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# Block-Sparse Impulsive Noise Reduction in OFDM Systems—A Novel Iterative Bayesian Approach.

—M. Korki, J. Zhang, C. Zhang, and H. Zayyani

- *Problem:* Mitigation of the impact of impulsive noise in OFDM systems.
- *Contribution:* Novel block-IBA Rx for optimal estimation & removal of impulsive noise.
- *Model:*  $\mathbf{y} = \mathbf{FHF}^H \mathbf{x} + \mathbf{F}\mathbf{e} + \mathbf{Fz} = \mathbf{D}\mathbf{x} + \mathbf{F}\mathbf{e} + \tilde{\mathbf{z}}$ .  
 $\mathbf{e} = \text{diag}\{\mathbf{s}\}\boldsymbol{\theta}$ ,  $\mathbf{s}$ : Order-1 MP  $p_{01}, p_{10}$ ;  $p(e_i) = p\delta(e_i) + (1 - p)\mathcal{CN}(0, \sigma_\theta^2)$  [BGHMM].
- *Impulsive noise Rx:*  
 $\mathbf{y}_J = \mathbf{F}_J \mathbf{e} + \tilde{\mathbf{z}}_J$ ,  
 Estimation of  $\mathbf{e}$   $\begin{cases} \text{MAP estimation of } \mathbf{s}: \mathbf{s}_{\text{MAP}} = \arg \max_{\mathbf{s}} p(\mathbf{s})p(\mathbf{y}_J|\mathbf{s}) \\ \text{MAP estimation of } \boldsymbol{\theta} \text{ using SBL: Gaussian priors on } \boldsymbol{\theta}. \end{cases}$
- *Block IBA noise estimation:*  
 Estimation of  $\mathbf{s}$ : Search over  $2^N$  sets- convert the problem.  
 Model  $\mathbf{s}$  as GM with two GVs centered around 0 and 1.  
 Algo.: **E-step:** Estimation of  $\boldsymbol{\theta}$  based on  $\hat{\mathbf{s}}$ .  
**M-step:**  $\mathbf{s}_{\text{MAP}} = \arg \max_{\mathbf{s}} \underbrace{(\log p(\mathbf{s}) + \log p(\mathbf{y}_J|\mathbf{s}))}_{L(\mathbf{s})}$   
 Maximize  $L(\mathbf{s})$  with steepest ascent.
- Learning model parameters  $p_{01}, p_{10}, \sigma_\theta^2, \sigma_n^2$  with method of moments.

## Message Passing Algorithms for Phase Noise Tracking Using Tikhonov Mixtures.

-Shachar Shayovitz and Dan Raphaeli

- Problem:** Phase noise can limit the information rate of comm. systems, PLL unsuitable for coded systems in high phase noise.
- Contribution:** Approach for approximating phase noise f/w and b/w messages - Tikhonov mixtures; Hypothesis expansion & clustering; Limiting instantaneous algo. complexity.
- Model:**  $r_k = c_k e^{j\theta_k} + n_k$ ,  $\theta_k = \theta_{k-1} + \Delta_k$ .  
 Sum & Product messages computed:  $p_f(\theta_k) = \int_0^{2\pi} p_f(\theta_{k-1}) \cdots$ ;  $p_b(\theta_k) = \int_0^{2\pi} p_b(\theta_{k-1}) \cdots$
- Background used:** Directional statistics - RVs defined on circles & spheres.  
 Circular mean & variance:  $\mu_C = \angle \mathbb{E}[e^{j\theta}]$ ,  $\sigma_C^2 = \mathbb{E}[1 - \cos(\theta - \mu_C)]$ ; One CD: Tikhonov.  
 CMVM: In KL-divergence sense, the nearest Tikhonov dist. to any circular dist.  $f(\theta)$  has its circular mean & var. matched to  $f(\theta)$ .
- Tikhonov mixture:**  $p_f(\theta_k) \approx \sum_{i=1}^{N_f^k} \alpha_i^{k,f} t_i^{k,f}(\theta_k)$ ,  $t_i^{k,f}(\theta_k) = \frac{e^{\Re[z_i^{k,f} e^{-j\theta_k}]}}{2\pi I_0(|z_i^{k,f}|)}$ .
- Mixture reduction:** Given a Tikhonov mixture  $f(\theta)$  of order  $L$ , find a Tikhonov mixture  $g(\theta)$  of order  $N (< L)$  which minimizes  $D_{\text{KL}}(f(\theta) \| g(\theta))$ .
- Mixture reduction algo.:** Global and local algorithms; *Runnalls, Lehmann, West*.  
**Adaptive mixtures:** Limited # mixture components, use pilot symbols to regain tracking.

