zheng and C. Xiao, "Simulation models with correct statistical prop- erties for Rayleigh fading channels," *IEEE Trans. Commun.*, vol. 6, no. 51, pp. 920–928, Jun. 2003.
- [124] E. SMG, "Universal mobile telecommunications system (UMTS), selection procedures for the choice of radio transmission technologies of the UMTS," ETSI, Sophia-Antipolis, France, Tech. Rep. UMTS 21.01 version 3.0.1, Nov. 1997.

- [125] R. Rubinstein, A. M. Bruckstein, and M. Elad, "Dictionaries for sparse representation modeling," *Proc. IEEE*, vol. 98, no. 6, pp. 1045–1057, Jun. 2010.
- [126] M. Elad and M. Aharon, "Image denoising via sparse and redundant representations over learned dictionaries," *IEEE Trans. Image Process.*, vol. 54, no. 12, pp. 3736– 3745, Dec. 2006.
- [127] J. Mairal, M. Elad, and G. Sapiro, "Sparse representation for color image restoration," *IEEE Trans. Image Process.*, vol. 17, no. 1, pp. 53–69, Jan. 2008.
- [128] J. Mairal, F. Bach, J. Ponce, G. Sapiro, and A. Zisserman, "Nonlocal sparse models for image restoration," in *ICCV*, Sep. 2009, pp. 2272–2279.
- [129] R. Grosse, R. Raina, H. Kwong, and A. Y. Ng, "Shift-invariant sparse coding for audio classification," in Proc. Conf. Uncertainty in Artificial Intelligence, Jul. 2007.
- [130] M. Zibulevsky and B. Pearlmutter, "Blind source separation by sparse decomposition in a signal dictionary," *Neural Computation*, vol. 13, no. 4, pp. 863–882, Apr. 2001.
- [131] R. Raina, A. Battle, H. Lee, B. Packer, and A. Y. Ng, "Self-taught learning: Transfer learning from unlabeled data," in *Proc. ICML*, Jun. 2007, pp. 759–766.
- [132] J. Mairal, F. Bach, J. Ponce, G. Sapiro, and A. Zisserman, "Discriminative learned dictionaries for local image analysis," in *Proc. CVPR*, Jun. 2008, pp. 1–8.
- [133] —, "Supervised dictionary learning," in Proc. Adv. in Neural Inform. Process. Syst., Dec. 2009, pp. 1033–1040.
- [134] D. Bradley and J. Bagnell, "Differentiable sparse coding," in Proc. Adv. in Neural Inform. Process. Syst., Dec. 2009, pp. 113–120.

- [135] K. Kavukcuoglu, M. Ranzato, R. Fergus, and Y. LeCun, "Learning invariant features through topographic filter maps," in *Proc. CVPR*, Jun. 2009, pp. 1605–1612.
- [136] J. Yang, K. Yu, Y. Gong, and T. Huang, "Linear spatial pyramid matching using sparse coding for image classification," in *Proc. CVPR*, Jun. 2009, pp. 1794 – 1801.
- [137] J. Mairal, F. Bach, and J. Ponce, "Task-driven dictionary learning," IEEE Trans. Pattern Anal. Mach. Intell., vol. 34, no. 4, pp. 791–804, Apr. 2012.
- [138] K. Engan, S. O. Aase, and J. H. Husoy, "Method of optimal directions for frame design," in *Proc. ICASSP*, Mar. 1999.
- [139] M. Aharon, M. Elad, and A. Bruckstein, "K-SVD: An algorithm for designing overcomplete dictionaries for sparse representation," *IEEE Trans. Signal Process.*, vol. 54, no. 11, pp. 4311–4322, Nov. 2006.
- [140] M. Yaghoobi, T. Blumensath, and M. E. Davies, "Dictionary learning for sparse approximations with the majorization method," *IEEE Trans. Signal Process.*, vol. 57, no. 6, pp. 2178–2191, Jun. 2009.
- [141] W. Dai, T. Xu, and W. Wang, "Simultaneous codeword optimization (SimCO) for dictionary update and learning," *IEEE Trans. Signal Process.*, vol. 60, no. 12, pp. 6340–6353, Dec. 2012.
- [142] M. Sadeghi, M. Babaie-Zadeh, and C. Jutten, "Learning overcomplete dictionaries based on atom-by-atom updating," *IEEE Trans. Signal Process.*, vol. 62, no. 4, pp. 883–891, Feb. 2014.

