

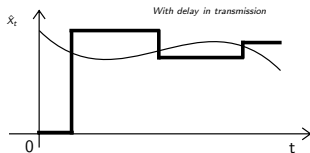
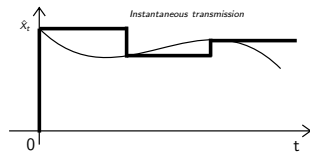
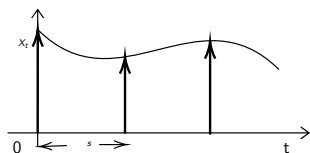
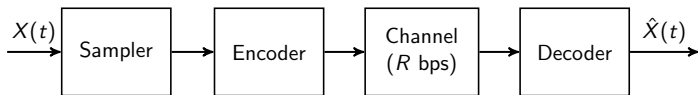
Tracking AR(1) Process with limited communication

Parimal Parag

Joint work with Rooji Jinan and Himanshu Tyagi

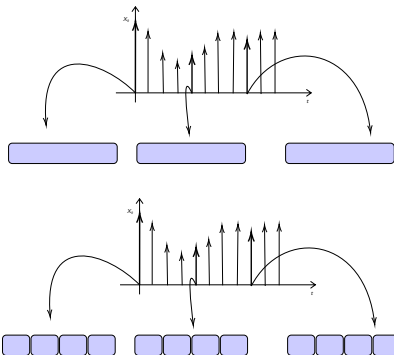


Remote real-time tracking



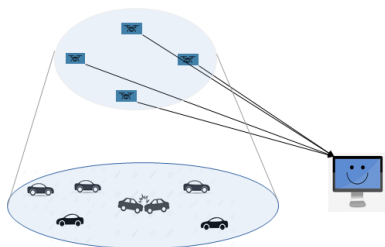
Fast or Precise?

- ▶ What is the optimal strategy for real-time tracking of a discrete time process under periodic sampling?
- ▶ **Slow and precise** or **Fast but loose**



Application

- ▶ Many cyber-physical systems often employ tracking of sensor data in real time
- ▶ Examples: sensing, surveillance, real-time control, ...



- ▶ Communication is limited by the following constraints:
 - ▶ Cost of frequent sampling
 - ▶ Limited channel resources

Existing Works

Sequential coding for correlated sources

- ▶ Rate-distortion region characterization [**Viswanathan2000TIT**]
- ▶ Real-time encoding for Gauss-Markov source [**Khina2017ITW**]

Remote estimation under communication constraints

- ▶ Real-time estimation of Wiener process [**Sun2017ISIT**]
- ▶ Real-time estimation of AR source [**Chakravorty2017TAC**]

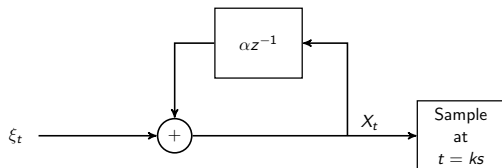
Sequential lossy coding for sequences under delay constraints

- ▶ Zero delay lossy coding [**Linder2001TIT**]
- ▶ Limited delay lossy coding [**Weissman2002TIT**]

Current setting

- ▶ Rate-limited channel with unit delay per channel use
- ▶ Real-time estimation of AR(1) process

Source Process

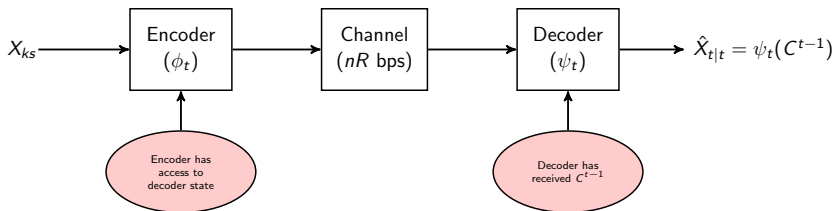


- ▶ Innovation process $\xi_t \in \mathbb{R}^n$ is *i.i.d.* and n -dimensional
- ▶ Discrete AR(1) n -dimensional source process

$$X_t = \alpha X_{t-1} + \xi_t \quad \text{for all } t \geq 0$$

- ▶ Source process X_t is sub-sampled at $1/s$, to obtain samples X_{ks} at $t = ks$

