## I. Approximate Set Identification: PAC Analysis for Group Testing II. SSR for OFDM channel estimation: Implementation in SDR

#### Monika Bansal Advisor: Dr. Chandra R Murthy

Department of ECE, IISc

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#### Overview

Approximate Set Identification: PAC Analysis for Group Testing

- Group Testing Framework
- Function Learning Model
- PAC Analysis: Approximate Set Identification
  - PAC type bound for CoMa
  - PAC type bound for DD

2 SSR for OFDM channel estimation: Implementation in SDR

- Sparsity in Channel
- SBL Framework
- Implementation in GNU Radio
- Experiment Results

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#### Group Testing

A set of N items with k defective items ( $k \ll N$ ).

- Group test: A group of items is tested in a group test
  - Test outcome 1 indicates presence of defective item(s)
  - Outcome 0 indicates all items are non-defective in the test

#### • Main issues:

- Sample Complexity
- Correctness of solution
- Pooling Design

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# Item Set 0 1 0 0 1 0





#### **Test Matrix**

Figure: A toy example for non-adaptive group testing.

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#### Group Testing Framework

## Group Testing Model



 $\mathbf{y} \in \{0,1\}^M$  is the binary test outcome vector  $\mathbf{a}_{c_j} \in \{0,1\}^M$  is the  $j^{th}$  column of  $\mathbf{A}$  $\mathbf{a}_{r_i} \in \{0,1\}^n$  is the  $i^{th}$  row of  $\mathbf{A}$  $\mathbf{A}(i,j) \sim \mathcal{B}(p)$  i.i.d.  $\mathbf{x} \in \{0,1\}^n$  is test item vector

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## Learning Problem

Learn unknown target function  $f(\cdot) \in \mathcal{C}$ 

Available items to the learner:

- Random examples: **a**<sub>i</sub>
- Corresponding label: y<sub>i</sub>

$$\begin{aligned} \mathbf{a}_i &\in \{0,1\}^n\\ y_i &= f(\mathbf{a}_i) \in \{0,1\}\\ f: \{0,1\}^n \to \{0,1\} \text{ is some boolean function} \end{aligned}$$

How many examples do we need to output an hypothesis  $f^*$  s.t. maximum error is  $\epsilon$  with confidence  $1 - \delta$ ?

## Figure Revisited





#### **Test Matrix**

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#### Group Testing as a Function Learning Problem

	Group testing	Learning model
Target function	x	$f(\cdot)$
Random example	a <sub>ri</sub>	a <sub>i</sub>
Label	<b>y</b> ( <i>i</i> )	Уі
Distribution	$\mathcal{B}(p)$	$\mathcal{D}$
Output hypothesis	Ŷ	$f^*$

Goal: Analyze group testing recovery algorithms using PAC (Probably Approximately Correct) framework applied to function learning problems.

### **PAC Analysis**

A learning algorithm is said to be PAC-learn C with approximation parameter  $\epsilon$  and confidence parameter  $\delta$  if  $\forall$  distributions D and all target functions  $f \in C$ , the algorithm draws M samples, runs for time at most tand outputs a function  $f^*$  s.t.

$$e(f^*, f) = Pr_{\mathbf{a} \sim \mathcal{D}} (f^*(\mathbf{a}) \neq f(\mathbf{a}))$$
$$Pr (e(f^*, f) > \epsilon) \le \delta$$

With prob.  $1 - \delta$  the output hypothesis  $f^*$  will make at most  $\epsilon$  error.

## Defective Set Recovery



- Column Matching (CoMa)<sup>1</sup> Algorithm
- Definite Defective (DD)<sup>2</sup> Algorithm

<sup>1</sup>Non-adaptive Group Testing: Explicit Bounds and Novel Algorithms <sup>2</sup>Group Testing Algorithms: Bounds and Simulations

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#### $\epsilon$ Approximate Set Identification

#### Allowed hidden non-defective items

$$egin{aligned} \mathbb{P}_{\mathbf{a}_i \sim \mathcal{B}}\left(\hat{\mathbf{x}}(\mathbf{a}_i) 
eq \mathbf{x}(\mathbf{a}_i)
ight) &= (1 - (1 - p)^G)(1 - p)^k \leq \epsilon \ g_\epsilon &= \left\lfloor rac{\ln\left(1 - \epsilon/\left(1 - p
ight)^k
ight)}{\ln\left(1 - p
ight)}
ight
floor \ \mathbb{P}(G \leq g_\epsilon) \geq 1 - \delta \end{aligned}$$