- [143] S. K. Sahoo and A. Makur, "Dictionary training for sparse representation as generalization of K-Means clustering," *IEEE Signal Process. Lett.*, vol. 20, no. 6, pp. 587–590, Jun. 2013.
- [144] M. Zhou, H. Chen, J. Paisley, L. Ren, L. Li, Z. Xing, D. Dunson, G. Sapiro, and L. Carin, "Nonparametric bayesian dictionary learning for analysis of noisy and incomplete images," *IEEE Trans. Image Process.*, vol. 21, no. 1, pp. 130–144, Jan. 2012.
- [145] J. T. Parker, P. Schniter, and V. Cevher, "Bilinear generalized approximate message passing - part II: Applications," *IEEE Trans. Signal Process.*, vol. 62, no. 22, pp. 5854–5867, Nov. 2014.
- [146] L. Yang, J. Fang, H. Cheng, and H. Li, "Sparse Bayesian dictionary learning with a Gaussian hierarchical model," *Signal Process.*, vol. 130, pp. 93–104, Jan. 2017.
- [147] P.-A. Absil, M. Robert, and R. Sepulchre, Optimization algorithms on matrix manifolds. Princeton University Press, 2008.
- [148] P.-A. Absil and J. Malick, "Projection-like retractions on matrix manifolds," SIAM J. Optim., vol. 22, no. 1, pp. 135–158, Oct. 2012.
- [149] T. Kaneko, S. Fiori, and T. Tanaka, "Empirical arithmetic averaging over the compact Stiefel manifold," *IEEE Trans. Signal Process.*, vol. 61, no. 4, pp. 883–894, Feb. 2013.
- [150] D. Wipf and S. Nagarajan, "Iterative reweighted  $\ell_1$  and  $\ell_2$  methods for finding sparse solutions," *IEEE J. Sel. Topics Sig. Proc.*, vol. 4, no. 2, pp. 317–329, Apr. 2010.

- [151] J. Cruz Neto, L. De Lima, and P. R. Oliveira, "Geodesic algorithms in Riemannian geometry," *Balkan J. Geom. Appl*, vol. 3, no. 2, pp. 89–100, 1998.
- [152] J. Cruz Neto, O. Ferreira, and L. R. Lucambio Perez, "A proximal regularization of the steepest descent method in Riemannian manifold," *Balkan J. Geom. Appl*, vol. 4, no. 2, pp. 1–8, 1999.
- [153] Y. Yang, "Globally convergent optimization algorithms on Riemannian manifolds: Uniform framework for unconstrained and constrained optimization," J. Optim. Theory Appl., vol. 132, no. 2, pp. 245–265, 2007.
- [154] C. Li and J. Wang, "Newton's method for sections on Riemannian manifolds: Generalized covariant α-theory," J. Complex., vol. 24, no. 3, pp. 423–451, Jun. 2008.
- [155] X.-b. Li, N.-j. Huang, Q. H. Ansari, and J.-C. Yao, "Convergence rate of descent method with new inexact line-search on Riemannian manifolds," J. Optim. Theory Appl., pp. 1–25, Sep. 2018.
- [156] H. Attouch and J. Bolte, "On the convergence of the proximal algorithm for nonsmooth functions involving analytic features," *Mat. Programming*, vol. 116, no. 1-2, pp. 5–16, Jan. 2009.
- [157] H. Liu, W. Wu, and A. Man-Cho So, "Quadratic optimization with orthogonality constraints: Explicit Lojasiewicz exponent and linear convergence of line-search methods," ArXiv e-prints, Oct. 2015. [Online]. Available: https://arxiv.org/abs/1510.01025
- [158] B. Gao, X. Liu, X. Chen, and Y. xiang Yuan, "On the Lojasiewicz exponent of the

quadratic sphere constrained optimization problem," ArXiv e-prints, Nov. 2016.
[Online]. Available: https://arxiv.org/abs/1611.08781

- [159] M. Razaviyayn, H.-W. Tseng, and Z.-Q. Luo, "Dictionary learning for sparse representation: Complexity and algorithms," in *Proc. ICASSP*, May 2014, pp. 5247–5251.
- [160] M. E. Muller, "A note on a method for generating points uniformly on N-dimensional spheres," *Commun. ACM*, vol. 2, no. 4, pp. 19–20, Apr. 1959.
- [161] B. W. Drinkwater and P. D. Wilcox, "Ultrasonic arrays for non-destructive evaluation: A review," NDT & E International, vol. 39, no. 7, pp. 525–541, Oct. 2006.
- [162] R. Geng, "Modern acoustic emission technique and its application in aviation industry," Ultrasonics, vol. 44, pp. e1025–e1029, Dec. 2006.
- [163] K. Worden, C. R. Farrar, G. Manson, and G. Park, "The fundamental axioms of structural health monitoring," in *Proc. Math. Phys. Eng. Sci.*, Jun. 2007.
- [164] E. V. Malyarenko and M. K. Hinders, "Fan beam and double crosshole Lamb wave tomography for mapping flaws in aging aircraft structures," J. Acoust. Soc. Am, vol. 108, no. 4, pp. 1631–1639, Oct. 2000.
- [165] K. R. Leonard, E. V. Malyarenko, and M. K. Hinders, "Ultrasonic Lamb wave tomography," *Inverse Probl.*, vol. 18, no. 6, pp. 1795–1808, Nov. 2002.
- [166] S. M. Prasad, K. Balasubramaniam, and C. Krishnamurthy, "Structural health monitoring of composite structures using Lamb wave tomography," *Smart Mater. Struct.*, vol. 13, no. 5, pp. N73–N79, Jul. 2004.