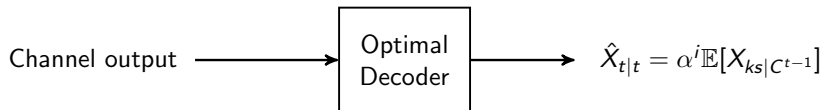
Communication Setting



- ▶ **Encoder:** $\phi_t : \mathcal{X}^{k+1} \rightarrow \{0, 1\}^{nR_s}$ at $t = ks$
- ▶ **Channel:** Error free, limited capacity causes delayed transmission
- ▶ **Decoder:** $\psi_t : \{0, 1\}^{nR_t} \rightarrow \mathcal{X}$ at $t = ks$
- ▶ **Performance metric:**

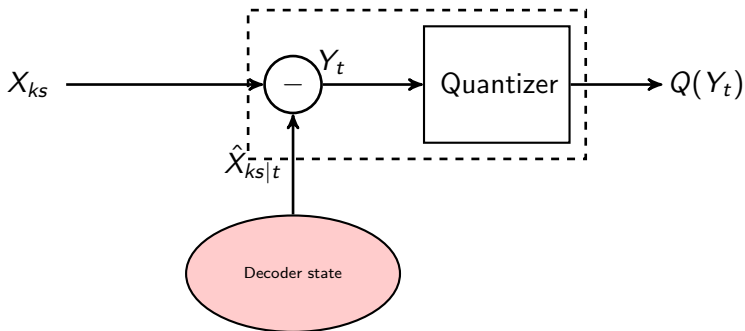
$$D_t(\phi, \psi, X) \triangleq \frac{1}{n} \mathbb{E} \|X_t - \hat{X}_{t|t}\|_2^2.$$

Optimal Decoder Structure



- ▶ Decoder at time $t = ks + i$ for $i \in \{1, \dots, s\}$
- ▶ For the mean squared error, estimate conditional mean
- ▶ Utilize the latest information to refine the last sample X_{ks}

Encoder Structure

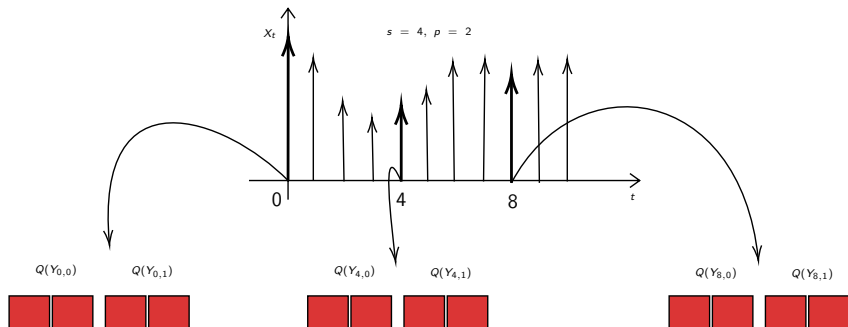


- ▶ Find the error in the decoder estimate of the last sample
- ▶ Transmit the quantized error

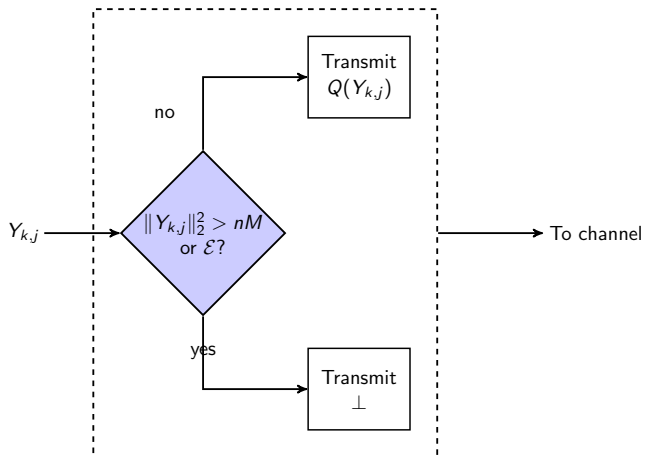
Periodic Successive Update Scheme

- ▶ Divide the sampling interval of duration s in smaller intervals of duration p
- ▶ For any time $t = ks + jp$, for $0 \leq j < s/p$, encode

