

Channel Coding Rate over a Block Fading Channel in the Finite Blocklength Regime

Deekshith P K

Joint work with: Vinod Sharma

Dept. of ECE, IISc, Bangalore

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Outline

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- ▶ Motivation & Setting

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- ▶ Prior & State of Art

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- ▶ Results in Numbers
- ▶ Final Remarks

Era of Short Packets

- ▶ Upcoming wireless systems ought to support novel traffic types (e.g., machine-to-machine)
- ▶ Typically short packets are involved (traffic generated by sensors)
- ▶ **Stringent** reliability and **latency** requirements
- ▶ Need to **refine** classical physical layer **performance metrics** (ergodic capacity, outage capacity etc.)

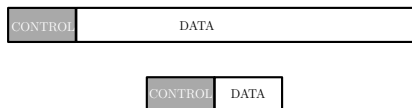


Figure 1: Short packet: Size of metadata and data are comparable

Information Theoretic Metrics

Delay sensitive data over a Wireless Channel

- ▶ Earlier: Delay-limited capacity, outage capacity, average capacity etc.
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- ▶ C – Channel capacity, V – Channel dispersion, $\Phi^{-1}(\cdot)$ – Inverse Gaussian cdf
- ▶ Question: Finite blocklength approximations for fading channels?

Channel Model

- ▶ Fading takes values in $\{\eta_0, \eta_1, \dots, \eta_N\}$, $\eta_i \neq 0$

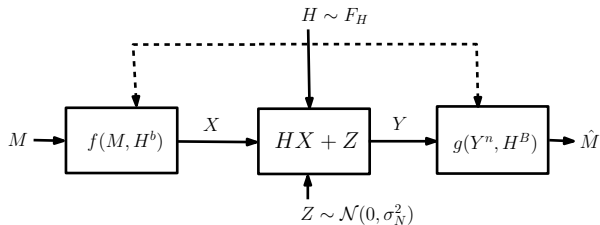


Figure 2: A blockfading channel

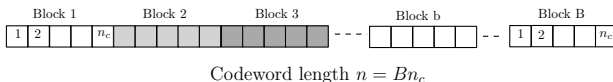


Figure 3: Independent, identically distributed gains across blocks

Code Constraints

- ▶ Average probability of error: $0 < \epsilon < \frac{1}{2}$
- ▶ Power constraint:
 - ▶ Short term (**ST**) constraint:

$$\sum_{b=1}^B \sum_{k=1}^{n_c} X_{[bk]}^2(m, H^b) \leq B n_c \bar{P}, \quad \forall m \in [1 : M].$$

- ▶ **ST** constraint: Corresponds to peak-power limitations of the circuitry

Related Work

- ▶ Dispersion of fading channel with CSIR and no CSIT (Polyanskiy and Verdú, ISIT'11)
- ▶ Second order coding rate of a DMC with finite states and with non causal CSIT and CSIR (Tomamichel & Tan, T-IT, Aug'14)
- ▶ Back off from outage capacity of block fading channel with a single fading block with CSIT and CSIR (Yang et al., T-IT Sep'15)

Channel Capacity & Optimal Power Allocation

- ▶ Capacity:

$$\mathbf{C}(\bar{P}) = \mathbb{E}_H \left[\frac{1}{2} \log \left(1 + \frac{H^2 \mathcal{P}_{\text{WF}}(H)}{\sigma_N^2} \right) \right]$$

- ▶ Water filling power allocation:

$$\mathcal{P}_{\text{WF}}(H) = \left(\lambda - \frac{\sigma_N^2}{H^2} \right)^+, \quad \mathbb{E}_H \left[\mathcal{P}_{\text{WF}}(H) \right] = \bar{P}$$

Question: Given a codeword length n , average probability of error ϵ and power constraint \bar{P} , what is the back off from $\mathbf{C}(\bar{P})$?