- [167] H. Gao, Y. Shi, J. Rose, D. O. Thompson, and D. E. Chimenti, "Guided wave tomography on an aircraft wing with leave in place sensors," in *Proc. AIP*, Apr. 2005.
- [168] A. B. Zoubi and V. J. Mathews, "Anomaly imaging using decomposed Lamb wave modes," in *Proc. IWSHM*, Sep. 2017.
- [169] A. B. Zoubi, S. Kim, D. O. Adams, and V. J. Mathews, "Lamb wave mode decomposition based on cross-Wigner-Ville distribution and its application to anomaly imaging for structural health monitoring," *IEEE Trans. Ultrason., Ferroelectr., Freq. Control*, 2018, (Accepted).
- [170] D. Wang, W. Zhang, X. Wang, and B. Sun, "Lamb-wave-based tomographic imaging techniques for hole-edge corrosion monitoring in plate structures," *Materials*, vol. 9, no. 11, p. 916, Nov. 2016.
- [171] Y. C. Eldar, P. Kuppinger, and H. Bolcskei, "Block-sparse signals: Uncertainty relations and efficient recovery," *IEEE Trans. Signal Process.*, vol. 58, no. 6, pp. 3042–3054, Jun. 2010.
- [172] Y. C. Eldar and M. Mishali, "Robust recovery of signals from a structured union of subspaces," *IEEE Trans. Inf. Theory*, vol. 55, no. 11, pp. 5302–5316, Nov. 2009.
- [173] M. Yuan and Y. Lin, "Model selection and estimation in regression with grouped variables," J. Royal Stat. Soc., vol. 68, no. 1, pp. 49–67, Feb. 2006.
- [174] Z. Zhang and B. D. Rao, "Sparse signal recovery with temporally correlated source

vectors using sparse bayesian learning," *IEEE J. Sel. Topics Signal Process.*, vol. 5, no. 5, pp. 912–926, Sep. 2011.

- [175] J. Fang, Y. Shen, H. Li, and P. Wang, "Pattern-coupled sparse Bayesian learning for recovery of block-sparse signals," *IEEE Trans. Signal Process.*, vol. 63, no. 2, pp. 360–372, Jan. 2015.
- [176] J. Fang, L. Zhang, and H. Li, "Two-dimensional pattern-coupled sparse Bayesian learning via generalized approximate message passing," *IEEE Trans. Image Process.*, vol. 25, no. 6, pp. 2920–2930, Jun. 2016.
- [177] A. B. Zoubi, V. J. Mathews, J. Harley, and D. Adams, "Lamb waves mode decomposition using the cross-Wigner-Ville distribution," in *Proc. IWSHM*, Sep. 2015.
- [178] V. J. Mathews, "Damage mapping in structural health monitoring using a multi-grid architecture," in *Proc. AIP*, vol. 1650, no. 1, Mar. 2015, pp. 1247–1255.
- [179] R. M. Dudley, "The sizes of compact subsets of Hilbert space and continuity of Gaussian processes," *Journal of Functional Analysis*, vol. 1, no. 3, pp. 290–330, Oct. 1967.
- [180] R. A. Horn and C. R. Johnson, *Matrix Analysis*. Cambridge University Press, 1990.
- [181] D. Williams, *Probability with martingales*. Cambridge University Press, 1991.
- [182] V. S. Borkar, Stochastic Approximation: A Dynamical Systems Viewpoint. Cambridge University Press, 2008.
- [183] B. Delyon, "General results on the convergence of stochastic algorithms," IEEE Trans. Autom. Control, vol. 41, no. 9, pp. 1245–1255, Sep. 1996.

[184] Y. Xu and W. Yin, "A block coordinate descent method for regularized multiconvex optimization with applications to nonnegative tensor factorization and completion," *SIAM J. Imaging Sci.*, vol. 6, no. 3, pp. 1758–1789, Sep. 2013.