

Journal Watch: Transactions on Signal Processing, August 2012
Sinchu P
28 Aug 2012

SNR and Noise Variance Estimation for MIMO Systems

Aniruddha Das, *ViaSat Inc., CA*

Bhaskar D Rao, *University of California, San Diego*

SNR and Noise Variance Estimation for MIMO Systems

System Model: $\mathbf{y}_i = \mathbf{H}\mathbf{x}_i + \mathbf{n}_i, \quad i = 1 \dots N$

$\mathbf{y}_i : M_R \times 1, \quad \mathbf{H} : M_R \times M_T, \quad \mathbf{x}_i : M_T \times 1$

$$\mathbf{X} = [\mathbf{x}_1, \mathbf{x}_2, \dots, \mathbf{x}_N]$$

SNR definition:

$$\rho_{xy} = \frac{\|h_{xy}\|^2}{\sigma^2}$$

$$\rho = \frac{\|\mathbf{H}\|_F^2}{\sigma^2}$$

CRLB for SNR estimation: Data aided(DA)

$$\Theta = [(\text{vec}(\mathbf{H}))^T (\text{vec}(\mathbf{H}^*))^T \sigma^2]^T$$

$$\text{CRLB of } \rho_{xy} = 2\alpha\rho_{xy} + \frac{\rho_{xy}^2}{NM_R}$$

$$\text{CRLB of } \rho = \frac{2\rho}{N} + \frac{\rho^2}{2N}$$

Non-data aided model(NDA)

- No closed form expression
- Numerical evaluation

CRLB of Noise Variance: DA

$$CRLB[\sigma^2] = \frac{\sigma^4}{NM_R}$$

$$CRLB[\sigma_k^2] = \frac{\sigma_k^4}{N}$$

Mixed DA and NDA model

$$N = N_p + N_d$$

$$\frac{1}{CRLB[N]_{DA_NDA}} = \frac{1}{CRLB[N_p]_{DA}} + \frac{1}{CRLB[N_d]_{NDS}}$$

SNR and Noise Variance Estimation for MIMO Systems

ML Estimators: DA

$$\widehat{SNR} = \frac{\|\hat{\mathbf{H}}\|^2}{\sigma^2}$$

1) Noise Variance

$$\hat{\sigma}_{DA_ML}^2 = \frac{1}{NM_R} \sum_{i=1}^N (\mathbf{y}_i - \mathbf{H}\mathbf{x}_i)^H (\mathbf{y}_i - \mathbf{H}\mathbf{x}_i)$$

2) Channel Coefficients

ML estimate is the one which maximizes $-\sum_{i=1}^N \|\mathbf{y}_i - \mathbf{H}\mathbf{x}_i\|^2$

ML Estimators: NDA

-No closed form solution

-Approximation in the low SNR and high SNR regimes

-EM algorithm for iteratively attaining ML estimate

Nondata-Aided Joint Channel Estimation and Equalization for OFDM Systems in Very Rapidly Varying Mobile Channels

Habib Senol, Erdal Panayirci, *Kadir Has University, Turkey*, H. Vincent Poor,
Princeton University

Nondata-Aided Joint Channel Estimation and Equalization for OFDM Systems in Very Rapidly Varying Mobile Channels

The received signal model is $\mathbf{y}(m) = \mathbf{S}(m)\mathbf{h}(m) + \mathbf{w}(m)$

Estimate $\mathbf{h} = [\mathbf{h}^T(0), \mathbf{h}^T(1), \dots, \mathbf{h}^T(M-1)]^T \in \mathcal{C}^{MNL}$ from
 $\mathbf{y} = [\mathbf{y}^T(0), \mathbf{y}^T(1), \dots, \mathbf{y}^T(M-1)]^T \in \mathcal{C}^{MN}$

Basis Expansion: $\tilde{\mathbf{h}}_l = \Psi \mathbf{c}_l$ and $\mathbf{c}_l = \Psi^\dagger \tilde{\mathbf{h}}_l$

where Ψ is the orthonormal discrete Legendre polynomial basis (DLP-BEM)

- ▶ Time domain implementation
- ▶ Exploit time diversity of the channel
- ▶ CRLB for the estimator of channel coefficient vector

Nondata-Aided Joint Channel Estimation and Equalization for OFDM Systems in Very Rapidly Varying Mobile Channels

SAGE-MAP

1. Expectation step: Compute an estimate of the log-likelihood function of 'hidden data' given \mathbf{y} and a current estimate of \mathbf{c}
2. Maximization step: c is updated as the value that maximizes this estimate of the log likelihood function

The initial channel estimate $\mathbf{c}^{(0)}$ can be determined with the aid of pilot symbols

Also computes

1. Bayesian MSE lower bound
2. Convergence rate
3. Complexity

Sparse Bayesian Methods for Low-Rank Matrix Estimation

S. Derin Babacan, *University of Illinois*,

Martin Leussi, *Harvard Medical School*,

Rafael Molina, *Universidad de Granada*,

Aggelos K. Katsaggelos, *NorthWestern University, IL*

$\mathbf{X} \in \mathbb{R}^{m \times n}$ with rank $r \ll \min(m, n)$

1. Matrix Completion: $\mathbf{Y} = f(\mathbf{X})$ such that $Y_{ij} = X_{ij} : (i, j) \in \Omega$
2. Robust PCA: $\mathbf{Y} = \mathbf{X} + \mathbf{E}$

$$\mathbf{X} = \mathbf{AB}^T$$

1. The goal is to achieve column sparsity in A and B.
2. Approximate Bayesian Inference: Compute posterior distribution approximations by minimizing the KL divergence for each latent variable
1. Estimate the effective rank
2. Sparsity of the estimate of E
3. Computational complexity

Block-Sparse Recovery via Convex Optimization
Ehsan Elhamifar, Rene Vidal
The Johns Hopkins University, Baltimore

$$\mathbf{y} = \mathbf{B}\mathbf{c}, \quad \mathbf{B} \in \mathbb{R}^{D \times N}, N > D$$

- Block-Sparse recovery using convex programs

$$P_{l_q/l_1} : \min \sum_{i=1}^n \|c[i]\|_q \text{ s.t. } \mathbf{y} = \mathbf{B}\mathbf{c}$$

and

$$P'_{l_q/l_1} \min \sum_{i=1}^n \|B[i]c[i]\|_q \text{ s.t. } \mathbf{y} = \mathbf{B}\mathbf{c}$$

for $q \geq 1$

- Does not restrict the blocks of a dictionary to have linearly independent columns
- No restriction on the length of the blocks
- Uniqueness of blocksparse representation