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#### Allowed unidentified defective items

$$\mathbb{P}_{\mathbf{a}_i \sim \mathcal{B}}\left(\hat{\mathbf{x}}(\mathbf{a}_i) \neq \mathbf{x}(\mathbf{a}_i)\right) = (1 - (1 - p)^D)(1 - p)^{k - D} \le \epsilon$$
 $d_{\epsilon} = \left\lfloor rac{\ln(1 + \epsilon/(1 - p)^k)}{\ln(1/(1 - p))} 
ight
floor$ 
 $\mathbb{P}(D \le d_{\epsilon}) \ge 1 - \delta$ 

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## PAC type bound for CoMa

#### Theorem 1

The sufficient number of tests such that estimated set using CoMa does not agree with the true defective set on the future group tests with probability at most  $\epsilon$  with confidence parameter  $1 - \delta$  is given as,

$$M_{g_{\epsilon}} = \frac{\log \binom{n-k}{g_{\epsilon}+1} + \log \frac{1}{\delta}}{\log \left(1/(1-(1-p)^k + (1-p)^{g_{\epsilon}+k+1})\right)},$$
$$\mathbb{P}(e(\hat{\mathbf{x}}, \mathbf{x}) > \epsilon) = \mathbb{P}(G \ge g_{\epsilon}) \le \binom{n-k}{g_{\epsilon}+1} P^h_{g_{\epsilon}+1}(M)$$
where, 
$$P^h_{g_{\epsilon}+1}(M) = \left(1-(1-p)^k + (1-p)^{g_{\epsilon}+1+k}\right)^M$$

# Bound on success probabilities



#### Bound on number of tests in PAC setting



## PAC type bound for DD

#### Theorem 2

The sufficiency bound on the number of tests such that estimated set using DD do not agree with the true defective set on the future group tests with probability at most  $\epsilon$  with confidence parameter  $1-\delta$  is given as,

$$egin{split} {k \ d_\epsilon+1} (1-(d_\epsilon+1)p(1-p)^{k-1+ar{g}})^M &\leq \delta, \ \mathbb{P}(e(\hat{\mathbf{x}},\mathbf{x})>\epsilon/G=g) &\leq {k \ d_\epsilon+1} (1-(d_\epsilon+1)p(1-p)^{k-1}(1-p)^g)^M, \ \mathbb{P}(e(\hat{\mathbf{x}},\mathbf{x})>\epsilon) &\leq {k \ d_\epsilon+1} (1-(d_\epsilon+1)p(1-p)^{k-1}(1-p)^{ar{g}})^M. \end{split}$$

where  $\bar{g} = (N - k)(1 - p(1 - p)^k)^M$  and  $\tilde{g}$  is a tuning parameter which depends on  $d_{\epsilon}$ .

#### Bound on success probabilities



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#### Bound on number of tests in PAC setting



#### Conclusions

- PAC analysis resulted in approximate set identification analysis
- Full defective set can always be recovered from approximate set
- Two stage procedure has more flexibility
- Once we identify a big number of non-defective items, random pooling does not give much further information

#### Wireless Communication Channel



#### Sparse in lag domain

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## **OFDM Channel Model**



 $\mathbf{y} = \mathbf{XFh} + \mathbf{v}$ 

 $\mathbf{y} \in \mathbb{C}^{N \times 1}$  is the received vector after FFT  $\mathbf{X} \in \mathbb{C}^{N \times N}$  contains data symbol and pilot symbols along the diagonal  $\mathbf{F} \in \mathbb{C}^{N \times L} (N > L)$  contains the first *L* columns of  $N \times N$  DFT matrix  $\mathbf{v} \in \mathbb{C}^{N \times 1} \sim C\mathcal{N}(0, \sigma^2 I)$  is the AWGN noise  $\mathbf{h} \in \mathbb{C}^{L \times 1}$  is the time domain channel response

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#### Channel Model using *Pilots* only:



$$\begin{aligned} \mathbf{y}_{P} &= \mathbf{X}_{p} \mathbf{F}_{p} \mathbf{h} + \mathbf{v}_{p}, \quad (P < L) \\ &= \phi_{p} \mathbf{h} + \mathbf{v}_{p} \end{aligned}$$

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#### SBL Framework

$$\mathbf{h} \sim \mathcal{CN}(\mathbf{0}, \mathbf{\Gamma}), \quad \mathbf{\Gamma} = \mathsf{diag}(\gamma(1), ..., \gamma(L))$$

#### SBL estimation problem:

$$\begin{split} \hat{\mathbf{h}} &= \arg \max_{\mathbf{h}, \gamma \in \mathbb{R}_{+}^{L \times 1}} p(\mathbf{y}_{\rho} | \mathbf{h}; \gamma) p(\mathbf{h}; \gamma) \\ &= \arg \min_{\mathbf{h}, \gamma \in \mathbb{R}_{+}^{L \times 1}} \frac{\|\mathbf{y}_{\rho} - \mathbf{X}_{\rho} \mathbf{F}_{\rho} \mathbf{h}\|_{2}^{2}}{\sigma^{2}} + \log |\mathbf{\Gamma}| + \mathbf{h}^{H} \mathbf{\Gamma}^{-1} \mathbf{h} \end{split}$$