$$Y_{k,j} = X_{ks} - \hat{X}_{ks|ks+jp}$$



Encoder at time $t = ks + jp$



\mathcal{E} is the codebook failure event

(θ, ε) -quantizer

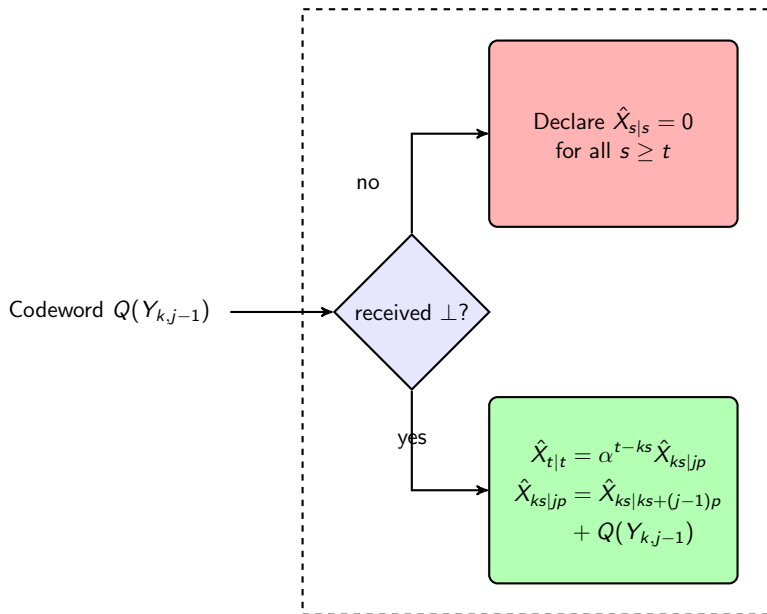
Definition

Fix $0 < M < \infty$. A quantizer $Q : \mathbb{R}^n \rightarrow \{0, 1\}^{nR}$ constitutes an nR bit (θ, ε) -quantizer if for every vector $y \in \mathbb{R}^n$ such that $\frac{1}{n}\|y\|_2^2 \leq M^2$, we have

$$\frac{1}{n}\mathbb{E}\|y - Q(y)\|_2^2 \leq \frac{1}{n}\|y\|_2^2\theta(R) + \varepsilon^2.$$

for $0 \leq \theta \leq 1$ and $0 \leq \varepsilon$.

Decoder at time $t = ks + jp + i$



Performance of Periodic Successive Update Scheme

Lemma

For a fixed time horizon T , periodic successive update scheme with a (θ, ϵ) quantizer gives

$$\frac{1}{T} \sum_{t=0}^T D_t(\phi_p, \psi_p, \mathbf{X}) \leq \sigma^2 \left[1 - \frac{g(s) \alpha^{2p}}{1 - \alpha^{2p} \theta(Rp)} \left(1 - \frac{\epsilon^2}{\sigma^2} - \theta(Rp) \right) \right]$$

for a very low probability of encoder failure and $g(s) \triangleq \frac{1 - \alpha^{2s}}{s(1 - \alpha^2)}$.

