

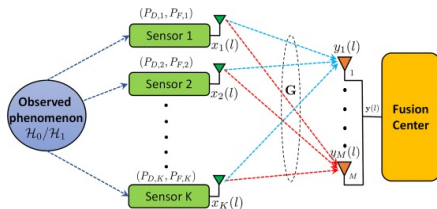
# Journal Watch

## IEEE TSP, Aug 1, 2019

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August 31, 2019

# Distributed Detection in Massive MIMO Wireless Sensor Networks Under Perfect and Imperfect CSI - Apoorva Chawla, Adarsh Patel, Aditya K. Jagannatham, Pramod K. Varshney



## Contributions

- Fusion rules for the low channel SNR regime derived based on linear filtering at the FC which also incorporates the  $P_D$  and  $P_{FA}$  for local sensor decisions.
- Close-form expressions for  $P_D$ ,  $P_{FA}$  for both perfect and imperfect CSI.
- Asymptotic system performance for the large antenna regime.
- The signaling matrices are derived for the WSN to maximize the detection performance at the FC.

## System Model

- The WSN consists of  $K$  single antenna sensors, based on local decision, the  $k$ th sensor transmits  $x_k(1), x_k(2), \dots, x_k(L)$  over  $L$  time instants to a fusion centre having  $M$  antennas such that  $M \gg K$ .

$$\mathbf{y}(l) = \sqrt{p_u} \mathbf{G} \mathbf{x}(l) + \mathbf{n}(l),$$

where,  $n(l) \sim \mathcal{CN}(0, \sigma_n^2 I_M)$ ,  $y(l) \in \mathbb{C}^{M \times 1}$  is the signal received by FC at  $l$ th time instant. The channel coefficient between the  $m$ th antenna and  $k$ th sensor is given by  $g_{mk} = h_{mk} \sqrt{\beta_k}$ .

The resulting channel matrix can be written as  $\mathbf{G} = \mathbf{H} \mathbf{D}^{1/2}$  and the received signal

$$\mathbf{Y} = [y(1), y(2), \dots, y(L)] \in \mathbb{C}^{M \times L}$$

$$\mathbf{Y} = \sqrt{p_u} \mathbf{G} \mathbf{X} + \mathbf{N}$$

$$\frac{1}{M} \mathbf{G}^H \mathbf{G} \approx \mathbf{D}, \quad \text{for } M \gg K.$$

## Fusion Rule with Perfect CSI

$$\mathbf{Z} = \mathbf{G}^H \mathbf{Y} = \sqrt{p_u} \mathbf{G}^H \mathbf{G} \mathbf{X} + \tilde{\mathbf{N}},$$

$$\mathbf{z}_k = \sqrt{p_u} M \beta_k \mathbf{x}_k + \tilde{\mathbf{n}}_k.$$

$$T(\mathbf{Z}) = \ln \left[ \frac{p(\mathbf{Z}|\mathcal{H}_1)}{p(\mathbf{Z}|\mathcal{H}_0)} \right] = \ln \left[ \prod_{k=1}^K \frac{p(\mathbf{z}_k|\mathcal{H}_1)}{p(\mathbf{z}_k|\mathcal{H}_0)} \right],$$

## Antipodal Signalling

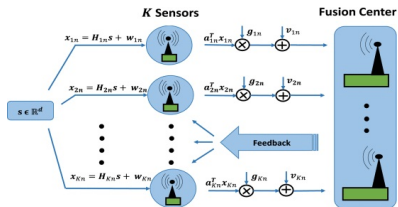
$$T_A(\mathbf{Z}) = \sum_{k=1}^K \ln \left[ \frac{P_{D,k} \exp\left(-\frac{\|\mathbf{z}_k - \sqrt{p_u} M \beta_k \mathbf{u}_k\|^2}{M \beta_k \sigma_n^2}\right) + (1 - P_{D,k}) \exp\left(-\frac{\|\mathbf{z}_k + \sqrt{p_u} M \beta_k \mathbf{u}_k\|^2}{M \beta_k \sigma_n^2}\right)}{P_{F,k} \exp\left(-\frac{\|\mathbf{z}_k - \sqrt{p_u} M \beta_k \mathbf{u}_k\|^2}{M \beta_k \sigma_n^2}\right) + (1 - P_{F,k}) \exp\left(-\frac{\|\mathbf{z}_k + \sqrt{p_u} M \beta_k \mathbf{u}_k\|^2}{M \beta_k \sigma_n^2}\right)} \right]$$

$$T_A(\mathbf{Z}) = \sum_{k=1}^K a_k \mathfrak{R}(\mathbf{z}_k^H \mathbf{u}_k) \underset{\mathcal{H}_0}{\overset{\mathcal{H}_1}{\gtrless}} \gamma_{AP},$$

where,  $a_k = P_{D,k} - P_{F,k}$

# Online Design of Optimal Precoders for High Dimensional Signal Detection

- Prashant Khanduri, Lakshmi Narasimhan Theagarajan, Pramod K. Varshney



## Contributions

- An online algorithm to design precoders (both scalar and vector precoding) for WSNs that perform distributed detection of high-dimensional unknown deterministic signals using spatio-temporal data.
- Analytical expressions for the error exponents of the distributed detector for the proposed precoder design and other state-of-the-art precoder design strategies and show that the proposed precoders are asymptotically optimal.

