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Goal:

• To propose a new hybrid precoding technique for massive MIMO systems using spatical channel covariance matrices

System Model:

• Base station with N antennas, M RF chains. U users with single antenna

 $\mathbf{y} = \mathbf{H}^* \mathbf{F}_{RF} \mathbf{F}_{BB} \mathbf{P} \mathbf{s} + \mathbf{n}$

Contributions:

- Regularized zero forcing (RZF) precoding for digital precoding
- Unconstrained analog precoder design using only the users' spatial channel covariance matrices
- Divide the unconstrained analog precoder into two separate matrices:
 - Constrained analog precoding matrix
 - Additional baseband precoding matrix
- Analysis of the signal-to-leakage-plus-noise ratio (SLNR) loss caused by the hybrid precoding

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- Hybrid Precoding using Spatial Channel Covariance Matrices:
 - RZF for digital precoding:

$$\mathbf{F}_{BB} = (\mathbf{F}_{RF}^* \mathbf{H} \mathbf{H}^* \mathbf{F}_{RF} + \beta \mathbf{I})^{-1} \mathbf{F}_{RF}^* \mathbf{H}$$

- Goal is to find F_{RF} to maximize the SLNR. F_{RF} is decomposed into product of two semi-unitary matrices.
- As $N \to \infty$, SLNR converges to a deterministic value which depends only on the spatial channel covariance
- Random matrix theory tools are used to obtain the optimal analog precoding matrix without phase shifter constraint
- Hybrid Precoding under Phase Shifter constraint:

$$\min_{\mathbf{F}_{RF,C}, \left[\left[\mathbf{F}_{RF,C} \right]_{i,j} \right| = \frac{1}{\sqrt{N}}, \mathbf{A}} \left\| \mathbf{F}_{RF,UC} \mathbf{A} - \mathbf{F}_{RF,C} \right\|_{F}^{2}$$

- A introduced to increase the degrees of freedom in the design
- Alternating minimization used to solve for F_{RF,C}, A
- For maintaining the orthogonality of the unconstrained solution, a compensation matrix is introduced in the baseband
- RZF digital precoder designed after the precompensation matrix design
- Asymptotic analysis for the SLNR loss caused by the hybrid structure

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Fast Approximation Algorithms for a Class of Non-convex QCQP Problems Using First-Order Methods

Goal:

• To solve a special class of non-convex QCQP optimization problems:

$$\max_{\mathbf{x}\in\mathbb{C}^{N}} \min_{m\in\mathcal{M}} \mathbf{x}^{H} \mathbf{A}_{m} \mathbf{x}$$
s.t. $\mathbf{x}\in\mathcal{F}$ (1)

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where $\mathcal{M} = \{1, 2, \dots, M\}, \mathbf{A}_m \in \mathbb{C}^{N \times N} \succeq \mathbf{0}, \mathcal{F} \subseteq \mathbb{C}^N$ is a compact, convex set

Contributions:

- Proposed low complexity, high performance SCA algorithms for solving (1)
 - Fast, first order methods (FOMs) for solving each non-smooth, convex subproblem
 - Subproblem expressed as the maximization of a bilinear function over a convex set
 - Nesterov smoothing technique, Nemirovski's saddle point reformulation approach to develop FOMs to solve each SCA subproblem
 - ADMM approach
- Convergence analysis using two different techniques
 - KKT conditions of the local smooth convex problem
 - Establish the link between the first order properties of the objective and its non-smooth, convex surrogate (first time in the literature)

- SCA: Find a lower (upper) bound and maximize (minimize) the convex surrogate function iteratively
- First order Taylor series used to find a convex, non-smooth upper bound for the objective function of the reformulated problem
- Existing Approaches:
 - Eprigraph representation: High computational complexity
 - Projected subgradient methods: Slow convergence rate
- Problem structure utilized to express the problem as

$$egin{aligned} \min_{\mathbf{ ilde{x}}\in\mathbb{R}^{2N}}&\max_{\mathbf{ ilde{y}}\in\Delta_M}\phi^{(n)}\left(\mathbf{ ilde{x}},\mathbf{ ilde{y}}
ight)\ ext{s.t.}&\mathbf{ ilde{x}}\in\mathcal{ ilde{F}} \end{aligned}$$

- New approach using
 - Smoothing via conjugation Nesterov's smoothing technique
 - Convex-concave saddle point reformulation Nemirovski's formulation
 - ADMM

Goal:

• To propose a Tx precoding scheme for the CR Z-channel to improve the QoS of SUs System Model:

- Single cell CR Z-channel system, 1 N-antenna SBS, K single antenna SUs, L single antenna PUs
- Rx signal at the *i*th SU:

$$y_{i} = \mathbf{h}_{i}^{T} \mathbf{x} + n_{i} = \underbrace{\mathbf{h}_{i}^{T} \mathbf{w}_{i} b_{i}}_{\text{desired signal}} + \underbrace{\sum_{j=1, j \neq i}^{K} \mathbf{h}_{i}^{T} \mathbf{w}_{j} b_{j} + n_{i}}_{\text{interference plus noise}}$$

Contributions:

- Proposed a precoder design to minimize the worst case SU's SEP (WSUSEP)
- Derived conditions to reformulate the problem to a convex problem
- Derived a computationally efficient approximation technique that achieves close to optimal performance

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Conventional SINR based CR DL Beamforming design: Max-Min fair problem:

- Maximize the min SINR subj. to average interference and total Tx power constraints
- Problem reformulated to a quasi-convex optimization problem and solved using sequential SOCP

WSUSEP Approach:

- Precoder design to constructively utilize the interfering signal to enhance the desired signal
- To design the precoder to steer the rx signals of SUs into the corresponding decision regions to reduce the SEP

$$\begin{split} \min_{\mathbf{x},\rho} & \rho \\ \text{s.t.} & \Pr\left(\psi_i\left(\mathbf{x}, n_i\right) \in \mathcal{A}_{\theta}^{2\pi-\theta}\right) \leq \rho, i = 1, \dots, K, \\ & \|\mathbf{x}\|^2 \leq P, \left|\mathbf{g}_I^T \mathbf{x}\right|^2 \leq \epsilon_I, I = 1, \dots, L \end{aligned}$$

$$\end{split}$$

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- The above problem reformulated in which the constraint depends on the joint normal CDF of the noise
- Solved using barrier method

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Goal:

• To propose a successive pseudoconvex approximation algorithm to efficiently compute stationary points for a large class of nonconvex optimization problems given below:

$$\min_{\mathbf{x}} \quad f(\mathbf{x}) + g(\mathbf{x}) \tag{3}$$
s.t. $x \in \mathcal{X}$

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where f is differentiable (convex or non convex), and g is non-differentiable and convex

Contributions:

- Developed a novel iterative pseudoconvex method to solve problem (3)
- Proposed new exact/successive line search methods to determine the step size when the objective is non-differetiable
- Applications:
 - Sum rate maximization of MIMO broadcast channel
 - Energy efficiency maximization of a massive MIMO system
 - LASSO (to illustrate the advantage of the proposed step size)

Proposed Successive Pesudoconvex Approximation Algorithm:

- In iteration t, let $\tilde{f}(\mathbf{x}; \mathbf{x}^{t})$ be the approximate function of $f(\mathbf{x})$ around \mathbf{x}^{t}
- Assumptions on *f* (x; x^t): pseudoconvex, continuously differentiable, and gradient of *f* and *f* are identical at x^t
- Two more assumptions on the sequence {x^t} ensures that any limit point of {x^t} is a stationary point of the original problem
- Nondifferentiable optimization problems: Reformulate using auxiliary variables

$$\min_{\mathbf{x}} \quad f(\mathbf{x}) + y$$
s.t. $x \in \mathcal{X}, g(\mathbf{x}) \le y$

$$(4)$$

- Assumption: $\tilde{f}(\mathbf{x}, \mathbf{y})$ is convex
- g replaced by auxiliary variable
- Convexity of g and Jensen's inequality used to replace the auxiliary variables (Pg. 3317, Step 2 in the paper for more details)
- Converges to a stationary point of f

July 1, 2017

- **Q** Projected Nesterov's Proximal-Gradient Algorithm for Sparse Signal Recovery
- **Q** Optimal Training Sequences for Large-Scale MIMO-OFDM Systems
- Overview Article) Tensor Decomposition for Signal Processing and Machine Learning

July 15, 2017

- Joint Sensing Matrix and Sparsifying Dictionary Optimization for Tensor Compressive Sensing
- **②** Sparsity-Driven Laplacian-Regularized Outlier Identification for Dictionary Learning
- An Information Theoretic Approach to Robust Constrained Code Design for MIMO Radars
- An Efficient Global Algorithm for Single-Group Multicast Beamforming
- Maximum-Likelihood Detection for MIMO Systems Based on Differential Metrics