

Journal Watch
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Discrete Linear Canonical Transform Based on Hyperdifferential Operators (Aykut Koc, Burak Bartan, and Haldun M. Ozaktas)

Achievements:

- ▶ A new approach to define the discrete linear canonical transform (DLCT) by employing operator theory

Basics :

- ▶ $L = \begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} \frac{\gamma}{\beta} & \frac{1}{\beta} \\ -\beta + \frac{\alpha\gamma}{\beta} & \frac{\alpha}{\beta} \end{bmatrix} = \begin{bmatrix} \frac{\alpha}{\beta} & -\frac{1}{\beta} \\ \beta - \frac{\alpha\gamma}{\beta} & \frac{\gamma}{\beta} \end{bmatrix},$
 $AD - BC = 1$ (Unitary transform)

- ▶ LCT as a linear integral transform :

$$C_L f(u) = \sqrt{\beta} e^{-i\frac{\pi}{4}} \int_{-\infty}^{\infty} \exp i\pi(u^2 - 2\beta uu' + \gamma u'^2) f(u') du'$$

- ▶ Every triplet (α, β, γ) corresponds to a different LCT.

Continue

- ▶ $C_{L_1 L_2} f(u) = C_{L_1} C_{L_2} f(u)$ and $C_{L_2 L_1} f(u) = f(u)$, if $L_2 = L_1^{-1}$
- ▶ Scaling : $L_M = \begin{bmatrix} M & 0 \\ 0 & \frac{1}{M} \end{bmatrix}$ $C_{L_M} = M_M f(u) = \sqrt{\frac{1}{M}} f\left(\frac{u}{M}\right)$
- ▶ Fractional Fourier Transform (FRT) : $L_{F_{1c}^a} = \begin{bmatrix} \cos \theta & \sin \theta \\ -\sin \theta & \cos \theta \end{bmatrix}$,
 $\theta = \pi a/2$ and a is the fractional order, $a = 1 \implies$ FRT = FT.
 $F_{1c}^a f(u) = \int_{-\infty}^{\infty} K(u, u') f(u') du'$
- ▶ Chirp Multiplication : $L_{Q_p} = \begin{bmatrix} 1 & 0 \\ -q & 1 \end{bmatrix}$,
 $C_{Q_q} f(u) = Q_q f(u) = \exp(-i\pi q u^2) f(u)$
- ▶ Iwasawa decomposition :
 $C_L = Q_q M_M F_{1c}^a \implies L = L_{Q_p} L_M L_{F_{1c}^a}$, a, q, M be chosen appropriately.

Hyperdifferential forms

- ▶ $Q_q = \exp(-i2\pi q \frac{U^2}{2})$, $M_M = \exp(-i2\pi \ln(M) \frac{UD+DU}{2})$ and $F_{1c}^a = \exp(-ia\pi^2 \frac{U^2+D^2}{2})$
- ▶ $Uf(u) = uf(u)$ and $Df(u) = \frac{1}{i2\pi} \frac{df(u)}{du}$ and $U = FDF^{-1}$
- ▶ abstract operators being replaced by matrix operators,
 $C_L = Q_q M_M F_{1c}^a$
- ▶ $Q_q = \exp(-i2\pi q \frac{U^2}{2})$, $M_M = \exp(-i2\pi \ln(M) \frac{UD+DU}{2})$ and $F_{1c}^a = \exp(-ia\pi^2 \frac{U^2+D^2}{2})$
- ▶ $u = nh$, $U_{m,n} = \begin{cases} \frac{\sqrt{N}}{\pi} \sin(\frac{\pi n}{N}) & \text{for } m = n \\ 0, & \text{for } m \neq n \end{cases}$
- ▶ $D = FUF^{-1}$, $F : DFT$
- ▶ C_L : Unitary/ D, U : Hermitian/ Q_q, M_M, F_{1c}^a : Unitary

Sinusoidal Parameter Estimation from Signed Measurements via Majorization-Minimization Based RELAX† (Jiaying Ren, Tianyi Zhang, Jian Li, and Petre Stoica)

Achievements:

- ▶ present a majorization-minimization (MM) based 1bRELAX algorithm, referred to as 1bMMRELAX, for sinusoidal parameter estimation using signed measurements obtained via one-bit sampling
- ▶ introduce a proper majorizing function and develop the MM procedure for minimizing the negative log-likelihood function for the signed measurements.