Result

Lower bound on maximal coding rate $R^*(n, \bar{P}, \epsilon)$

- ▶ **ST** constraint, i.e., $\|\mathbf{X}\|^2 \leq n\bar{P}$:

$$R^*(n, \bar{P}, \epsilon) \geq \mathbf{C}(\bar{P}) + \sqrt{\frac{V_{\text{BF}}(\bar{P})}{n}} \Phi^{-1}(\epsilon) - \frac{c_\epsilon(\bar{P})}{\sqrt{n}} + \frac{\log n}{2n} + O\left(\frac{1}{n}\right)$$

$$V_{\text{BF}}(\bar{P}) \triangleq \mathbb{E}\left[V(G)\right] + n_c \text{Var}\left[C(G)\right] + \frac{1}{2} \text{Var}\left[\mathcal{L}(G)\right]$$

$$\mathcal{L}(x) = \frac{\sigma_N^2}{x + \sigma_N^2}, V(x) = \frac{1 - \mathcal{L}^2(x)}{2}, C(x) = \frac{1}{2} \log\left(1 + \frac{x}{\sigma_N^2}\right), G = H^2 \mathcal{P}_{\text{WF}}(H)$$

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$$V_{\text{BF}}'(\bar{P}) \triangleq \mathbb{E}_G[V(G)] + n_c \text{Var}\left[C(G) + \frac{\bar{P}}{2\lambda} - \frac{\mathcal{L}(G)}{2}\right]$$

Obtaining Lower Bound

1) Decouple Coding & Power Control

- ▶ Sample codewords *uniformly* from $\{\mathbf{x}' : \|\mathbf{x}'\|^2 = n(1 - \delta_n)\}$
- ▶ Power control: $X_k = \sqrt{\mathcal{P}_{\text{WF}}(H_k)} X'_k$
- ▶ **ST** constraint violation at time k : If power transmitted till $k > n\bar{P}$

$$\begin{aligned} \mathbb{P}[\text{violating } \mathbf{ST} \text{ constraint}] &= \mathbb{P}\left[\bigcup_{b=1}^B \left\{ \sum_{l=1}^b \|\underline{X}'_l\|^2 \mathcal{P}_{\text{WF}}(H_l) > Bn_c \bar{P} \right\}\right] \\ &\leq c_1 \exp\left(-c_2 n \delta_n^2\right) = \epsilon_n^* < \epsilon. \end{aligned}$$

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- ▶ Probability of error in decoding: $\epsilon - \epsilon^*$

Obtaining Lower Bound

2) The $\beta - \beta$ bound

- ▶ The $\beta - \beta$ bound [Yang et al., 2016]: For any $\tau \in (0, \epsilon')$, $\epsilon' = \epsilon - \epsilon^*$, any auxiliary channel \mathbb{Q} ,

$$M^* \geq \frac{\beta_{\tau}(\tilde{\mathbb{P}}_1, \tilde{\mathbb{P}}_2)}{\beta_{1-\epsilon'+\tau}(\mathbb{P}_1, \mathbb{P}_2)},$$

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$\beta_\alpha(\mathbb{P}_A, \mathbb{P}_B)$: Minimum *false alarm* probability in deciding \mathbb{P}_A against \mathbb{P}_B , with minimum detection probability α

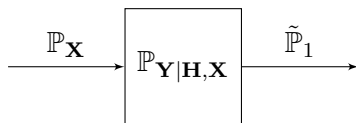
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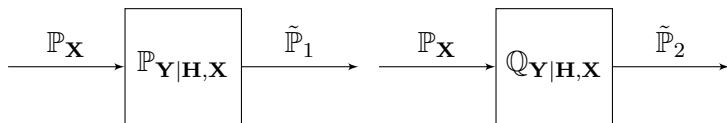
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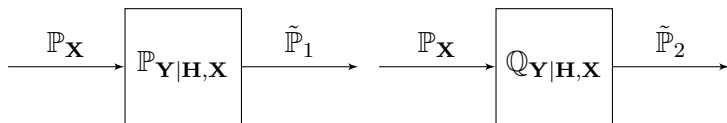


Figure 4: Joint distributions \mathbb{P}_1 and \mathbb{P}_2 resp.

Obtaining Lower Bound

3) Choosing \mathbb{Q} and bounding β terms

- ▶ Fix auxiliary channel $\mathbb{Q}_{\mathbf{Y}|\mathbf{X},\mathbf{H}} \equiv \prod_{b=1}^B \prod_{k=1}^{n_c} \mathcal{N}(0, \sigma_N^2 + H_b^2 \mathcal{P}_{\text{WF}}(H_b))$
- ▶ Lower bound on $\log \beta_\tau(\tilde{\mathbb{P}}_1, \tilde{\mathbb{P}}_2)$:

$$\log \beta_\tau(\tilde{\mathbb{P}}_1, \tilde{\mathbb{P}}_2) \geq \tau^{-1} \left[-1 - D(\tilde{\mathbb{P}}_1 \| \tilde{\mathbb{P}}_2) \right] = O(1).$$