Instead of estimating  ${\bf h}$  directly, we first estimate  $\gamma$  using type II ML estimate as given below

$$\hat{\gamma}_{\textit{ML}} = \operatorname*{arg\ max}_{\gamma \in \mathbb{R}_{+}^{L \times 1}} p(\mathbf{y}_{\textit{p}}; \gamma)$$

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## EM algorithm

$$p(\mathbf{h};\gamma) = \prod_{i=1}^{L} (\pi\gamma(i))^{-1} \exp\left(-\frac{|h(i)|^2}{\gamma(i)}\right)$$

$$\begin{split} \mathbf{E}\text{-step:} \ & Q\left(\gamma|\gamma^{(r)}\right) = \mathbb{E}_{\mathbf{h}|\mathbf{y}_{p};\gamma^{(r)}}\left[\log \ p(\mathbf{y}_{p},\mathbf{h};\gamma)\right] \\ \mathbf{M}\text{-step:} \ & \gamma^{(r+1)} = \arg\max_{\gamma \in \mathbb{R}_{+}^{L \times 1}} Q\left(\gamma|\gamma^{(r)}\right) \\ & \gamma^{(r+1)}(i) = \Sigma(i,i) + |\mu(i)|^{2} \end{split}$$

**Probability densities:** 

$$p\left(\mathbf{h}|\mathbf{y}_{p};\gamma^{(r)}\right) = \mathcal{CN}(\mu,\Sigma)$$
  
$$\Sigma = \Gamma^{(r)} - \Gamma^{(r)}\phi_{p}^{H}(\sigma^{2}\mathbf{I}_{P_{b}} + \phi_{p}\Gamma^{(r)}\phi_{p}^{H})^{-1}\phi_{p}\Gamma^{(r)}, \ \mu = \sigma^{-2}\Sigma\phi_{p}^{H}\mathbf{y}_{p}$$

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#### SBL Algorithm for OFDM Channel Estimation

Algorithm 1 SBL for estimating time domain channel taps

Input:  $\mathbf{y}_{p}, \phi_{p}, r_{max}$  and  $\epsilon$ . Initialize  $\Gamma^{(0)} = \mathbf{I}_{L}$ , Set difference = 1, r = 0while (difference>  $\epsilon$  and  $r < r_{max}$ ) E-step:  $\mu = \sigma^{-2} \Sigma \phi_{p}^{H} \mathbf{y}_{p}$   $\Sigma = \Gamma^{(r)} - \Gamma^{(r)} \phi_{p}^{H} (\sigma^{2} \mathbf{I}_{P_{b}} + \phi_{p} \Gamma^{(r)} \phi_{p}^{H})^{-1} \phi_{p} \Gamma^{(r)}$ M-step:  $\gamma^{(r+1)}(i) = \Sigma(i, i) + |\mu|^{2}$  for i = 1, 2, ..., Ldifference  $\triangleq ||\gamma^{(r+1)} - \gamma^{(r)}||_{2}^{2}, r \leftarrow r + 1$  end output:  $\mu, \gamma^{(r)}$ 

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#### Introduction to GNU Radio

- A software development tool kit with **signal processing** blocks written in C++/Python.
- GRC (GNU Radio Companion) is the user interface for GNU Radio.
- Can be used with external RF hardware (such as USRP N210) to create Software Defined Radio(SDR).

#### System architecture



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Implementation in SDR

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## OFDM Chain in GNU Radio



OFDM-Transmitter chain for DATA



OFDM-Receiver chain for DATA

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#### OFDM Transmit Chain in GNU Radio



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## OFDM Receive Chain in GNU Radio



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#### Experiment Results

## Experiment Results

#### System Parameters:

- Number of OFDM sub-carriers (N) = 64
- Cyclic prefix (CP) = 16
- Packet size = 92 Bytes
- Cyclic redundancy check (CRC) = 32 bits = 4 Bytes
- Number of sync words per packet = 2 OFDM symbols
- Header length = 1 OFDM symbol
- Header modulation = BPSK
- Data Modulation = QPSK
- Centre frequency  $(f_c) = 1.1 \text{GHz}$
- Bandwidth (BW) = 500kHz
- Sampling frequency ( $f_s$ ) = 1MS/sec

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#### Real Time Channel



#### Future Work

- Joint data detection and channel estimation algorithms
- Throughput analysis

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