Problem formulations:

- ▶ 1 – D sinusoidal signal

$$s_t(\theta) = \sum_{k=1}^K A_k \sin(\omega_k t + \phi_k) = \sum_{k=1}^K a_k \sin(\omega_k t) + b_k \cos(\omega_k t),$$

K : the number of sinusoids, $A_k \in \mathbb{R}^+$: amplitude,
 $\omega_k \in [0, \pi]$: frequency, $\phi_k \in [0, 2\pi]$: phase (of the k -th sinusoid component)

- ▶ $y_n = \text{sign}(s_n(\theta) + e_n - h_n)$, additive noise $e \in \mathbb{R}^N$: iid Gaussian with mean zero and unknown variance σ^2 , $h_n \in \mathbb{R}^N$: known threshold vector.
- ▶ likelihood function of the signed measurements :
$$L_\beta = \prod_{n=0}^{N-1} \Phi\left(y_n \frac{s_n(\theta) - h_n}{\sigma}\right) =$$
$$\prod_{n=0}^{N-1} \Phi\left[y_n \frac{(\sum_{k=1}^K a_k \sin(\omega_k n) + b_k \cos(\omega_k n)) - h_n}{\sigma}\right],$$
- ▶ $\Phi(x)$: cdf of standard normal distribution, unknown parameter : $\beta = [\theta^T, \sigma]^T$

continue ..

- ▶ Goal : estimating the parameter vector β and K based on the signed measurement vector $y \in [-1, 1]^N$
- ▶ Maximum Likelihood Estimation: ML estimate of the parameter vector by minimizing the negative log-likelihood

function: $\hat{\tilde{\beta}} = \arg \min_{\tilde{\beta}} l(\tilde{\beta}) =$
 $\arg \min_{\tilde{\beta}} \prod_{n=0}^{N-1} \Phi[y_n((\sum_{k=1}^K \tilde{a}_k \sin(\omega_k n) + \tilde{b}_k \cos(\omega_k n)) - \lambda h_n)]$

- ▶ $\lambda = \frac{1}{\sigma}$, $\tilde{a}_k = \frac{1}{\sigma} a_k$, $\tilde{b}_k = \frac{1}{\sigma} b_k$, $\tilde{\beta} = [\tilde{\theta}^T, \lambda]^T$, and $\tilde{\theta} = [\tilde{a}_1, \tilde{b}_1, \tilde{\omega}_1, \dots, \tilde{a}_K, \tilde{b}_K, \tilde{\omega}_K]$

Majorization function

- ▶ With an auxiliary vector $x(\tilde{\beta}) = (x_n(\tilde{\beta}))_{n=0}^{N-1} = y_n(s_n(\tilde{\theta}) - \lambda h)_{n=0}^{N-1}$ and $x^i = x(\tilde{\beta}^i)$, the estimate obtained at the i -th MM iteration,

continue

- ▶ $G(\tilde{\beta}|\tilde{\beta}^i) = \tilde{G}(x(\tilde{\beta})|x^i) = \sum_{n=0}^{N-1} f(x_n^i) + f'(x_n^i)(x_n(\tilde{\beta}) - x_n^i) + \frac{1}{2}(x_n(\tilde{\beta}) - x_n^i)^2$, majorizes the objective function
- ▶ computational complexity: 1bRELAX($O(CKN^2 + CK^2N^2)$) where C is the number of iterations required to achieve practical convergence
- ▶ 1bMMRELAX : $O(KN^2)$

Group Greedy Method for Sensor Placement (Chaoyang Jiang, Zhenghua Chen, Rong Su, and Yeng Chai Soh)

Achievements:

Providing necessary and sufficient conditions for convergence of existing greedy algorithms and newly proposed group greedy algorithm for sensor placement

Problem Statement

- ▶ physical field $\zeta = \tilde{\Phi}\alpha \in \mathbb{R}^N$, where $\tilde{\Phi} \in \mathbb{R}^{N \times n}$ with $n < N$
- ▶ aim is to estimate α from

$$y = H(\zeta + \nu) = H\tilde{\Phi}\alpha + H\nu,$$

$y \in \mathbb{R}^M$, M is the number of sensor observations, $H \in \mathbb{R}^{M \times N}$ whose i -th row is $e_{s_i}^T$, $\Phi = H\tilde{\Phi} = [\phi_{s_1}, \dots, \phi_{s_M}]^T \in \mathbb{R}^{M \times n}$, $s_i \in \mathbf{N} = \{1, 2, \dots, N\}$ corresponds to i -th sensing location, $\phi_{s_i}^T$ is the observation vector, noise $\nu \sim \mathcal{N}(0, \sigma^2 I)$

Measure of recovery performance

- ▶ mean squared error (MSE):

$$MSE(\tilde{\alpha}) = \sigma^2 \text{tr}((\Phi^T \Phi)^{-1}) = \sum_{i=1}^n \frac{\sigma^2}{\lambda_i}$$

- ▶ log volume of the confidence ellipsoid (VCE):

$$VCE(\tilde{\alpha}) = \beta - \frac{1}{2} \log \det(\Phi^T \Phi)$$

- ▶ worst case error variance (WCEV)

$$WCEV(\tilde{\alpha}) = \sigma^2 \lambda_{\max}((\Phi^T \Phi)^{-1})$$

- ▶ Goal : error of the estimated $\tilde{\alpha}$ is less than predefined threshold and M is minimized

