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Beamforming Design for Full-Duplex Two-Way Amplify-and-Forward MIMO Relay

(Yeonggyu Shim, Wan Choi, Hyuncheol Park)

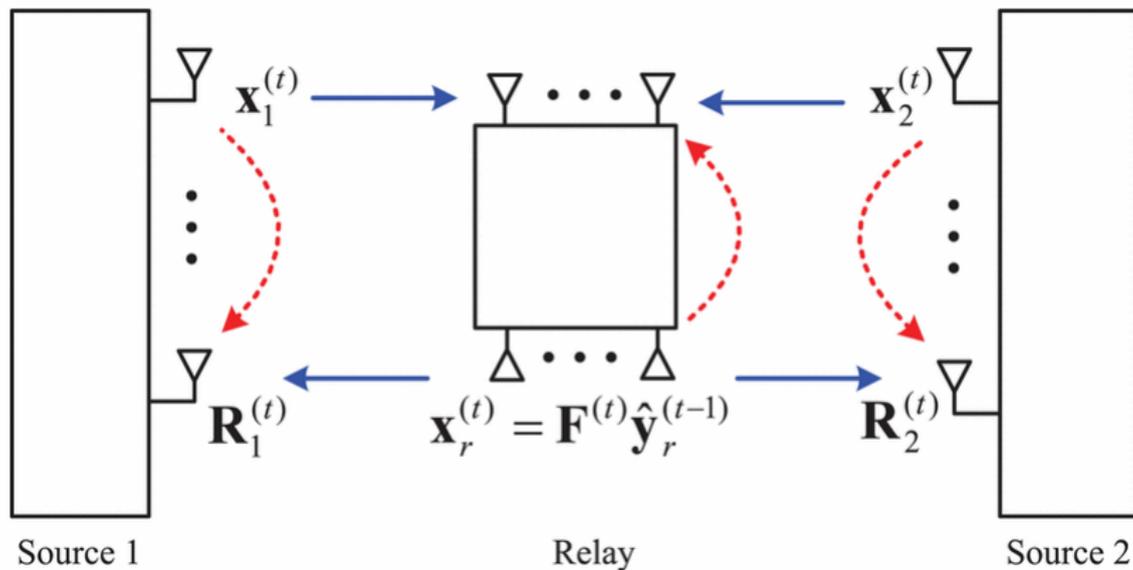
▶ Contributions

- ▶ Optimized Relay Beamforming matrix
- ▶ Optimized Receive Beamforming matrices at sources

▶ System Model

- ▶ Consider a two-way FD AF relaying system consisting of two source nodes s_1 and s_2 with N_s antennas, and one relay node r with N_r antennas, where all nodes operate in the FD mode.
- ▶ Channel links are modelled as Frequency-flat fading channels and assumed to be static in Time slot.
- ▶ CSI between two nodes is assumed to be perfectly known.
- ▶ CSI of the loopback channels is assumed to be imperfect.
- ▶ focuses on loopback Self Interference.

System



Received signals with imperfect loopback SI cancellation:

- ▶ At relay in time slot $t - 1 (t \geq 2)$ is given by,

$$\begin{aligned} \hat{\mathbf{y}}_r^{(t-1)} = & \mathbf{H}_{1,r}^{(t-1)} \mathbf{x}_1^{(t-1)} + \mathbf{H}_{2,r}^{(t-1)} \mathbf{x}_2^{(t-1)} + \mathbf{n}_r^{(t-1)} \\ & + \sum_{i=0}^{t-2} \left\{ \prod_{j=1}^{t-1-i} (\Delta_{r,r}^{(t-j)} \mathbf{F}^{(t-j)}) \right. \\ & \left. \times (\mathbf{H}_{1,r}^{(i)} \mathbf{x}_1^{(i)} + \mathbf{H}_{2,r}^{(i)} \mathbf{x}_2^{(i)} + \mathbf{n}_r^{(i)}) \right\} \end{aligned} \quad (1)$$