# Training Sequence Design for Feedback Assisted Hybrid Beamforming in Massive MIMO Systems.

~S. Noh, M. D. Zoltowski, and D. J. Love

- Problem:** Channel estimation in massive MIMO FDD systems is challenging; schemes are highly complex, training signal overhead is huge.
- Contribution:** Training scheme design with reduced dimensionality of training sequence & transmit precoding; suboptimal algo. to minimize max. steady state channel MSE.
- Model:**  $y_k = \mathbf{h}_l^H \mathbf{s}_k + w_k$ ; block Tx with  $M = M_p + M_d$  symbols;  $\mathbf{s}_k = \mathbf{v}_l x_k$ .  
 Block fading channel:  $\mathbf{h}_{l+1} = a\mathbf{h}_l + \sqrt{1-a^2} \mathbf{b}_{l+1}$ ;  $\mathbf{b}_l \sim \mathcal{CN}(\mathbf{0}, \mathbf{R}_h)$ ,  $\mathbf{R}_h = \mathbf{U}\mathbf{\Lambda}\mathbf{U}^H$ ;  
 Optimal channel estimation: Kalman filtering.  
 Goal: Design  $\mathbf{F}$  that supports a spatial matched filter Tx beamforming.  
 Prop.: Use of dominant eigenvectors of Kalman prediction error cov. matrix as training signals minimize channel MSE.
- Proposed scheme:** Channel estimate: Lin. comb. of all used training symbols - should be designed to capture  $n_d$  dominant channel eig. modes.  
 Eigenvectors of  $\mathbf{R}_h$  selected as the training signals.  
 Min. and max. steady-state channel MSE derived - using Riccati equation.  
 MMSE upper bound minimization; Max. steady-state channel MSE is reduced by transmitting the training signal more frequently.  
 Hybrid analog-digital b/f:  $\mathbf{R}_h \approx \tilde{\mathbf{F}}\mathbf{\Lambda}\tilde{\mathbf{F}}^H$ ; DFT-based training signal.  
 Beamformer  $\mathbf{F}$  will span subspace of training signal - matrix determined by distinct DFT columns.
- Extension to multiuser massive MIMO.

## Permutation Trellis Coded Multi-Level FSK Signaling to Mitigate Primary User

## Interference in Cognitive Radio Networks. –R. El-Bardan, E. Masazade, O. Ozdemir, Y. S. Han, P. K. Varshney

- *Problem:* Inefficient usage of assigned frequency bands; performance degradation for SUs due to interference by PUs.
- *Contribution:* Use of PTC-based multi-level FSK signaling (no overhead); analytical BER and throughput; SU transmissions robust and resilient.
- *Model:*  $x_k(t) = s_k^r(t) + \sum_j i_j^r(t) + w(t)$ , [ $i_j = 0$  if PU is absent].  
For each convolution coded o/p symbols, PTC encoder uses a code matrix.  $\mathbf{T}_i$ ;  $H$ -FSK  
SUs transmit concurrently with PUs; interference from PUs independent for each SU bit.
- *BER analysis:* Through special properties of Viterbi decoder.  
Hamming distance b/w any two PTC matrices is lower-bounded by the Hamming distance b/w the convolution coded o/p symbols.  
$$P_e \leq \sum_{d=d_{\text{free}}}^{\infty} a_d P(d) \approx \sum_{d=d_{\text{free}}}^{d_{\text{free}}+z} a_d P(d).$$
- *Throughput analysis:* Average throughput of the SU session in the presence of multiple PUs (activities are modeled as a 2- state Markov chain).  
SU Tx has no information regarding the presence or absence of PUs in the network.  
$$T_e = (1 - \text{PER}) \times R_p \times L \approx (1 - \hat{P}_e)^L \times R_p \times L.$$

## Other interesting papers

- Omnidirectional Precoding Based Transmission in Massive MIMO Systems.  
- X. Meng, X. Gao, and X.-G. Xia
- Power Allocation for Distributed Antenna Systems in Frequency-Selective Fading Channels.  
- Q.-Y. Yu, Y.-T. Li, W. Xiang, W.-X. Meng, and W.-Y. Tang
- Adaptive Transmission Rate With a Fixed Threshold Decoder for Diffusion-Based Molecular Communication.  
- M. Movahednasab, M. Soleimanifar, A. Gohari, M. Nasiri-Kenari, and U. Mitra
- A Sparse Recovery Method for Initial Uplink Synchronization in OFDMA Systems.  
- M. M. Hyder and K. Mahata
- Full-Duplex Decode-and-Forward Relay-Assisted Interference Management: A Diversity Gain Region Perspective.  
- R. P. Sirigina and A. S. Madhukumar