- ▶ Upper bound on $\beta_{1-\epsilon'+\tau}(\mathbb{P}_1, \mathbb{P}_2)$: For appropriate choice of γ_0 ,

$$\log \beta_{1-\epsilon'+\tau}(\mathbb{P}_1, \mathbb{P}_2) \leq \log \mathbb{P}_2 \left[\frac{d\mathbb{P}_1}{d\mathbb{P}_2} \geq \gamma_0 \right] \leq -\frac{\log n}{2} - \log \gamma_0 + \text{constant}.$$

Obtaining Upper Bound

1) Meta Converse & Choice of Auxiliary Channel

- ▶ Assume non causal CSIT.
- ▶ $\mathcal{C}(M, n, \epsilon, \bar{P})$: code satisfy **ST** constraint with equality.
- ▶ Fix auxiliary channel $\mathbb{Q}_{\mathbf{Y}|\mathbf{X},\mathbf{H}} \equiv \prod_{b=1}^B \prod_{k=1}^{n_c} \mathcal{N}(0, \sigma_N^2 + H_b^2 \mathcal{P}_{\text{WF}}(H_b))$.
- ▶ Avg. prob. of error of $\mathcal{C}(M, n, \epsilon, \bar{P})$ over \mathbb{Q} : $\frac{M-1}{M}$
- ▶ Meta-converse and its relaxation: For any \mathbb{Q} , $\gamma > 0$,

$$M \leq \frac{1}{\beta_{1-\epsilon}(\mathbb{P}_{\mathbf{X},\mathbf{H},\mathbf{Y}}, \mathbb{Q}_{\mathbf{X},\mathbf{H},\mathbf{Y}})} \leq \frac{\gamma}{\mathbb{P}[I_\gamma] - \epsilon},$$

$$\mathbb{P}[I_\gamma] = \mathbb{P}\left[\frac{d\mathbb{P}_{\mathbf{X},\mathbf{H},\mathbf{Y}}}{d\mathbb{Q}_{\mathbf{X},\mathbf{H},\mathbf{Y}}} \leq \gamma\right]$$

Obtaining Upper Bound

2) Lower bound on $\mathbb{P}[\mathcal{I}_\gamma]$: Two step approximation

- ▶ View as a parallel channel with N sub channels
 - ▶ Constraint set $\subset \mathbb{R}^N$ (instead of \mathbb{R}^n)
- ▶ Using Berry Esseen, for some $\mu, \nu, c_3 > 0$

$$\mathbb{P}[\mathcal{I}_\gamma] \geq \mathbb{E}_{\mathbf{H}} \left[\Phi \left(\frac{\log \gamma - \mu(\mathbf{H})}{\sqrt{\nu(\mathbf{H})}} \right) \right] - \frac{c_3}{\sqrt{n}}$$

- ▶ $\mathbb{E}_{\mathbf{H}} \left[\Phi \left(\frac{\log \gamma - \mu(\mathbf{H})}{\sqrt{\nu(\mathbf{H})}} \right) \right] \geq \Phi \left(\frac{\log \gamma - n\mathbf{C}(\bar{P})}{\sqrt{nV'_{\text{BF}}(\bar{P})}} \right) - \frac{c_4}{\sqrt{n}}$
 - ▶ Infinite Taylor series, concentration bounds, Berry Esseen
- ▶ Choosing $\gamma = n\mathbf{C}(\bar{P}) + \sqrt{nV'_{\text{BF}}(\bar{P})} \Phi^{-1}(\epsilon + \frac{2c_4}{\sqrt{n}})$, the result follows.

Numerical example

Finite blocklength approximation: Rate versus power

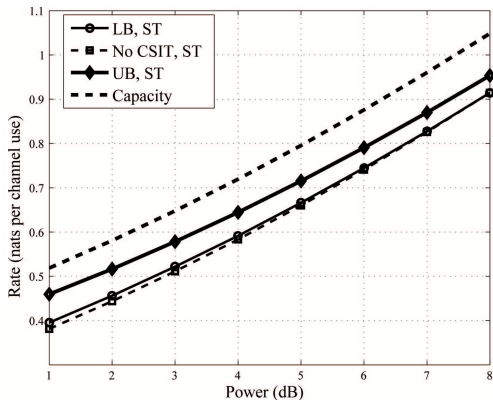


Figure 5: Rate versus power

- Fading distribution: A quantized version of Rayleigh distribution with parameter 0.75, 10 mass points

Concluding remarks

- ▶ Result holds for for general fading state space, long term power constraint
- ▶ Quest for matching second order coefficients...