Average Distortion Upper Bound for Gain-Shape Quantizer

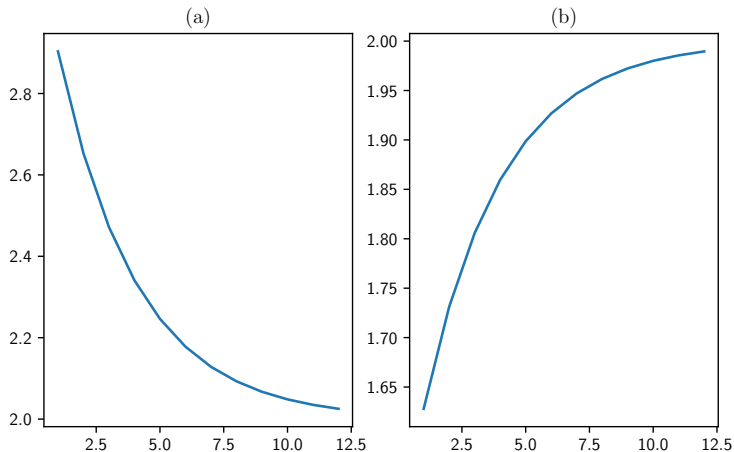
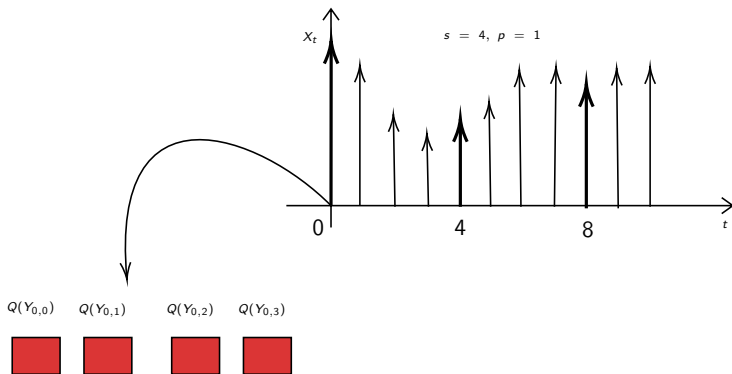


Figure: (a) gives a case where $p = s$ is the best and in (b) $p = 1$ minimizes the bound

Special Case: Successive Update scheme

- ▶ Fast and Loose
- ▶ Set $p = 1$



Performance of the scheme

Lemma

Let $t = ks + i$, for $i \in [1, s]$, for n sufficiently large, the successive update scheme used with a (θ, ϵ) quantizer realisation with $\theta(R) = 2^{-2R}$ satisfies

$$D_t(\phi, \psi, X) \leq \alpha^{2i} 2^{-2Ri} D_{ks}(\phi, \psi, X) + \sigma^2(1 - \alpha^{2i}) + f_n$$

where $f_n \rightarrow 0$ for large n .

Optimum min-max tracking accuracy

Definition

We define the accuracy

$$\delta^T(\phi, \psi, \mathbb{X}_n) \triangleq 1 - \frac{\frac{1}{T} \sum_{t=0}^{T-1} D_t(\phi, \psi, \mathbf{X})}{\sigma^2}$$

The optimum asymptotic maxmin tracking accuracy is defined

$$\delta^*(R, s, \mathbb{X}) \triangleq \lim_{T \rightarrow \infty} \lim_{n \rightarrow \infty} \left[\sup_{(\phi, \psi)} \inf_{\mathbf{X} \in \mathbb{X}_n} \delta^T(\phi, \psi, \mathbb{X}_n) \right].$$

Main Result: Lower Bound

Theorem (Lower bound for maxmin tracking accuracy: The achievability)

For $R > 0$ and $s \in \mathbb{N}$, the asymptotic minmax tracking accuracy is bounded below as

$$\delta^*(R, s, \mathbb{X}) \geq \delta_0(R)g(s).$$

for $\delta_0(R) \triangleq \frac{\alpha^2(1-2^{-2R})}{(1-\alpha^2)2^{-2R}}$ and $g(s) \triangleq \frac{(1-\alpha^{2s})}{s(1-\alpha^2)}$ for all $s > 0$.

This bound is achieved using successive update scheme for $p = 1$ and a specific (θ, ϵ) quantizer.

Main Result: Upper Bound

Theorem (Upper bound for maxmin tracking accuracy: The converse)

For $R > 0$ and $s \in \mathbb{N}$, the asymptotic minmax tracking accuracy is bounded above as

$$\delta^*(R, s, \mathbb{X}) \leq \delta_0(R)g(s).$$

The upper bound is obtained by considering the Gauss-Markov Processes.

Conclusion

- ▶ We provide an information theoretic upper bound for maxmin tracking accuracy for a fixed rate and sampling frequency.
- ▶ It is shown that for a fixed rate, high dimensional setting, the strategy of being *fast but loose* achieves this bound.
- ▶ We outline the performance requirements of the quantizer needed for achieving the optimal performance.
- ▶ For non-asymptotic regime our studies show that the optimal strategy might differ.