## System Model

- WSN with  $K$  sensors. Detect the presence or absence of an unknown signal  $s \in \mathbb{R}^d$ .
- Each sensor observes and forwards a precoded version of the observations to a FC at each time instant ( $N$  time instants).

$$\begin{aligned}\mathcal{H}_0 : \mathbf{x}_{kn} &= \mathbf{w}_{kn}, \\ \mathcal{H}_1 : \mathbf{x}_{kn} &= \mathbf{H}_{kn}\mathbf{s} + \mathbf{w}_{kn} \\ \mathbf{y}_{kn} &= \mathbf{G}_{kn}\mathbf{A}_{kn}\mathbf{x}_{kn} + \mathbf{v}_{kn},\end{aligned}$$

where,  $\mathbf{H}_{kn} \in \mathbb{R}^{l_k \times d}$  is the observation matrix,  $\mathbf{G}_{kn} \in \mathbb{R}^{q_k \times q_k}$  is the wireless fading channel matrix and  $\mathbf{A}_{kn} \in \mathbb{R}^{q_k \times l_k}$  denotes the precoding matrix.

$$\begin{aligned}\mathbb{P}(\mathbf{y}_{kn}|\mathcal{H}_i) &\sim \mathcal{N}(\boldsymbol{\mu}_{kn}, \boldsymbol{\Sigma}_{kn}) \\ \boldsymbol{\mu}_{kn} &= \begin{cases} 0 & \text{under } \mathcal{H}_0 \\ \mathbf{T}_{kn}\mathbf{s} & \text{under } \mathcal{H}_1 \end{cases}, \quad \mathbf{T}_{kn} \triangleq \mathbf{G}_{kn}\mathbf{A}_{kn}\mathbf{H}_{kn} \\ \text{and } \boldsymbol{\Sigma}_{kn} &= \boldsymbol{\Sigma}_{v_{kn}} + \mathbf{G}_{kn}\mathbf{A}_{kn}\boldsymbol{\Sigma}_{w_{kn}}\mathbf{A}_{kn}^T\mathbf{G}_{kn}^T.\end{aligned}$$

The ML estimate of  $s$  is given by-

$$\hat{\mathbf{s}} = \left( \sum_{k=1}^K \sum_{n=1}^N \mathbf{T}_{kn}^T \boldsymbol{\Sigma}_{kn}^{-1} \mathbf{T}_{kn} \right)^\dagger \left( \sum_{k=1}^K \sum_{n=1}^N \mathbf{T}_{kn}^T \boldsymbol{\Sigma}_{kn}^{-1} \mathbf{y}_{kn} \right). \quad (6)$$

The probability of detection is computed as below -

$$\beta_D = Q \left( \frac{\sqrt{2d} Q^{-1}(\alpha) - \Lambda_{KN}}{\sqrt{2d + 4\Lambda_{KN}}} \right). \quad (14)$$

where,

$$\Lambda_{KN} = \mathbf{s}^T \left( \sum_{k=1}^K \sum_{n=1}^N \mathbf{T}_{kn}^T \boldsymbol{\Sigma}_{kn}^{-1} \mathbf{T}_{kn} \right) \mathbf{s}. \quad (10)$$

Scalar Case

$$\mathbf{a}_{kn} = \left( \sigma_{v_{kn}}^2 \mathbf{I}_d + \boldsymbol{\Sigma}_{w_{kn}} \right)^{-1} \mathbf{s}/c, \quad (20)$$

With each received sample, the ML estimation error decreases, i.e.,  $s^n \xrightarrow{P} s$  in probability as  $N \rightarrow \infty$ .

$$\begin{aligned} \mathbf{a}_{kn} &= \left( \sigma_{v_{kn}}^2 \mathbf{I}_d + \boldsymbol{\Sigma}_{w_{kn}} \right)^{-1} \hat{\mathbf{s}}_{n-1}/c \\ \hat{\mathbf{s}}_n &= \left( \sum_{k=1}^K \sum_{i=1}^n \mathbf{a}_{ki} \left( \sigma_{v_{ki}}^2 + \mathbf{a}_{ki}^T \boldsymbol{\Sigma}_{w_{ki}} \mathbf{a}_{ki} \right)^{-1} \mathbf{a}_{ki}^T \right)^\dagger \\ &\quad \left( \sum_{k=1}^K \sum_{i=1}^n \mathbf{a}_{ki} \left( \sigma_{v_{ki}}^2 + \mathbf{a}_{ki}^T \boldsymbol{\Sigma}_{w_{ki}} \mathbf{a}_{ki} \right)^{-1} \mathbf{y}_{ki} \right) \end{aligned}$$

