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Chirag Ramesh

SPC Lab, Indian Institute of Science

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Image: A matrix and a matrix

### Contributions

- Iteratively reweighted super-resolution mmWave channel estimation
- Low complexity SVD based preconditioning

System Model

$$\mathbf{Y} = \mathbf{W}^{H}\mathbf{H}\mathbf{P}\mathbf{S} + \mathbf{N} = \mathbf{W}^{H}\mathbf{H}\mathbf{X} + \mathbf{N}$$

$$\mathbf{H} = \sum_{l=1}^{L} z_{l} \mathbf{a}_{R}(\theta_{R,l}^{az}, \theta_{R,l}^{el}) \mathbf{a}_{T}^{H}(\theta_{T,l}^{az}, \theta_{T,l}^{el}) = \mathbf{A}_{R}(\theta_{R}^{\prime}) \ diag(\mathbf{z}^{\prime}) \ \mathbf{A}_{T}^{H}(\theta_{T}^{\prime})$$
$$\min_{\mathbf{z}, \theta_{R}, \theta_{T}} \|\mathbf{z}\|_{0} \ s.t. \ \|\mathbf{Y} - \mathbf{W}^{H}\mathbf{H}\mathbf{X}\|_{F} \le \epsilon$$

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#### Solution

$$\begin{split} \min_{\mathbf{z}, \theta_{R}, \theta_{T}} G(\mathbf{z}, \theta_{R}, \theta_{T}) &\triangleq \left\{ \sum_{l=1}^{L} log(|z_{l}|^{2} + \delta) + \lambda \|\mathbf{Y} - \mathbf{W}^{H} \mathbf{H} \mathbf{X}\|_{F}^{2} \right\} \\ \min_{\mathbf{z}, \theta_{R}, \theta_{T}} S^{(i)}(\mathbf{z}, \theta_{R}, \theta_{T}) &\triangleq \left\{ \lambda^{-1} \mathbf{z}^{H} \mathbf{D}^{(i)} \mathbf{z} + \|\mathbf{Y} - \mathbf{W}^{H} \mathbf{H} \mathbf{X}\|_{F}^{2} \right\} \\ \mathbf{D}^{(i)} &\triangleq diag \left\{ (log(|z_{l}^{(i)}|^{2} + \delta))^{-1} \right\} \\ \mathbf{z}^{(i)}_{opt}(\theta_{R}, \theta_{T}) &\triangleq \arg\min_{\mathbf{z}} S^{(i)}(\mathbf{z}, \theta_{R}, \theta_{T}) \end{split}$$

• 
$$\lambda = \min(d/r^{(i)}, \lambda_{max})$$

- $\theta_R \& \theta_T$  found via Gradient Descent
- SVD based preconditioning performed to reduce complexity

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#### Contributions

- Precoder design for MU-MIMO Systems with Per-Antenna Power Constraints (PAPC)
- Minimization of distance between Sum Power Constraint and PAPC decoders
- Iterative algorithm with Newton's method to solve the dual

## System Model

$$\mathbf{y}_k = \mathbf{H}_k \mathbf{x} + \mathbf{n}_k, \ \mathbf{x} = \sum_{k=1}^{K} \mathbf{W}_k \mathbf{b}_k$$

• 
$$\mathbf{W} = [\mathbf{W}_1, \mathbf{W}_2, \dots, \mathbf{W}_K]$$

• SPC:  $\mathbb{E}[tr(\mathbf{x}\mathbf{x}^H)] = tr(\mathbf{W}\mathbf{W}^H) \le p_t$ 

• PAPC: 
$$\mathbb{E}[diag(\mathbf{x}\mathbf{x}^{H})] = diag(\mathbf{W}\mathbf{W}^{H}) \leq \mathbf{p}$$

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## Solution

- ZF precoder for SPC:  $\mathbf{W}_{SPC,k} = \mathbf{V}_k \mathbf{A}_k$
- ZF precoder for PAPC:  $\mathbf{W}_k = \mathbf{V}_k \mathbf{B}_k$
- Distance between SPC and PAPC:  $d(\mathbf{W}, \mathbf{W}_{SPC}) = \|\mathbf{W} - \mathbf{W}_{SPC}\|^2 = tr((\mathbf{W} - \mathbf{W}_{SPC})(\mathbf{W} - \mathbf{W}_{SPC})^H)$

$$\min_{\mathbf{W}} d(\mathbf{W}, \mathbf{W}_{SPC}) \ \ s.t. \ \ diag(\mathbf{WW}^{H}) \leq \mathbf{p}$$

$$Primal: \max_{\mathbf{B}} \sum_{k=1}^{K} tr(\mathbf{A}_{k}\mathbf{B}_{k}^{H} + \mathbf{B}_{k}\mathbf{A}_{k}^{H})$$

s.t. 
$$diag(\sum_{k=1}^{H} \mathbf{V}_k \mathbf{B}_k \mathbf{B}_k^H \mathbf{V}_k^H) \leq \mathbf{p}$$