Combinatorial optimization problem

- ▶ min $M = |S|$
subject to $S \subset \mathbf{N}, f(S) \geq \gamma$
- ▶ $f_{MSE}(S) = \frac{n}{\epsilon} - \text{tr}(\Psi^{-1}) \geq \frac{n}{\epsilon} - \gamma_A, f_{VCE}(S) = \log \det \Psi \geq \gamma_D,$
 $f_{WVEV} = \lambda_n \geq \frac{1}{\gamma_E}$
- ▶ Greedy algorithm : Update stage:
 $s^* = \arg \max_{j \in \mathbf{N} \setminus S} f(S \cup \{j\}), S = S \cup \{s^*\}$
run till constraint is not satisfied
- ▶ Dual Problem:
max $f(S)$
subject to $S \subset \mathbf{N}, |S| = M$
- ▶ A and B be both optimal solutions of dual problem with
 $M_A = |A| < M_B = |B|$
 $A = \arg \max_{|S|=M_A, S \subset \mathbf{N}} f(S)$ and $b = \arg \max_{|S|=M_b, S \subset \mathbf{N}} f(S)$
- ▶ Necessary and sufficient condition: The greedy algorithm can obtain the optimal solution of dual problem iff $\forall M_A M_B,$
 $A \subset B.$

Group greedy Algorithm

- ▶ when determining each sensing location select a group of L suboptimal sensor configurations instead of the current optimal one
- ▶ Update stage :
 $\forall i \in [1 : L], S_{k+1i} = S_{kl_i^*} \cup \{j_i^*\}$ where
 $\{l_i^*, j_i^*\} = \underset{l_i \in [1:M], j_i \in \mathbf{N} \setminus S_{kl_i}}{\operatorname{arg\,rankde}_i} f(S_{kl_i} \cup \{j_i\})$
 $k = k = 1, S = S_{k1}$ and $M = k$
- ▶ 'rankdei' represents an operator which ranks all elements of a set in a descending order and returns the i -th element
- ▶ Necessary and sufficient condition: Denote the optimal solution of dual problem by S_M^* . Let $\mathbb{S}_k = \{S_{k1}, \dots, S_{kL}\}$. Then group greedy algorithm can provide the optimal solution of dual problem iif for every $k \in [1, M - 1]$ there exists $i \in [1, L]$ such that $S_{ki} \subset S_{k+1}^*$

Other papers:

- ▶ Information-Theoretic Pilot Design for Downlink Channel Estimation in FDD Massive MIMO Systems (Yujie Gu ; Yimin D. Zhang)
- ▶ Unimodality-Constrained Matrix Factorization for Non-Parametric Source Localization (Junting Chen ; Urbashi Mitra)
- ▶ Delta-Ramp Encoder for Amplitude Sampling and its Interpretation as Time Encoding (Pablo Martinez-Nuevo ; Hsin-Yu Lai ; Alan V. Oppenheim)
- ▶ Localization from Incomplete Euclidean Distance Matrix: Performance Analysis for the SVD-MDS Approach (Huan Zhang ; Yulong Liu ; Hong Lei)
- ▶ Advanced Low-Complexity Multicarrier Schemes Using Fast-Convolution Processing and Circular Convolution Decomposition (AlaaEddin Y. M. Loulou ; Juha Yli-Kaakinen ; Markku K. Renfors)

Other papers:

- ▶ Correction of Corrupted Columns through Fast Robust Hankel Matrix Completion (Shuai Zhang ; Meng Wang)
- ▶ On the Use of the Z Transform of LTI Systems for the Synthesis of Steered Beams and Nulls in the Radiation Pattern of Leaky-Wave Antenna Arrays (Rafael Verdu-Monedero ; Jose Luis Gomez-Tornero)
- ▶ Semi-Blind Inference of Topologies and Dynamical Processes over Dynamic Graphs (Vassilis N. Ioannidis ; Yanning Shen ; Georgios B. Giannakis)
- ▶ Covariance Matrix Estimation from Linearly-Correlated Gaussian Samples (Wei Cui ; Xu Zhang ; Yulong Liu)
- ▶ New Saddle-Point Technique for Non-Coherent Radar Detection with Application to Correlated Targets in Uncorrelated Clutter Speckle (Josef Alexander Zuk ; Stephen Bocquet ; Luke Rosenberg)

Other papers:

- ▶ Antithetic Dithered 1-bit Massive MIMO Architecture: Efficient Channel Estimation via Parameter Expansion and Pseudo Maximum Likelihood (David K. W. Ho ; Bhaskar D. Rao)
- ▶ Frequency Synchronization for Uplink Massive MIMO with Adaptive MUI Suppression in Angle-domain (Yinghao Ge ; Weile Zhang ; Feifei Gao ; Geoffrey Ye Li)
- ▶ Parameter Estimation of Heavy-Tailed AR Model with Missing Data via Stochastic EM (Junyan Liu ; Sandeep Kumar ; Daniel P. Palomar)
- ▶ Learning To Detect (Neev Samuel ; Ami Wiesel ; Tzvi Diskin)
- ▶ Underdetermined DOA Estimation for Wide-Band Stationary Sources in Unknown Noise Environment (L. Huang ; Q. Zhang ; S. Wu ; So)