# Simultaneous Safe Feature and Sample Elimination for Sparse Support Vector Regression - Hongmei Wang, Xianli Pan, Yitian Xu

## Goal & Contributions

- To reduce the computation complexity of Sparse Support Vector Regression (SSVR) by identifying the inactive features and samples.
- A safe Feature and Sample Screening Rule (FSSR1) is proposed for accelerating the SSVR. Through FSSR1 most inactive features and samples can be discarded before training the model.
- FSSR2 is proposed to continuously discard the remained inactive features and samples during the reduced model training process.

**System Model** Given training set  $T = \{(x_1, y_1), (x_2, y_2), \dots, (x_n, y_n)\}$ , where  $x_i \in \mathbb{R}^p$  and  $y_i \in \mathbb{R}$ . The formulation of SSVR is displayed as follows :

$$\min_w P_{\alpha, \beta}(w) = \frac{\alpha}{2} \|w\|_2^2 + \beta \|w\|_1 + \frac{1}{n} \sum_{i=1}^n l(x_i^T w - y_i), \quad (1)$$

$$l(t) = \begin{cases} 0, & \text{if } |t| < \epsilon, \\ \frac{1}{2\gamma} (|t| - \epsilon)^2, & \text{if } \epsilon \leq |t| \leq \epsilon + \gamma, \\ |t| - \epsilon - \frac{\gamma}{2}, & \text{if } |t| > \epsilon + \gamma, \end{cases}$$



$$\min_{w,z} P_{\alpha,\beta}(w) = \frac{\alpha}{2} \|w\|_2^2 + \beta \|w\|_1 + \frac{1}{n} \sum_{i=1}^n l(z_i)$$

$$\text{s.t. } z = Xw - Y.$$

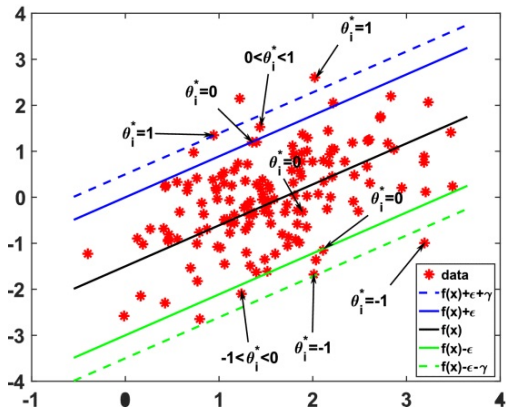
$$L = \frac{\alpha}{2} \|w\|_2^2 + \beta \|w\|_1 + \frac{1}{n} \sum_{i=1}^n l(z_i) + \frac{1}{n} \langle z - Xw + Y, \theta \rangle$$

By differentiating w.r.t  $w$  and equating to 0 -

$$w_j = \frac{1}{\alpha} \begin{cases} \frac{1}{n} \langle x^j, \theta \rangle - \beta, & \text{if } \frac{1}{n} \langle x^j, \theta \rangle > \beta, \\ 0, & \text{if } |\frac{1}{n} \langle x^j, \theta \rangle| \leq \beta, \\ \frac{1}{n} \langle x^j, \theta \rangle + \beta, & \text{if } \frac{1}{n} \langle x^j, \theta \rangle < -\beta. \end{cases}$$

By differentiating w.r.t  $z$  and equating to 0 -

$$\theta_i = \begin{cases} 1, & \text{if } z_i < -\epsilon - \gamma, \\ -\frac{1}{\gamma}(z_i + \epsilon), & \text{if } -\epsilon - \gamma \leq z_i \leq -\epsilon, \\ 0, & \text{if } -\epsilon < z_i < \epsilon, \\ -\frac{1}{\gamma}(z_i - \epsilon), & \text{if } \epsilon \leq z_i \leq \epsilon + \gamma, \\ -1, & \text{if } z_i > \epsilon + \gamma, \end{cases}$$



## Other Interesting Papers

- Self-Tuning Algorithms for Multisensor-Multitarget Tracking Using Belief Propagation.
- Asymptotic Task-Based Quantization With Application to Massive MIMO
- Energy Efficiency Optimization in MIMO Interference Channels: A Successive Pseudoconvex Approximation Approach.
- Nonsubsampled Graph Filter Banks: Theory and Distributed Algorithms.
- Truncated Sequential Non-Parametric Hypothesis Testing Based on Random Distortion Testing.