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$$Dual: \min_{\boldsymbol{\lambda}} tr(\sum_{k=1}^{K} (\mathbf{V}_{k}^{H} diag(\boldsymbol{\lambda}) \mathbf{V}_{k})^{-1} \mathbf{A}_{k} \mathbf{A}_{k}^{H}) + \boldsymbol{\lambda}^{T} \boldsymbol{p}$$

s.t.  $\lambda \geq 0$ 

- Slater's condition satisfied
- Newton Update:  $\lambda(i+1) = \lambda(i) \beta \phi^{-1} \bigtriangledown g$

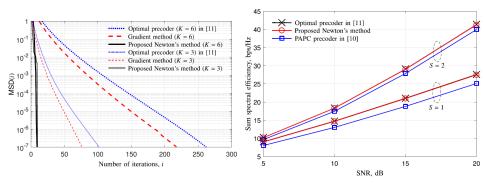
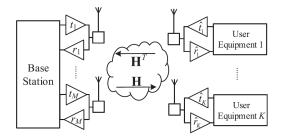


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# A General Matched Filter Design for Reciprocity Calibration in Multiuser Massive MIMO Systems

# Contributions

- Design of Calibration matrix for TDD MU-MIMO System
- Maximization of SINR in the presence of reciprocity mismatch



# System Model

- $\mathbf{H}_{UL} = diag(\{r_i\}_{i=1}^M) \mathbf{H}^T diag(\{\hat{t}_j\}_{j=1}^K) = \mathbf{R}\mathbf{H}^T \mathbf{\hat{T}}$
- $\mathbf{H}_{DL} = diag(\{\hat{r}_j\}_{j=1}^K) \mathbf{H} \ diag(\{r_i\}_{i=1}^M) = \mathbf{\hat{R}HT}$

• Downlink Channel:

$$\mathbf{y} = \sqrt{
ho_D} \mathbf{H}_{DL} \mathbf{W} \mathbf{x} + \mathbf{n}$$

• Precoding Matrix:  $\mathbf{W} = \alpha \mathbf{C} \mathbf{H}_{UL}^*$ 

$$\Rightarrow \mathbf{y} = \sqrt{\rho_D} \alpha \mathbf{HTCR}^* \mathbf{H}^H \mathbf{x} + \mathbf{n} = \sqrt{\rho_D} \mathbf{\tilde{H}} \mathbf{x} + \mathbf{n}$$

- Matched Filter:  $\mathbf{C}_1 = \lambda \mathbf{T}^* \mathbf{R}^{-*} \Rightarrow \tilde{\mathbf{H}}_1 = \alpha \lambda \mathbf{HT} (\mathbf{HT})^H$
- Decorrelator:  $\mathbf{C}_2 = \lambda \mathbf{T}^{-1} \mathbf{R}^{-*} \Rightarrow \tilde{\mathbf{H}}_2 = \alpha \lambda \mathbf{H} \mathbf{H}^H$
- Optimal C:

$$\max_{\mathbf{C}} R \equiv \max_{\mathbf{C}} \frac{|tr(\mathbf{TCR}^*)|^2}{\|\mathbf{TCR}^*\|_F^2 + \|\mathbf{CR}^*\|_F^2/\rho_D}$$

(1)

#### Results

• Proposed design yields the below for perfect CSI

$$\mathbf{C}^{GMF} = \epsilon \mathbf{T}^* \mathbf{R}^{-*} \left( \mathbf{T} \mathbf{T}^* + \frac{\mathbf{I}_M}{\rho_D} \right)^{-1}$$

• Asymptotics:

$$\mathbf{C}_{\rho_D \to 0}^{GMF} = \epsilon \mathbf{T}^* \mathbf{R}^{-*}$$

$$\mathbf{C}_{\rho_D \to \infty}^{GMF} = \epsilon \mathbf{T}^{-1} \mathbf{R}^{-*}$$

• For imperfect CSI

$$\mathbf{C}^{RGMF} = \epsilon \mathbf{T}^* \mathbf{R} \left( \mathbf{T} \mathbf{T}^* \mathbf{R} \mathbf{R}^* + \frac{\mathbf{T} \mathbf{T}^*}{\rho_U} + \frac{\mathbf{R} \mathbf{R}^*}{\rho_D} + \frac{\mathbf{I}_M}{\rho_U \rho_D} \right)^{-1}$$
$$\mathbf{C}^{RGMF}_{\rho_D \to \infty, \rho_D \to 0} = \epsilon \mathbf{T}^{-1} \mathbf{R}$$

- Compressed Sensing of Underwater Acoustic Signals via Structured Approximation *k*-Norm
- Propagation Models and Performance Evaluation for 5G Millimeter-Wave Bands
- On the Performance of Sliding Window TD-LMMSE Channel Estimation for 5G Waveforms in High Mobility Scenario
- Partial CSI Acquisition for Size-Constrained Massive MIMO Systems With User Mobility
- Secure Beamformer Designs in MU-MIMO Systems With Multiuser Interference Exploitation
- O Deep Learning for Super-Resolution Channel Estimation and DOA Estimation Based Massive MIMO System

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