- ▶ At source l is,

$$\begin{aligned} \hat{\mathbf{y}}_l^{(t)} = & \mathbf{H}_{r,l}^{(t)} \mathbf{F}^{(t)} \mathbf{H}_{l,r}^{(t-1)} \mathbf{x}_l^{(t-1)} + \mathbf{H}_{r,l}^{(t)} \mathbf{F}^{(t)} \\ & \sum_{i=0}^{t-2} \left\{ \prod_{j=1}^{t-1-i} (\Delta_{r,r}^{(t-j)} \mathbf{F}^{(t-j)}) (\mathbf{H}_{l,r}^{(i)} \mathbf{x}_l^{(i)} + \mathbf{H}_{l,r}^{(i)} \mathbf{x}_l^{(i)} + \mathbf{n}_r^{(i)}) \right\} \\ & + \mathbf{H}_{r,l}^{(t)} \mathbf{F}^{(t)} \mathbf{n}_r^{(t-1)} + \Delta_{l,l}^{(t)} \mathbf{x}_l^{(t)} + \mathbf{n}_l^{(t)}. \end{aligned} \quad (2)$$

$\mathbf{F}^{(t)}$, $\mathbf{R}_l^{(t)}$ are the relay Beamforming matrix and receive Beamforming matrix respectively.

▶ **MMSE Based Beamforming design**

- ▶ Proposes an iterative algorithm which decouples the primal problem into two subproblems and solve them alternately; one is for relay beamforming design and the other is for receive beamforming design at sources.

▶ **Results:**

- ▶ Closed form expressions for Relay Beamforming and Receive Beamforming matrices are derived.
- ▶ An Iterative Algorithm is proposed for joint Beamforming Design.

Resource Allocation for D2D Communication Underlaid Cellular Networks Using Graph-Based Approach

(Tuong Duc Hoang, Long Bao Le, Tho Le-Ngoc)

► Contributions

focuses on the radio resource allocation for D2D communications in cellular networks for the first scenario and the developed algorithm for scenario I is employed to tackle the resource allocation for scenario II

- Novel Iterative Rounding algorithm is proposed solve the subband assignment problem based on the combination of linear programming and efficient rounding techniques. Specifically, each iteration solves a relaxed version of the subband assignment problem for unallocated subbands and network links.
- *Scenario I*: Each active (admitted) D2D link is assigned one subband and each subband is exploited by at most one D2D link.
- *Scenario II*: Each active (admitted) D2D link can be assigned multiple subbands and each subband is exploited by at most one D2D link.

- ▶ *Scenario III*: Each active (admitted) D2D link can be assigned multiple subbands and each subband can be exploited by multiple D2D links.
- ▶ **System model**:
 - ▶ The spectrum sharing problem among multiple D2D and cellular links in the uplink direction.
 - ▶ Let $\mathcal{N} = \{1, 2, \dots, N\}$ with size $|\mathcal{N}| = N$ be the set of subbands in the system.
 - ▶ We denote $\mathcal{K}_c = \{1, 2, \dots, K_c\}$ as the set of cellular links, $\mathcal{K}_d = \{K_c + 1, \dots, K_c + K_d\}$ as the set of D2D links, and $\mathcal{K} = \mathcal{K}_c \cup \mathcal{K}_d$ as the set of all communications links with size $|\mathcal{K}| = K_c + K_d = K$.
 - ▶ Assumption: Each subband can be allocated to at most one cellular and one D2D link.

- ▶ The signal to interference plus noise ratio (SINR) achieved by link $k \in \mathcal{K}$ on subband n can be expressed as

$$\Gamma_k^{[n]}(\mathbf{p}^{[n]}, \boldsymbol{\rho}^{[n]}) = \frac{\rho_k^{[n]} p_k^{[n]} h_{kk}^{[n]}}{\sigma_k^{[n]} + \sum_{l \in \mathcal{K} \setminus \mathcal{K}_k} \rho_l^{[n]} p_l^{[n]} h_{kl}^{[n]}}, \quad (3)$$

- ▶ The achievable rates of link $k \in \mathcal{K}$ on subband n and all the subbands can be expressed as

$$r_k^{[n]}(\mathbf{p}^{[n]}, \boldsymbol{\rho}^{[n]}) = \log_2 \left(1 + \Gamma_k^{[n]}(\mathbf{p}^{[n]}, \boldsymbol{\rho}^{[n]}) \right),$$
$$r_k(\mathbf{p}, \boldsymbol{\rho}) = \sum_{n \in \mathcal{N}} r_k^{[n]}(\mathbf{p}^{[n]}, \boldsymbol{\rho}^{[n]}),$$

- ▶ The considered resource allocation problem can now be formulated as

$$\max_{\mathbf{p}, \rho} \mathcal{R} = \sum_{k \in \mathcal{X}_c} \alpha r_k(\mathbf{p}, \rho) + \sum_{k \in \mathcal{X}_d} (1 - \alpha) r_k(\mathbf{p}, \rho)$$

- ▶ Solution:
 - ▶ we first characterize the optimal power allocation solution for a given subband assignment.
 - ▶ Based on this result, we formulate the subband assignment problem by using the graph-based approach, in which each link corresponds to a vertex and each subband assignment is represented by a hyper-edge.
 - ▶ We then propose an iterative rounding algorithm and an optimal branch-and-bound (BnB) algorithm to solve the resulting graph-based problem.

Secure Transmission in Cooperative Relaying Networks With Multiple Antennas

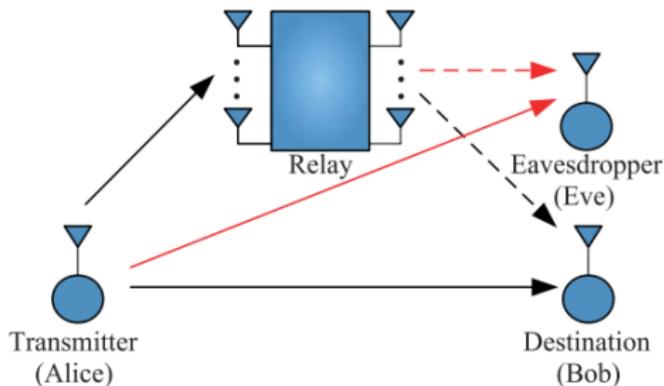
*(Yuzhen Huang; Jinlong Wang; Caijun Zhong; Trung Q. Duong;
George K. Karagiannidis)*

► Contributions

- For ZF/MRC and ZF/SC, we present novel closed-form lower and upper bounds for the secrecy outage probability and the probability of non-zero secrecy capacity, respectively, as well as a simple high SNR secrecy outage analysis.
- For MRT/MRC and MRT/SC, closed-form approximations for the secrecy outage probability and the probability of non-zero secrecy capacity are provided, respectively.
- For the CJ/ZF scheme, new exact closed-form expressions for the secrecy outage probability and the probability of non-zero secrecy capacity are derived.

- ▶ The analytical results suggest that the ZF/MRC (MRT/MRC) scheme always achieves better performance than that of the corresponding ZF/SC (MRT/SC) scheme. In addition, the ZF/MRC (ZF/SC) scheme outperforms the corresponding MRT/MRC (MRT/SC) scheme in the low SNR regime, while in the high SNR regime, the MRT/MRC (MRT/SC) scheme attains better secrecy performance than the corresponding ZF/MRC (ZF/SC) scheme.

System model: We consider a dual-hop multiple antenna AF relaying network, where both Alice (A), Bob (B), and Eve (E) are equipped with a single antenna, while the relay (R) is equipped with M antennas.



SNR of $A \rightarrow R \rightarrow B$ link is given by,

$$\gamma_{ARB} = \frac{P_s \left| \mathbf{h}_{RB}^\dagger \mathbf{W} \mathbf{h}_{AR} \right|^2}{\sigma^2 \left(1 + \left\| \mathbf{h}_{RB}^\dagger \mathbf{W} \right\|_F^2 \right)}. \quad (4)$$

SNR of $A \rightarrow R \rightarrow E$ link can be derived as,

$$\gamma_{ARE} = \frac{P_s \left| \mathbf{h}_{RE}^\dagger \mathbf{W} \mathbf{h}_{AR} \right|^2}{\sigma^2 \left(1 + \left\| \mathbf{h}_{RE}^\dagger \mathbf{W} \right\|_F^2 \right)}. \quad (5)$$

instantaneous SNRs of the main channel under MRC,

$$\gamma_{BMRC} = \gamma_{AB} + \gamma_{ARB}$$

$$\gamma_{EMRC} = \gamma_{AE} + \gamma_{ARE}.$$

The achievable secrecy capacity of the relaying wiretap channels is defined as,

$$C_S \triangleq \frac{1}{2} [\log_2 (1 + \gamma_{B_i}) - \log (1 + \gamma_{E_{MRC}})]^+,$$

The SNR at B and E is maximized using different schemes.